

# **FAST CMOS CCD DRIVER**

## **FEATURES**

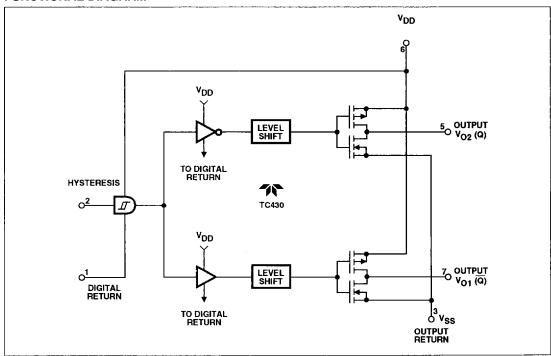
Operating Range	$4.5V \le (V_{DD} - V_{SS}) \le 12V$
TTL/CMOS-Compatible	Inputs
Low Delay Time	15 ns Typ
Rise and Fall Times	2200 pF Load, 25 ns Typ
Peak Output Current	3A

- Output Can Be Floated Below Digital Return
- Level Shifting for Split-Supply Operation
- Guaranteed Skew
- Complementary Outputs
- 10 MHz Operation With Adequate Heat Sink
- Drives 1000 pF at 4 MHz, in CerDIP With No External Heat Sink (10V V<sub>DD</sub> - V<sub>SS</sub>)
- Low Output Impedance.....5Ω Max
- Low Quiescent Current......5 mA Max

#### **APPLICATIONS**

- **CCD Driver**
- **MOSFET Driver**
- Laser Diode Driver
- Differential Line Driver
- **PIN Diode Driver**
- **Level Shifting Driver**

## **FUNCTIONAL DIAGRAM**



## **TC430**

#### **GENERAL DESCRIPTION**

The TC430 is a super-fast CMOS power driver for driving CCDs and other loads. The TC430 operates at frequencies to 10 MHz and drives loads greater than 2200 pF. Peak current output is 3A.

The input is TTL/CMOS compatible. Digital return and output return can be at different voltages, allowing operation with output swings between positive and negative supplies without sacrificing AC performance when driven from TTL. The ability to swing negative is important when driving CCD devices.

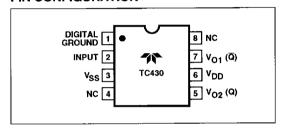
The output stages have been designed so the rising edge of one output crosses the 50% point of the transition within 5 ns of the other. This makes the TC430 ideal for driving CCDs and achieving high contrast images.

CMOS construction achieves low quiescent power (less than 5 mA at 15V and 25°C) and low input current requirements. This device requires fewer external components than bipolar devices like the DS0026 which need external speed-up capacitors.

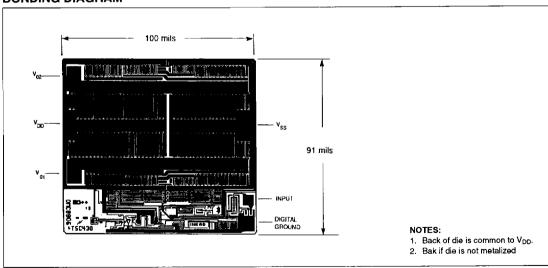
#### **ORDERING INFORMATION**

Part No.	Package	Temperature Range
TC430CPA	8-Pin Plastic	0°C to +70°C
TC430IJA	8-Pin CerDIP	-25°C to +85°C
TC430MJA	8-Pin CerDIP	-55°C to +125°C

#### PIN CONFIGURATION



#### **BONDING DIAGRAM**



## **ABSOLUTE MAXIMUM RATINGS**

Power Dissipation at +25°C	
Plastic	1000 mW
CerDIP	800 mW
Derating Factors	
Plastic	8 mW/°C
CerDIP	6.4 mW/°C
Supply Voltage	$V_{DD} - V_{SS} \le 15V$
	$V_{DD} - V_{DR} \leq 15V$
Input Voltage, Any Terminal	V <sub>DD</sub> +0.3V
Operating Temperature Range	
M Version	55°C to +125°C
Version	25°C to +85°C
C Version	0°C to +70°C
Maximum Chip Temperature	150°C

Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 10 sec).	300°C
CerDIP Re-JA	150°C/W
Plastic Re-JA	125°C/W
ESD Protection	2000V

Static-sensitive device. Unused devices must be stored in conductive material. Protect devices from static discharge and static fields. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to Absolute Maximum Rating Conditions for extended periods may affect device reliability.

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# **ELECTRICAL CHARACTERISTICS:** $T_A = +25$ °C with $4.5V \le (V_{DD} - V_{DR}) \le 12V$ , $4.5V \le (V_{DD} - V_{SS}) \le 12V$

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
VIH	Logic 1, High Input Voltage		2.4	1.6		٧
VIL	Logic 0, Low Input Voltage			1.3	0.8	٧
I <sub>IN</sub>	Input Current	$0V \leq V_{IN} \leq V_{DD}$	-10		10	μА
V <sub>OH</sub>	High Output Voltage		V <sub>DD</sub> -0.025			V
Vol	Low Output Voltage				V <sub>SS</sub> +0.025	V
Ro	Output Resistance	$V_{IN} = 0V$ , $I_{OUT} = 10$ mA, $V_{SS} = 0V$ , $V_{DD} = 12V$		3	5	Ω
		$V_{IN} = 3V$ , $I_{OUT} = 10$ mA, $V_{SS} = 0V$ , $V_{DD} = 12V$		3	5	Ω
lpk	Peak Output Current	V <sub>DD</sub> = 12V		3		Α
tskew1	Output Pulse Skew	Figure 2		3	5	ns
tskew2	Output Pulse Skew	Figure 2		3	5	ns
t <sub>R</sub>	Rise Time	Figure 1 C <sub>L</sub> = 2200 pF, V <sub>DD</sub> = 12V		22	30	ns
tF	Fall Time	Figure 1 $C_L = 2200 \text{ pF}, V_{DD} = 12V$		22	30	ns
t <sub>D1</sub>	Delay Time	Figure 1		18	25	ns
t <sub>D2</sub>	Delay Time	Figure 1		18	25	ns
ls	Quiescent Power Supply Current	$V_{IN} = 3V$ , $V_{DD} = 12V$ , $V_{SS} = 0V$ $V_{IN} = 0V$ , $V_{DD} = 12V$ , $V_{SS} = 0V$		2.9	5 0.3	mA mA

NOTE: Switching times are guaranteed by design.

# TC430

# **ELECTRICAL CHARACTERISTICS:** $4.5V \le (V_{DD} - V_{DR}) \le 12V$ ; $4.5V \le (V_{DD} - V_{SS}) \le 12V$ , $T_A = -55^{\circ}C$ to $+125^{\circ}C$

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
V <sub>IH</sub>	Logic 1, High Input Voltage		2.4			٧
V <sub>IL</sub>	Logic 0, Low Input Voltage				0.8	٧
l <sub>IN</sub>	Input Current	$0V \le V_{IN} \le V_{DD}$	-10		10	μА
V <sub>OH</sub>	High Output Voltage		V <sub>DD</sub> 0.025			٧
V <sub>OL</sub>	Low Output Voltage				V <sub>SS</sub> +0.025	٧
Ro	Output Resistance	$V_{IN} = 0V$ , $I_{OUT} = 10$ mA, $V_{SS} = 0V$ , $V_{DD} = 12V$		4.5	7	Ω
		$V_{IN} = 3V$ , $I_{OUT} = 10$ mA, $V_{SS} = 0V$ , $V_{DD} = 12V$		4.5	7	Ω
l <sub>PK</sub>	Peak Output Current	V <sub>DD</sub> = 12V		3		Α
t <sub>SKEW1</sub>	Output Pulse Skew	Figure 2		5	10	ns
t <sub>SKEW2</sub>	Output Pulse Skew	Figure 2		5	10	ns
t <sub>R</sub>	Rise Time	Figure 1 C <sub>L</sub> = 2200 pF, V <sub>DD</sub> = 12V			40	ns
t <sub>F</sub>	Fall Time	Figure 1 C <sub>L</sub> = 2200 pF, V <sub>DD</sub> = 12V			40	ns
t <sub>D1</sub>	Delay Time	Figure 1		T .	35	ns
t <sub>D2</sub>	Delay Time	Figure 1			35	ns
Is	Quiescent Power Supply Current	$V_{IN} = 3V$ , $V_{DD} = 12V$ , $V_{SS} = 0V$ $V_{IN} = 0V$ , $V_{DD} = 12V$ , $V_{SS} = 0V$		5	8 0.5	mA mA

NOTE: Switching times are guaranteed by design.

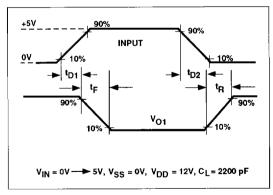


Figure 1 Driver Switching Time

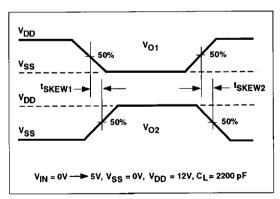
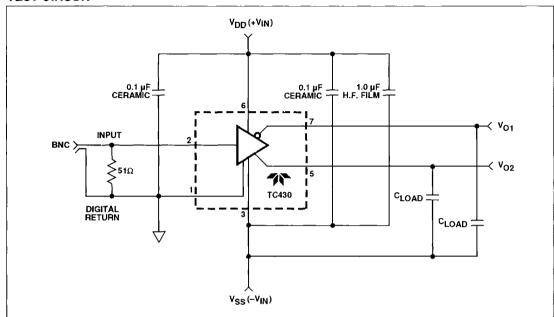


Figure 2 Output Drive Skew

#### **TEST CIRCUIT**



# APPLICATIONS INFORMATION Functional Description

The TC430 is fabricated with a super-fast silicon gate process. The input stage consists of a Schmitt trigger which drives a level shift circuit. The level shift circuit allows the input signal to be referenced to some point other than the output return, pin 3 ( $V_{SS}$ ). This allows the output to swing positive and negative relative to the digital return (pin 1).

The output stage is a low-impedance MOSFET totem pole that can source or sink currents up to 3A peak. This type of output can swing to within millivolts of either rail when driving capacitive loads. Output rise times are on the order of 3 ns, while propagation delays are in the 15 ns region.

## **Application Tips**

Due to its high speed and short transition times, proper layout of the PC board is critical. See Application Note 28 for further information on the effects of layout.

Additional precautions that must be made in addition to those in Application Note 28 are:

 Decoupling between the digital return and V<sub>DD</sub> is critical.

- (2) A minimum 4.5V must be maintained between digital return and V<sub>DD</sub>.
- (3) Decoupling between V<sub>DD</sub> and V<sub>SS</sub> is critical.
- (4) Single-point (star) ground systems should be used.

For decoupling between digital return and  $V_{DD}$  [item (1) above] a 1  $\mu$ F 50V polyester film cap (such as a Wima MKS-2) in parallel with a multilayer ceramic 0.1  $\mu$ F 50 X7R (such as an AVX dip guard) will work well. These capacitors have to be mounted as close as possible to the respective pins on the TC430 to minimize circuit inductance.

Circuits that are improperly decoupled will exhibit oscillations on the output.

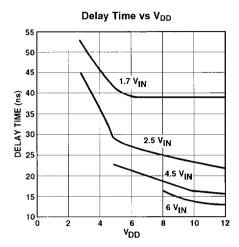
A minimum 4.5V between digital return and V<sub>DD</sub> [item (2) above] is necessary to ensure that the level shifting and hysteresis circuits have enough voltage to function properly. Put another way, the input circuit is referenced to the positive supply, not the negative supply.

Decoupling of the  $V_{SS}$  to  $V_{DD}$  [item (3) above] is important because of the high peak current capability of the output of the TC430. The suggested decoupling is a low ESR polyester film capacitor (such as the 1  $\mu$ F 50V MKS-2) and a ceramic capacitor (such as the AVX 0.1  $\mu$ F 50V dip quard).

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# **TC430**

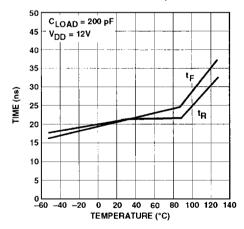


NOTE: Digital return tied to Vss.

## **Delay Time vs Temperature** 25.0 V<sub>DD</sub> = 12V 22.5 t<sub>D1</sub> 20.0 t<sub>D1</sub> 17.5 DELAY TIME (ns) 15.0 12.5 10.0 7.5 5.0 2,5 -60 -40 -20 0 20 40 60 80 100 120 140 TEMPERATURE (°C)

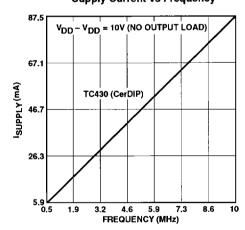
NOTE: Digital return tied to V<sub>SS</sub>.

#### Rise/Fall Times vs Temperature



NOTE: Digital return tied to  $V_{\rm SS}$ .

## **Supply Current vs Frequency**



The parallel combination of the two capacitors forms a low-impedance source of power across a broad frequency range for the output stage. This will ensure that for any load and frequency of operation the output will be as "clean" as is practical.

The use of single-point grounds [item (4) above] is very critical. Due to the high peak currents that the TC430 is capable of generating, any additional trace or wire length can cause L di/dt drops that can effect the output and in the extreme cause the device to fail due to voltage breakdown.

Application Note 28 explains parasitic inductance problems further.

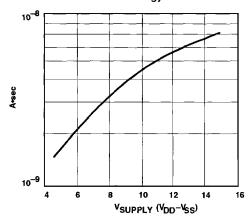
## **Operation From a Single Supply**

If the TC430 is operated from a single supply voltage, the digital return pin must be tied to the  $V_{SS}$  pin. This eliminates the need for the decoupling capacitors from  $V_{DD}$  to digital return.

#### **Load Return Path**

It is very important to return the load currents directly to, and in the shortest possible distance to  $V_{SS}$ , pin 3. Again, this is due to the parasitic inductance of the PC board trace or wire. The test circuit shows how the load capacitors,  $C_{LOAD}$ , are returned to the same point as the decoupling capacitors which is directly on pin 3.

#### **Crossover Energy Loss**



## **Input Signal Considerations**

The amplitude of the input signal has a significant effect on the propagation delay through the IC.

While the device can be driven with a signal as small as 2V, propagation delays will be in the 40 ns region. If the input is increased to 5V, delays will be in the 15 ns region.

The input stage of the TC430 is a MOSFET gate. Thus, it is of high impedance and requires little drive current. This eliminates the need for speed-up capacitors as with older bipolar parts. The use of speed-up capacitors is not recommended, as they can cause voltage-doubling effects that can be detrimental to the life of the device.

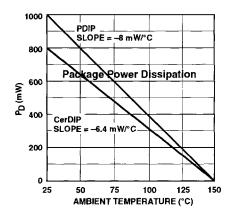
Table I Maximum Operating Frequency

Vs	Max Frequency
12V	4 MHz
10V	9.1 MHz
5 <b>V</b>	20 MHz

Conditions: 1. CerDIP Package (θ<sub>JA</sub> = 150°C/W)

2. T<sub>A</sub> = 25°C

3. No load



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# TC430

## **APPLICATIONS**

