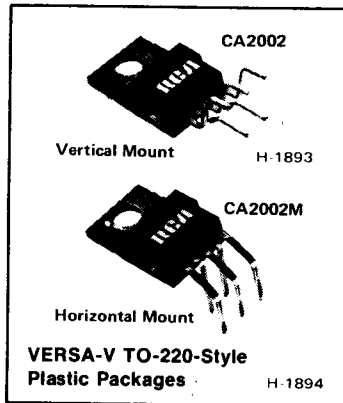


# Linear Integrated Circuits

## CA2002, CA2002M



## 8-Watt Audio Power Amplifier

Especially suited for automobile and other mobile applications

### FEATURES:

- Output short-circuit and thermal overload protection
- Drives load impedance as low as  $1.6\Omega$
- Load dump voltage surge protection
- Output current capability of up to 3.5A
- Few external components
- Versa-V power transistor package-requires no electrical insulation

The RCA-CA2002 is a monolithic silicon class B audio power amplifier designed for driving loads as low as  $1.6\Omega$ . It provides a high output current capability (up to 3.5A), very low harmonic and cross-over distortion, and load-dump voltage-surge protection.

The maximum operating supply-voltage of the CA2002 is 18 V, and internal protection is provided for peaks of up to 40 V, as shown in Fig. 18. Supply-voltage peaks of more than 40 V will require an LC network between the supply and terminal 5. An LC network, such as the one shown in Fig. 18, provides protection against supply-voltage surges of up to 120 V for 2 ms. This type of protection is ON when the supply voltage (pulsed or dc exceeds 18 V).

Thermal shut-down occurs if the output overloads (temporary or permanent), the ambient temperature is

excessive, or the junction temperature is excessive. None of these conditions results in device damage. They merely cause a temporary automatic reduction of output power and drain current, as shown in Figs. 16 and 17.

A heater fan motor run-down hysteresis circuit is included for automotive applications. Typical starting voltage is 10 volts; typical drop-out voltage is 6.5 volts.

The CA2002 is supplied in a 5-lead plastic TO-220-style VERSA-V package. All leads (except term. 3) are electrically insulated from the mounting flange, eliminating the need for insulating hardware. The VERSA-V package is available with two lead configurations. The CA2002 has a vertical-mount lead form, and the CA2002M has a horizontal-mount lead form.

### MAXIMUM RATINGS, Absolute-Maximum Values:

PEAK SUPPLY VOLTAGE (50 ms)	40 V
DC SUPPLY VOLTAGE	28 V
OPERATING SUPPLY VOLTAGE	18 V
OUTPUT PEAK CURRENT:	
REPETITIVE	3.5 A
NON-REPETITIVE	4.5 A
POWER DISSIPATION, $P_D$ at $T_A = 90^\circ\text{C}$	15 W
THERMAL RESISTANCE, JUNCTION TO CASE	$4^\circ\text{C/W}$
AMBIENT-TEMPERATURE RANGE:	
OPERATING	See Figures 16 & 17
STORAGE	$-40$ to $+150^\circ\text{C}$
LEAD TEMPERATURE (DURING SOLDERING):	
At distance $1/16 \pm 1/32$ inch ( $1.59 \pm 0.79$ mm)	
from case for 12 s max.	$260^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$ ,  $V^+ = 14.4\text{ V}$**   
Unless otherwise specified (See Figure 2)

CHARACTERISTIC	TEST CONDITIONS	LIMITS			UNITS	
		Min.	Typ.	Max.		
Supply Voltage, $V^+$		11	—	18	V	
Quiescent Output Voltage, $V_O$	Measure at Term. 4	6.4	7.2	8	V	
Quiescent Drain Current, $I_D$	Measure at Term. 5	—	45	80	mA	
Output Power, $P_O$	THD = 10%, A = 40 dB, f = 1 KHz  $V^+ = 14.4\text{ V}$ $V^+ = 16\text{ V}$	$R_L = 4\ \Omega$	4.8	5.2	—	W
		$R_L = 2\ \Omega$	7	8	—	
		$R_L = 4\ \Omega$	—	6.5	—	
		$R_L = 2\ \Omega$	—	10	—	
Input Saturation Voltage, $V_{I(RMS)}$		400	—	—	mV	
Input Sensitivity, $e_i$	A = 40 dB, f = 1 KHz  $P_O = 0.5\text{ W}, R_L = 4\ \Omega$ $P_O = 0.5\text{ W}, R_L = 2\ \Omega$ $P_O = 5.2\text{ W}, R_L = 4\ \Omega$ $P_O = 8\text{ W}, R_L = 2\ \Omega$		—	15	—	mV
			—	11	—	
			—	55	—	
			—	50	—	
Frequency Response (-3 dB)	$R_L = 4\ \Omega$ (See Fig. 19)	25000			Hz	
Input Resistance, $R_I$ (Term. 1)	f = 1 KHz	70	150	—	K $\Omega$	
Open-Loop Voltage Gain, $A_{OL}$	$R_L = 4\ \Omega$ , f = 1 KHz	—	80	—	dB	
Closed-Loop Voltage Gain, A	$R_L = 4\ \Omega$ , f = 1 KHz	39.5	40	40.5	dB	
Input Noise Voltage, $e_N$	Freq. Resp. = 40 to 15,000 Hz (-3 dB)	400	4	—	$\mu\text{V}$	
Input Noise Current, $i_N$	Freq. Resp. = 40 to 15,000 Hz (-3 dB)	—	60	—	pA	
Efficiency, $\eta$	A = 40 dB, f = 1 KHz  $P_O = 5.2\text{ W}, R_L = 4\ \Omega$ $P_O = 8\text{ W}, R_L = 2\ \Omega$		—	68	—	%
			—	58	—	
Power Supply Rejection Ratio, PSRR	$R_L = 4\ \Omega$ , A = 40 dB, $R_g = 10\text{ K}\Omega$ , $f_{\text{ripple}} = 100\text{ Hz}$ , $V_{\text{ripple}} = 0.5\text{ V}$	30	35	—	dB	

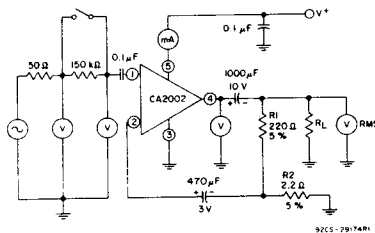


Fig. 1 - Test circuit.

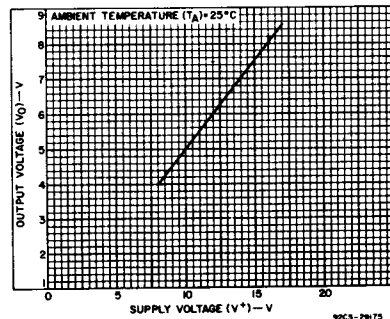


Fig. 2 - Typical quiescent output voltage as a function of supply voltage.

# Linear Integrated Circuits

## CA2002, CA2002M

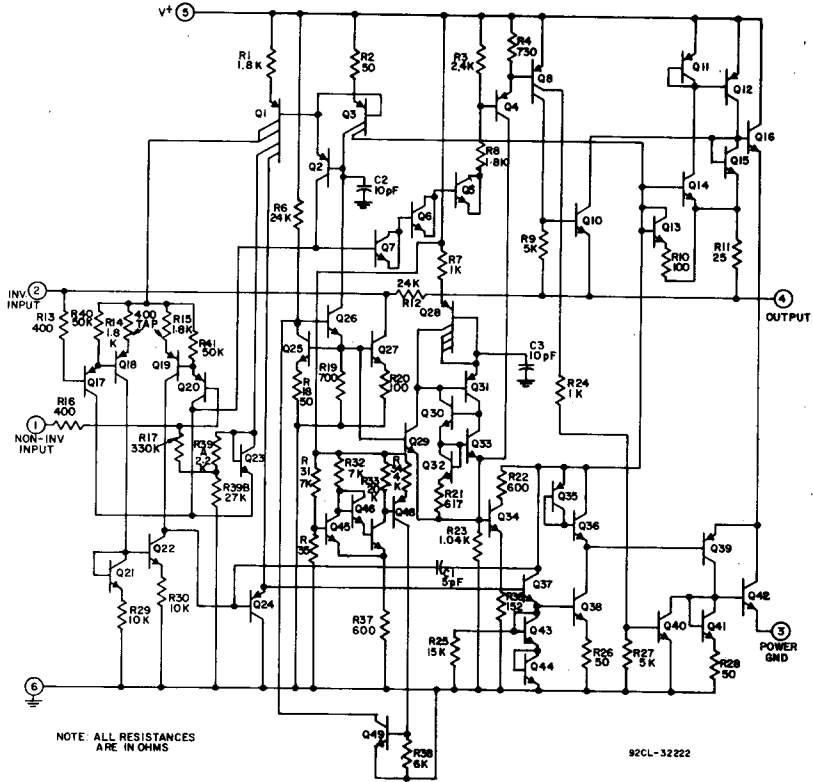


Fig. 3 - Schematic diagram.

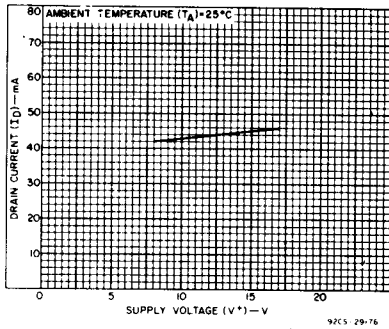


Fig. 4 - Typical quiescent drain current as a function of supply voltage.

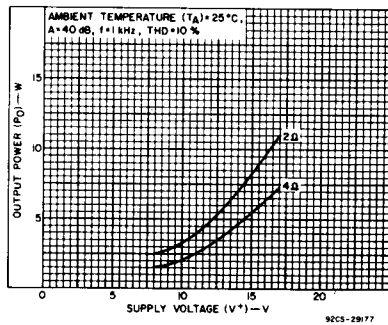


Fig. 5 - Typical output power as a function of supply voltage.

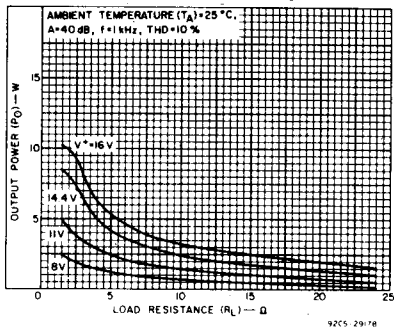


Fig. 6 - Typical output power as a function of load resistance.

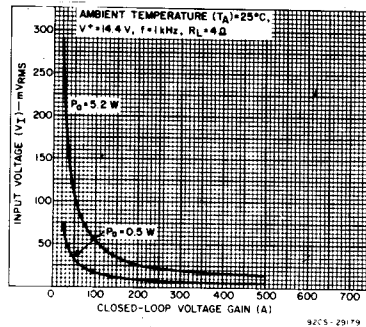


Fig. 7 - Typical input voltage as a function of closed-loop voltage gain.

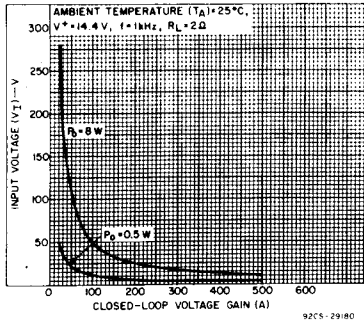


Fig. 8 - Typical input voltage as a function of closed-loop voltage gain.

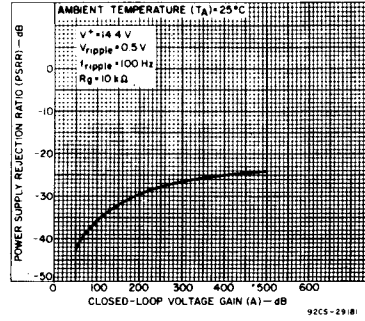


Fig. 9 - Typical power supply rejection ratio as a function of closed-loop voltage gain.

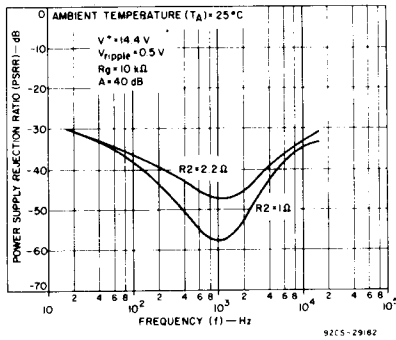


Fig. 10 - Typical power supply rejection ratio as a function of frequency.

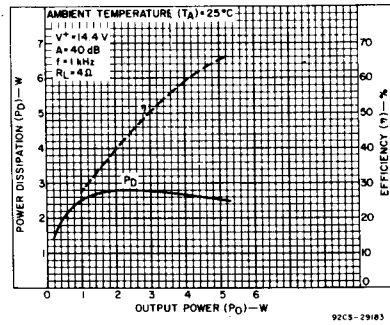


Fig. 11 - Typical power dissipation and efficiency as a function of output power.

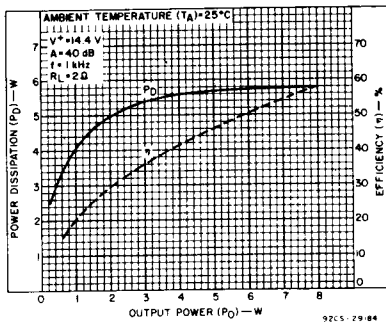


Fig. 12 - Typical power dissipation and efficiency as a function of output power.

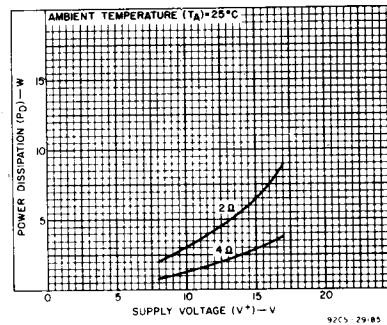


Fig. 13 - Maximum power dissipation as a function of supply voltage (sine-wave operation).

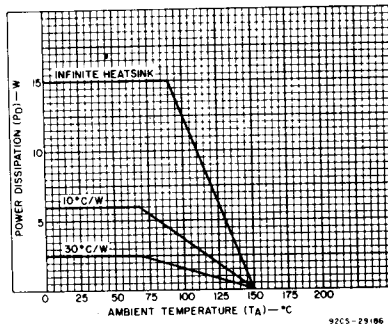


Fig. 14 - Maximum allowable power dissipation as a function of ambient temperature.

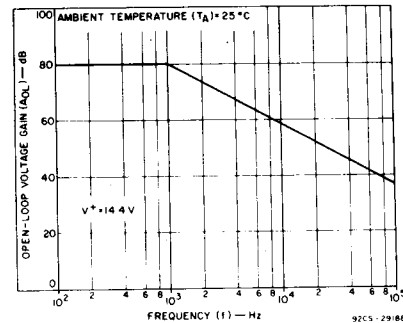


Fig. 15 - Open-loop voltage gain as a function of frequency.

# Linear Integrated Circuits

## CA2002, CA2002M

### Load-Dump Voltage-Surge Protection

The maximum operating supply-voltage of the CA2002 is 18 V, and internal protection is provided for peaks of up to 40 V, as shown in Fig. 18. Supply-voltage peaks of more than 40 V will require an LC network between the supply and terminal 5. An LC network, such as the one shown in Fig. 18, provides protection against supply-voltage surges of up to 120 V for 2 ms. This type of protection is ON when the supply voltage (pulsed or

dc) exceeds 18 V.

### Thermal Shut-Down

Thermal shut-down occurs if the output overloads (temporary or permanent), the ambient temperature is excessive, or the junction temperature is excessive. None of these conditions results in device damage. They merely cause a temporary automatic reduction of output power and drain current, as shown in Figs. 16 and 17.

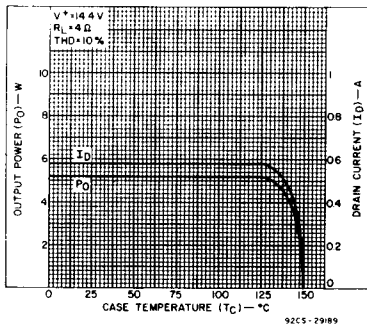


Fig. 16 - Output power and drain current as a function of case temperature.

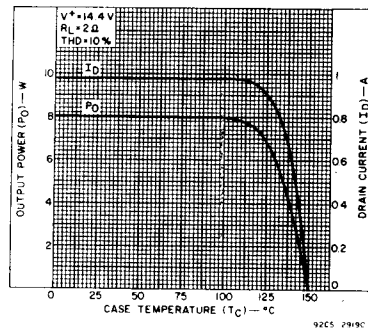


Fig. 17 - Output power and drain current as a function of case temperature.

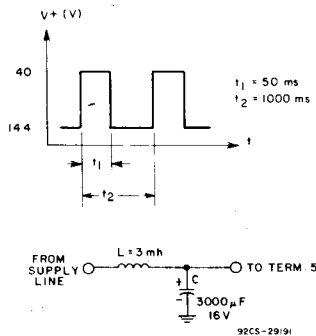


Fig. 18 - Supply-voltage surge protection network and timing diagram.

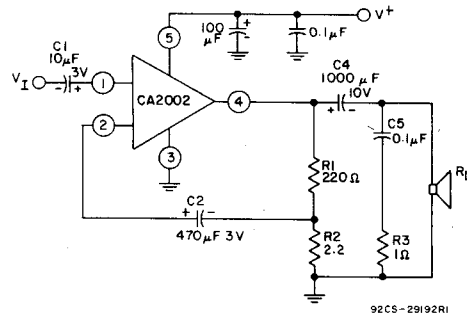


Fig. 19 - Typical application.

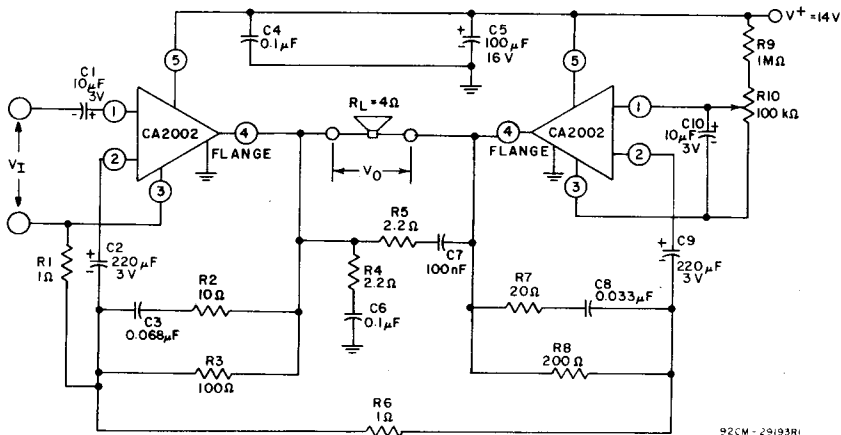


Fig. 20 - 15 W circuit-bridge application.