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- High-Performance Static CMOS Technology
   25-ns Instruction Cycle Time (40 MHz)
  - 40-MIPS Performance
  - Low-Power 3.3-V Design
- Based on TMS320C2xx DSP CPU Core
   Code-Compatible With F243/F241/C242
  - Instruction Set and Module Compatible With F240/C240
- Flash (LF) and ROM (LC) Device Options
   LF240xA<sup>†</sup>: LF2407A, LF2406A, LF2402A
  - LC240xA<sup>†</sup>: LC2406A, LC2404A, LC2402A
- On-Chip Memory
  - Up to 32K Words x 16 Bits of Flash EEPROM (4 Sectors) or ROM
  - Programmable "Code-Security" Feature for the On-Chip Flash/ROM
  - Up to 2.5K Words x 16 Bits of Data/Program RAM
    - 544 Words of Dual-Access RAM
    - Up to 2K Words of Single-Access RAM
  - Boot ROM (LF240xA Devices) – SCI/SPI Bootloader
- Two Event-Manager (EV) Modules (EVA and EVB), Each Include:
  - Two 16-Bit General-Purpose Timers
  - Eight 16-Bit Pulse-Width Modulation (PWM) Channels Which Enable:
    - Three-Phase Inverter Control
    - Center- or Edge-Alignment of PWM Channels
    - Emergency PWM Channel Shutdown With External PDPINTx Pin
  - Programmable Deadband (Deadtime)
     Prevents Shoot-Through Faults
  - Three Capture Units For Time-Stamping of External Events
  - Input Qualifier for Select Pins
  - On-Chip Position Encoder Interface Circuitry
  - Synchronized A-to-D Conversion
  - Designed for AC Induction, BLDC, Switched Reluctance, and Stepper Motor Control
  - Applicable for Multiple Motor and/or Converter Control

- External Memory Interface (LF2407A)
   192K Words x 16 Bits of Total Memory: 64K Program, 64K Data, 64K I/O
- Watchdog (WD) Timer Module
- 10-Bit Analog-to-Digital Converter (ADC)
  - 8 or 16 Multiplexed Input Channels
  - 500 ns Minimum Conversion Time Selectable Twin 8-Input Sequencers
  - Triggered by Two Event Managers
- Controller Area Network (CAN) 2.0B Module
- Serial Communications Interface (SCI)
- 16-Bit Serial Peripheral Interface (SPI)
- Phase-Locked-Loop (PLL)-Based Clock Generation
- Up to 40 Individually Programmable, Multiplexed General-Purpose Input/Output (GPIO) Pins
- Five External Interrupts (Power Drive Protection, Reset, and Two Maskable Interrupts)
- Power Management:
  - Three Power-Down Modes
  - Ability to Power Down Each Peripheral Independently
- Real-Time JTAG-Compliant Scan-Based Emulation, IEEE Standard 1149.1<sup>‡</sup> (JTAG)
- Development Tools Include:
  - Texas Instruments (TI) ANSI C Compiler, Assembler/Linker, and Code Composer Studio<sup>™</sup> Debugger
  - Evaluation Modules
  - Scan-Based Self-Emulation (XDS510<sup>™</sup>)
  - Broad Third-Party Digital Motor Control Support
- Package Options
  - 144-Pin LQFP PGE (LF2407A)
  - 100-Pin LQFP PZ (LC2404A, LC2406A, LF2406A)
  - 64-Pin QFP PG (LC2402A and LF2402A)
- Extended Temperature Options (A and S)
  - − A: − 40°C to 85°C
  - S: 40°C to 125°C



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

Code Composer Studio and XDS510 are trademarks of Texas Instruments.

<sup>†</sup>Throughout this data sheet, 240xA is used as a generic name for the LF240xA/LC240xA generation of devices.

<sup>‡</sup>IEEE Standard 1149.1–1990, IEEE Standard Test-Access Port

ADVANCE INFORMATION concerns new products in the sampling or preproduction phase of development. Characteristic data and other specifications are subject to change without notice.



**ADVANCE INFORMATION** 

# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

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# **ADVANCE INFORMATION**

REVISION HISTORY										
REVISION	DATE	PRODUCT STATUS	HIGHLIGHTS							
В	November 2000	Advance Information	The on-chip ROM size for the LC2402A device has been changed to 6K words. The description for the V <sub>CCP</sub> pin has been modified. This informa- tion can be found in Table 2, LF240xA and LC240xA Pin List and Package Options. The conditions for high-impedance state for the strobe signals have been changed. This information can be found in Table 2, LF240xA and LC240xA Pin List and Package Options. All 240xA parts will be rated for a maximum clock speed of 40 MHz. There will be no 30-MHz 240xA parts. The t <sub>w</sub> (CAP), t <sub>w</sub> (PDP), t <sub>w</sub> (INT), and t <sub>w</sub> (PDP-WAKE) parameters have been changed. Ready-on-Read and Ready-on-Write timings for one software wait state and one external wait state have been added. Bits 15 and 8 of the SCSR1 register are now reserved (see Table 18, LF240xA/LC240xA DSP Peripheral Register Description). A new section, <i>Migrating From 240x Devices to 240xA Devices</i> , has been added. The boot ROM description has changed. It can now support the x2 or x4 option for the PLL, depending on the state of the SCITXD pin.							

### 



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### description

The TMS320LF240xA and TMS320LC240xA devices, new members of the TMS320C24x<sup>™</sup> generation of digital signal processor (DSP) controllers, are part of the TMS320C2000<sup>™</sup> platform of fixed-point DSPs. The 240xA devices offer the enhanced TMS320<sup>™</sup> DSP architectural design of the C2xx core CPU for low-cost, low-power, and high-performance processing capabilities. Several advanced peripherals, optimized for digital motor and motion control applications, have been integrated to provide a true single-chip DSP controller. While code-compatible with the existing C24x<sup>™</sup> DSP controller devices, the 240xA offers increased processing performance (40 MIPS) and a higher level of peripheral integration. See the *TMS320x240xA Device Summary* section for device-specific features.

The 240xA generation offers an array of memory sizes and different peripherals tailored to meet the specific price/performance points required by various applications. Flash devices of up to 32K words offer a cost-effective reprogrammable solution for volume production. The 240xA devices offer a password-based "code security" feature which is useful in preventing unauthorized duplication of proprietary code stored in on-chip Flash/ROM. Note that Flash-based devices contain a 256-word boot ROM to facilitate in-circuit programming. The 240xA family also includes ROM devices that are fully pin-to-pin compatible with their Flash counterparts.

All 240xA devices offer at least one event manager module which has been optimized for digital motor control and power conversion applications. Capabilities of this module include center- and/or edge-aligned PWM generation, programmable deadband to prevent shoot-through faults, and synchronized analog-to-digital conversion. Devices with dual event managers enable multiple motor and/or converter control with a single 240xA DSP controller. Select EV pins have been provided with an "input-qualifier" circuitry, which minimizes inadvertent pin-triggering by glitches.

The high-performance, 10-bit analog-to-digital converter (ADC) has a minimum conversion time of 500 ns and offers up to 16 channels of analog input. The autosequencing capability of the ADC allows a maximum of 16 conversions to take place in a single conversion session without any CPU overhead.

A serial communications interface (SCI) is integrated on all devices to provide asynchronous communication to other devices in the system. For systems requiring additional communication interfaces, the 2407A, 2406A, and 2404A offer a 16-bit synchronous serial peripheral interface (SPI). The 2407A and 2406A offer a controller area network (CAN) communications module that meets 2.0B specifications. To maximize device flexibility, functional pins are also configurable as general-purpose inputs/outputs (GPIOs).

To streamline development time, JTAG-compliant scan-based emulation has been integrated into all devices. This provides non-intrusive real-time capabilities required to debug digital control systems. A complete suite of code-generation tools from C compilers to the industry-standard Code Composer Studio<sup>™</sup> debugger supports this family. Numerous third-party developers not only offer device-level development tools, but also system-level design and development support.

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### TMS320x240xA device summary

Note that throughout this data sheet, 240xA is used as a generic name for the LF240xA/LC240xA generation of devices.

FEATURE		LF2407A	LF2406A	LF2402A	LC2406A	LC2404A	LC2402A
C2xx DSP Core	Yes	Yes	Yes	Yes	Yes	Yes	
Instruction Cycle	25 ns	25 ns	25 ns	25 ns	25 ns	25 ns	
MIPS (40 MHz)		40 MIPS	40 MIPS	40 MIPS	40 MIPS	40 MIPS	40 MIPS
	Dual-Access RAM (DARAM)	544	544	544	544	544	544
RAM (16-bit word)	Single-Access RAM (SARAM)	2К	2K	512	2K	1K	_
On-chip Flash (16-bit wor (4 sectors: 4K, 12K, 12K,	,	32K	32K	8K	_	_	_
On-chip ROM (16-bit wor	d)	_	_	—	32K	16K	6K
Code Security for On-Chi	ip Flash/ROM	Yes	Yes	Yes	Yes	Yes	Yes
Boot ROM (16-bit word)		Yes	Yes	Yes	—	_	_
External Memory Interfac	e	Yes	_	—	—	_	_
Event Managers A and B	(EVA and EVB)	EVA, EVB	EVA, EVB	EVA	EVA, EVB	EVA, EVB	EVA
General-Purpo	se (GP) Timers	4	4	2	4	4	2
Compare (CMF	P)/PWM	12/16	12/16	6/8	12/16	12/16	6/8
Capture (CAP)	Capture (CAP)/QEP		6/4	3/2	6/4	6/4	3/2
<ul> <li>Input qualifier circuitry on PDPINTx, CAPn, XINT1/2, and ADCSOC pins</li> </ul>		Yes	Yes	Yes	Yes	Yes	Yes
Status of PDPI in COMCONx	NTx pin reflected register	Yes	Yes	Yes	Yes	Yes	Yes
Watchdog Timer		Yes	Yes	Yes	Yes	Yes	Yes
10-Bit ADC		Yes	Yes	Yes	Yes	Yes	Yes
Channels		16	16	8	16	16	8
Conversion Tin	ne (minimum)	500 ns	500 ns	500 ns	500 ns	500 ns	500 ns
SPI		Yes	Yes	—	Yes	Yes	_
SCI		Yes	Yes	Yes	Yes	Yes	Yes
CAN		Yes	Yes	_	Yes	_	_
Digital I/O Pins (Shared)		41	41	21	41	41	21
External Interrupts		5	5	3	5	5	3
Supply Voltage		3.3 V	3.3 V	3.3 V	3.3 V	3.3 V	3.3 V
Packaging		144-pin PGE	100-pin PZ	64-pin PG	100-pin PZ	100-pin PZ	64-pin PG

### Table 1. Hardware Features of 240xA Devices

Denotes features that are different/new compared to 240x devices.



# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

### functional block diagram of the 2407A DSP controller

-				_
				PLLF
				PLLVCCA
XINT1/IOPA2		DARAM (B0)	PLL Clock	PLLF2
RS		256 Words		XTAL1/CLKIN
CLKOUT/IOPE0	_			XTAL2
TMS2				ADCIN00-ADCIN07
BIO/IOPC1	C2xx DSP	DARAM (B1)		ADCIN08-ADCIN15
MP/MC	Core	256 Words	10-Bit ADC	VCCA
BOOT_EN/XF			(With Twin	VSSA
			Autosequencer)	VREFHI
		DARAM (B2)		VREFLO
		32 Words		XINT2/ADCSOC/IOPD0
				SCITXD/IOPA0
\/ (2.2.\/)			SCI	SCIRXD/IOPA1
V <sub>DD</sub> (3.3 V)	SARA	M (2K Words)		SPISIMO/IOPC2
VSS	JAIVA			SPISOMI/IOPC3
			. SPI	SPICLK/IOPC4
				SPISTE/IOPC5
TP1				CANTX/IOPC6
TP2		lash/ROM 2K Words:	CAN	CANRX/IOPC7
V <sub>CCP</sub> (5V)		12K/12K/4K)		
			WD	
				Port A(0-7) IOPA[0:7]
A0-A15			il	Port B(0-7) IOPB[0:7]
D0D15			Digital I/O	Port C(0-7) IOPC[0:7]
PS, DS, IS			(Shared With	Port D(0) IOPD[0]
R/W			Other Pins)	Port E(0-7) IOPE[0:7]
RD				Port F(0-6) IOPF[0:6]
READY	- Exte	ernal Memory		TRST
STRB		Interface		TDO
WE				TDI
ENA_144			JTAG Port	TMS
				ТСК
VIS_OE				EMU0
W/R / IOPC0				EMU1
PDPINTA				PDPINTB
CAP1/QEP1/IOPA3				CAP4/QEP3/IOPE7
CAP2/QEP2/IOPA4				CAP5/QEP4/IOPF0
CAP3/IOPA5				CAP6/IOPF1
PWM1/IOPA6	_			PWM7/IOPE1
PWM2/IOPA7	Ever	nt Manager A	Event Manager B	PWM8/IOPE2
PWM3/IOPB0	• 3 ~	Capture Input	• 3 × Capture Input	PWM9/IOPE3
PWM4/IOPB1		Compare/PWM	• 6 × Compare/PWM	PWM10/IOPE4
PWM5/IOPB2	Out	put	Output	PWM11/IOPE5
PWM6/IOPB3	• 2 ×		• 2×GP	PWM12/IOPE6
T1PWM/T1CMP/IOPB4	Tim	ers/PWM	Timers/PWM	T3PWM/T3CMP/IOPF2
T2PWM/T2CMP/IOPB5				T4PWM/T4CMP/IOPF3
				TDIRB/IOPF4
TDIRA/IOPB6 TCLKINA/IOPB7	-			TCLKINB/IOPF5



Indicates optional modules.

The memory size and peripheral selection of these modules change for different 240xA devices. See Table 1 for device-specific details.



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TCLKINB/10PF5 BOOT\_EN/XF<sup>1</sup> XTAL1/CLKIN BIO/IOPC ADCIN00 ENA\_144 ADCIN09 PDPINTB VREFLO ADCIN08 ADCIN01 ADCIN10 VREFHI MP/MC XTAL2 READY Vsso VCCA Vbbo Vsso VSSA TMS TDO OPF V SS Ī 10K D6 D5 RS â 4 D2 D3 123 119 118 116 113 111 110 43 142 140 139 38 130 129 128 126 125 124 122 120 117 115 114 112 60 44 37 36 35 32 127 121 141 131  $\cap$ TRST 108 ADCIN11 TDIRB/IOPF4 107 ADCIN02 106 ADCIN12 Vsso 105 V<sub>DDO</sub> ADCIN03 D7 5 104 ADCIN13 103 T4PWM/T4CMP/IOPF3 ADCIN04 **PDPINTA** 102 ADCIN05 T3PWM/T3CMP/IOPF2 101 ADCIN14 D8 9 100 ADCIN06 10 99 ADCIN07 PLLF2 11 98 ADCIN15 PLLF 12 97 VIS\_OE PLLV<sub>CCA</sub> 13 96 STRB D9 TDIRA/IOPB6 14 95 ] v<sub>ddo</sub> D10 15 94 ] v<sub>sso</sub> 16 RD T1PWM/T1CMP/IOPB4 93 D11 17 92 ] R/W TMS320LF2407A PGE ] EMU1/OFF T2PWM/T2CMP/IOPB5 18 91 ] ЕМОО 19 W/R/IOPC0 90 20 89 ] WE D12 XINT2/ADCSOC/IOPDO 21 CAP4/QEP3/IOPE7 88 D13 22 87 DS 23 86 ] V<sub>DD</sub> XINT1/IOPA2 24 D14 85 ] v<sub>ss</sub> SCITXD/IOPA0 25 84 1 PS SCIRXD/IOPA1 26 83 CAP1/QEP1/IOPA3 D15 27 82 l is 28 81 CAP5/QEP4/IOPF0 VSS 29 80  $V_{\text{DD}}$ SPISIMO/IOPC2 30 79 CAP2/QEP2/IOPA4 31 78 A15 A1 SPISOMI/IOPC3 32 77 ] v<sub>ddo</sub> SPISTE/IOPC5 76 ] V<sub>SSO</sub> ] CAP3/*IOPA5* 33 A14 34 75 SPICLK/IOPC4 35 74 ] A2 TMS2 36 73 CLKOUT/IOPE0 38 39 40 42 43 4 45 

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 37 4 PWM6/*IOPB3* VSSO A12 PWM5/10PB2 PWM7/ IOPE1 VSSO VDDO VCCP VDDO A13 A10 VSS VDD A9 A8 A5 A4 A3

PGE PACKAGE<sup>†</sup> (TOP VIEW)

† Bold, italicized pin names indicate pin function after reset.

TCLKINA/IOPB7 PWM12/10PE6 A11

PWM11/10PE5 PWM4/10PB1

<sup>‡</sup>BOOT\_EN is available only on Flash devices.



PWM3/10PB0

-WM10/10PE4

PWM2/10PA7 PWM1/10PA6

A7

PWM9/10PE3

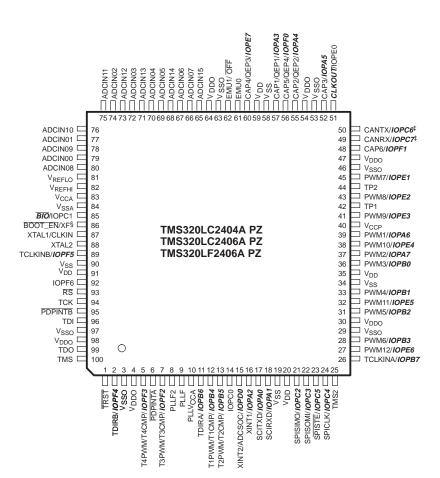
A6 PWM8/10PE2 TP2

Ē

CAP6/10PF1 CANRX/10PC7 CANTX/10PC6

### TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

PZ PACKAGE<sup>†</sup> (TOP VIEW)



† Bold, italicized pin names indicate pin function after reset.

‡CANTX and CANRX are not available on LC2404A devices.

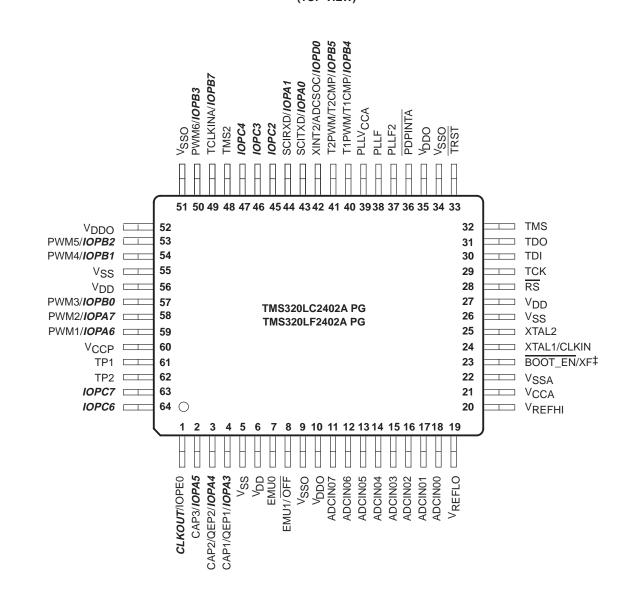
§ BOOT\_EN is available only on Flash devices.





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PG PACKAGE<sup>†</sup> (TOP VIEW)



**Bold, italicized pin names** indicate pin function after reset. **BOOT\_EN** is available only on Flash devices.



### pin functions

The TMS320LF2407A device is the superset of all the 240xA devices. All signals are available on the 2407A device. Table 2 lists the signals available in the 240xA generation of devices.

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
		-	EVENT	MANAGER	A (EVA)
CAP1/QEP1/ <i>IOPA3</i>	83	57	57	4	Capture input #1/quadrature encoder pulse input #1 (EVA) or GPIO $(\uparrow)$
CAP2/QEP2/ <i>IOPA4</i>	79	55	55	3	Capture input #2/quadrature encoder pulse input #2 (EVA) or GPIO (1)
CAP3/ <i>IOPA5</i>	75	52	52	2	Capture input #3 (EVA) or GPIO (1)
PWM1/ <i>IOPA6</i>	56	39	39	59	Compare/PWM output pin #1 (EVA) or GPIO (1)
PWM2/ <i>IOPA7</i>	54	37	37	58	Compare/PWM output pin #2 (EVA) or GPIO (1)
PWM3/ <i>IOPB0</i>	52	36	36	57	Compare/PWM output pin #3 (EVA) or GPIO (1)
PWM4/ <i>IOPB1</i>	47	33	33	54	Compare/PWM output pin #4 (EVA) or GPIO (1)
PWM5/ <i>IOPB2</i>	44	31	31	53	Compare/PWM output pin #5 (EVA) or GPIO (1)
PWM6/ <i>IOPB3</i>	40	28	28	50	Compare/PWM output pin #6 (EVA) or GPIO (1)
T1PWM/T1CMP/ <i>IOPB4</i>	16	12	12	40	Timer 1 compare output (EVA) or GPIO $(\uparrow)$
T2PWM/T2CMP/IOPB5	18	13	13	41	Timer 2 compare output (EVA) or GPIO (↑)
TDIRA/ <i>IOPB6</i>	14	11	11		Counting direction for general-purpose (GP) timer (EVA) or GPIO. If TDIRA = 1, upward counting is selected. If TDIRA = 0, downward counting is selected. $(\uparrow)$
TCLKINA/ <i>IOPB7</i>	37	26	26	49	External clock input for GP timer (EVA) or GPIO. Note that the timer can also use the internal device clock. $(\uparrow)$
		-	EVENT	MANAGER I	B (EVB)
CAP4/QEP3/ <i>IOPE7</i>	88	60	60		Capture input #4/quadrature encoder pulse input #3 (EVB) or GPIO $(\uparrow)$
CAP5/QEP4/ <i>IOPF0</i>	81	56	56		Capture input #5/quadrature encoder pulse input #4 (EVB) or GPIO (1)
CAP6/ <i>IOPF1</i>	69	48	48		Capture input #6 (EVB) or GPIO (1)
PWM7/ <i>IOPE1</i>	65	45	45		Compare/PWM output pin #7 (EVB) or GPIO (1)
PWM8/ <i>IOPE2</i>	62	43	43		Compare/PWM output pin #8 (EVB) or GPIO (1)
PWM9/ <i>IOPE3</i>	59	41	41		Compare/PWM output pin #9 (EVB) or GPIO (1)
PWM10/ <i>IOPE4</i>	55	38	38		Compare/PWM output pin #10 (EVB) or GPIO (1)
PWM11/ <i>IOPE5</i>	46	32	32		Compare/PWM output pin #11 (EVB) or GPIO (1)
PWM12/ <i>IOPE6</i>	38	27	27		Compare/PWM output pin #12 (EVB) or GPIO (1)

Table 2. LF240xA and LC240xA	Pin List and	Package	Options <sup>†‡</sup>
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† Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

 $\P$  Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION				
EVENT MANAGER B (EVB) (CONTINUED)									
T3PWM/T3CMP/ <i>IOPF2</i>	8	7	7		Timer 3 compare output (EVB) or GPIO ( $\uparrow$ )				
T4PWM/T4CMP/ <i>IOPF3</i>	6	5	5		Timer 4 compare output (EVB) or GPIO ( $\uparrow$ )				
TDIRB/ <i>IOPF4</i>	2	2	2		Counting direction for general-purpose (GP) timer (EVB) or GPIO. If TDIRB = 1, upward counting is selected. If TDIRB = 0, downward counting is selected. ( $\uparrow$ )				
TCLKINB/ <i>IOPF5</i>	126	89	89		External clock input for GP timer (EVB) or GPIO. Note that the timer can also use the internal device clock. $(\uparrow)$				
	ANA	ALOG-TO-D	IGITAL CO	NVERTER (A	NDC)				
ADCIN00	112	79	79	18	Analog input #0 to the ADC				
ADCIN01	110	77	77	17	Analog input #1 to the ADC				
ADCIN02	107	74	74	16	Analog input #2 to the ADC				
ADCIN03	105	72	72	15	Analog input #3 to the ADC				
ADCIN04	103	70	70	14	Analog input #4 to the ADC				
ADCIN05	102	69	69	13	Analog input #5 to the ADC				
ADCIN06	100	67	67	12	Analog input #6 to the ADC				
ADCIN07	99	66	66	11	Analog input #7 to the ADC				
ADCIN08	113	80	80		Analog input #8 to the ADC				
ADCIN09	111	78	78		Analog input #9 to the ADC				
ADCIN10	109	76	76		Analog input #10 to the ADC				
ADCIN11	108	75	75		Analog input #11 to the ADC				
ADCIN12	106	73	73		Analog input #12 to the ADC				
ADCIN13	104	71	71		Analog input #13 to the ADC				
ADCIN14	101	68	68		Analog input #14 to the ADC				
ADCIN15	98	65	65		Analog input #15 to the ADC				
VREFHI	115	82	82	20	ADC analog high-voltage reference input				
VREFLO	114	81	81	19	ADC analog low-voltage reference input				
VCCA	116	83	83	21	Analog supply voltage for ADC (3.3 V)§				
VSSA	117	84	84	22	Analog ground reference for ADC				

+ Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (VDD, VDDO, VSS, or VSSO) should be left unconnected. All power supply pins must be connected appropriately for proper device operation. (Typical active pullup/pulldown value is  $\pm 16~\mu A.)$ 

LEGEND: 1 – Internal pullup  $\downarrow$  – Internal pulldown



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME		LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
CONTROLLER AREA	NETWORK	(CAN), SERIA		NICATIONS	INTERFACE	(SCI), SERIAL PERIPHERAL INTERFACE (SPI)
CANRX/ <i>IOPC7</i>	CANRX	70	49	-	-	CAN reactive data or CDIO ( <sup>A</sup> )
CANRX/IOPC7	IOPC7	70	49	49	63	CAN receive data or GPIO (1)
CANTY/IODCC	CANTX	72	50	-	-	CAN transmit data at CDIO $(^{\uparrow})$
CANTX/ <i>IOPC6</i>	IOPC6	72	50	50	64	CAN transmit data or GPIO (1)
SCITXD/IOPA0		25	17	17	43	SCI asynchronous serial port transmit data or GPIO $(\uparrow)$
SCIRXD/IOPA1		26	18	18	44	SCI asynchronous serial port receive data or or GPIO $(\uparrow)$
001011///0004	SPICLK	35	24	24	-	
SPICLK/IOPC4	IOPC4	35	24	24	47	SPI clock or GPIO (↑)
	SPISIMO	30	21	21	-	
SPISIMO/IOPC2	IOPC2	30	21	21	45	SPI slave in, master out or GPIO (1)
	SPISOMI	32	22	22	-	
SPISOMI/IOPC3	IOPC3	32	22	22	46	SPI slave out, master in or GPIO (1)
·	SPISTE	33	23	23	-	
SPISTE/IOPC5		33	23	23	-	SPI slave transmit-enable (optional) or GPIO (1)
		•	EXTERNAL		PTS, CLOCK	
RS		133	93	93	28	Device reset. $\overline{RS}$ causes the 240xA to terminate execution and sets PC = 0. When $\overline{RS}$ is brought to a high level, execution begins at location zero of program memory. $\overline{RS}$ affects (or sets to zero) various registers and status bits. When the watchdog timer overflows, it initiates a system reset pulse that is reflected on the $\overline{RS}$ pin. ( $\uparrow$ )
PDPINTA		7	6	6	36	Power drive protection interrupt input. This interrupt, when activated, puts the PWM output pins (EVA) in the high-impedance state should motor drive/power converter abnormalities, such as overvoltage or overcurrent, etc., arise. PDPINTA is a falling-edge-sensitive interrupt. ( $\uparrow$ )
XINT1/ <b>IOPA2</b>		23	16	16		External user interrupt 1 or GPIO. Both XINT1 and XINT2 are edge-sensitive. The edge polarity is programmable. $(\uparrow)$
XINT2/ADCSOC/ <i>IOPD0</i>		21	15	15	42	External user interrupt 2 and ADC start of conversion or GPIO. External "start-of-conversion" input for ADC/GPIO. Both XINT1 and XINT2 are edge-sensitive. The edge polarity is programmable. (1)

<sup>†</sup> Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/ $\overline{\text{OFF}}$  is low,  $\overline{\text{TRST}}$  is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME		LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION			
EXTERNAL INTERRUPTS, CLOCK (CONTINUED)									
<i>CLKOUT</i> /IOPE(	0	73	51	51	1	Clock output or GPIO. This pin outputs either the CPU clock (CLKOUT) or the watchdog clock (WDCLK). The selection is made by the CLKSRC bit (bit 14) of the system control and status register (SCSR). This pin can be used as a GPIO if not used as a clock output pin. ( $\uparrow$ )			
PDPINTB		137	95	95		Power drive protection interrupt input. This interrupt, when activated, puts the PWM output pins (EVB) in the high-impedance state should motor drive/power converter abnormalities, such as overvoltage or overcurrent, etc., arise. PDPINTB is a falling-edge-sensitive interrupt. (1)			
		OSCI	LLATOR, P	LL, FLASH	, BOOT, AND	MISCELLANEOUS			
XTAL1/CLKIN		123	87	87	24	PLL oscillator input pin. Crystal input to PLL/clock source input to PLL. XTAL1/CLKIN is tied to one side of a reference crystal.			
XTAL2		124	88	88	25	Crystal output. PLL oscillator output pin. XTAL2 is tied to one side of a reference crystal. This pin goes in the high-impedance state when EMU1/OFF is active low.			
PLLVCCA		12	10	10	39	PLL supply (3.3 V)			
IOPF6		131	92	92		General-purpose I/O (↑)			
BOOT_EN /	BOOT_EN	121	86	-	23	Boot ROM enable, GPO, XF. This pin will be sampled as input (BOOT_EN) to update SCSR2.3 (BOOT_EN bit) during reset and then driven as an output signal for XF. After			
XF	XF	121	86	86	23	reset, XF is driven high. ROM devices do not have boot ROM, hence, no BOOT_EN modes. The $\overline{\text{BOOT}_\text{EN}}$ pin must be driven with a passive circuit only. (1)			
PLLF		11	9	9	38	Filter input 1			
PLLF2		10	8	8	37	Filter input 2			
V <sub>CCP</sub> (5V)		58	40	40	60	Flash programming voltage pin. This pin must be connected to a 5-V supply for Flash programming. The Flash cannot be programmed if this pin is connected to GND. When not programming the Flash (i.e., during normal device operation), this pin can either be left connected to the 5-V supply or it can be tied to GND. This pin must not be left floating at any time. Do not use any current-limiting resistor in series with the 5-V supply on this pin. This pin is a "no connect" (NC) on ROM parts and can be left open.			
TP1		60	42	42	61	Test pin 1. Do not connect.			
TP2		63	44	44	62	Test pin 2. Do not connect.			

*Bold, italicized pin names* indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
	OSCILL	ATOR, PLL	, FLASH, BO	DOT, AND MI	SCELLANEOUS (CONTINUED)
<b>BIO</b> /IOPC1	119	85	85		Branch control input. $\overline{\text{BIO}}$ is polled by the BCND pma, BIO instruction. If $\overline{\text{BIO}}$ is low, a branch is executed. If $\overline{\text{BIO}}$ is not used, it should be pulled high. This pin is configured as a branch control input by all device resets. It can be used as a GPIO, if not used as a branch control input. (1)
			EMUL	ATION AND	TEST
EMU0	90	61	61	7	Emulator I/O #0 with internal pullup. When TRST is driven high, this pin is used as an interrupt to or from the emulator system and is defined as input/output through the JTAG scan. ( $\uparrow$ )
EMU1/OFF	91	62	62	8	Emulator pin 1. Emulator pin 1 disables all outputs. When TRST is driven high, EMU1/OFF is used as an interrupt to or from the emulator system and is defined as an input/output through the JTAG scan. When TRST is driven low, this pin is configured as OFF. EMU1/OFF, when active low, puts all output drivers in the high-impedance state. Note that OFF is used exclusively for testing and emulation purposes (not for multiprocessing applications). Therefore, for the OFF condition, the following $\frac{apply:}{TRST = 0}$ EMU0 = 1 EMU1/OFF = 0 (1)
ТСК	135	94	94	29	JTAG test clock with internal pullup $(\uparrow)$
TDI	139	96	96	30	JTAG test data input (TDI) with internal pullup. TDI is clocked into the selected register (instruction or data) on a rising edge of TCK. $(\uparrow)$
TDO	142	99	99	31	JTAG scan out, test data output (TDO). The contents of the selected register (instruction or data) is shifted out of TDO on the falling edge of TCK. $(\downarrow)$
TMS	144	100	100	32	JTAG test-mode select (TMS) with internal pullup. This serial control input is clocked into the TAP controller on the rising edge of TCK. $(\uparrow)$
TMS2	36	25	25	48	JTAG test-mode select 2 (TMS2) with internal pullup. This serial control input is clocked into the TAP controller on the rising edge of TCK. Used for test and emulation only. This pin can be left unconnected in user applications. $(\uparrow)$
TRST	1	1	1	33	JTAG test reset with internal pulldown. TRST, when driven high, gives the scan system control of the operations of the device. If this signal is not connected or driven low, the device operates in its functional mode, and the test reset signals are ignored. $(\downarrow)$

<sup>†</sup> Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME		LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION			
ADDRESS, DATA, AND MEMORY CONTROL SIGNALS									
DS		87				Data space strobe. IS, DS, and PS are always high unless low-level asserted for access to the relevant external memory space or I/O. They are placed in the high-impedance state.¶			
ĪS		82				I/O space strobe. $\overline{IS}$ , $\overline{DS}$ , and $\overline{PS}$ are always high unless low-level asserted for access to the relevant external memory space or I/O. They are placed in the high-impedance state.¶			
PS		84				Program space strobe. IS, DS, and PS are always high unless low-level asserted for access to the relevant external memory space or I/O. They are placed in the high-impedance state.¶			
R/W		92				Read/write qualifier signal. $R/\overline{W}$ indicates transfer direction during communication to an external device. It is normally in read mode (high), unless low level is asserted for performing a write operation. $R/\overline{W}$ is placed in the high-impedance state.			
	W/R	19				Write/Read qualifier or GPIO. This is an inverted $R/\overline{W}$ signal useful for zero-wait-state memory interface. It is normally low, unless a memory write			
	IOPC0	19	14	14		operation is performed. See Table 12, Port C section, for reset note regarding LF2406A and LF2402A. (1)			
RD		93				Read-enable strobe. Read-select indicates an active, external read cycle. RD is active on all external program, data, and I/O reads. RD is placed in the high-impedance state.¶			
WE		89				Write-enable strobe. The falling edge of $\overline{\text{WE}}$ indicates that the device is driving the external data bus (D15–D0). $\overline{\text{WE}}$ is active on all external program, data, and I/O writes. $\overline{\text{WE}}$ is placed in the high-impedance state.¶			
STRB		96				External memory access strobe. STRB is always high unless asserted low to indicate an external bus cycle. STRB is active for all off-chip accesses. STRB is placed in the high-impedance state.¶			

† Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that VCCA be isolated from the digital supply voltage (and VSSA from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/ $\overline{\text{OFF}}$  is low,  $\overline{\text{TRST}}$  is low, and EMU0 is high

<sup>#</sup> No power supply pin (VDD, VDDO, VSS, or VSSO) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.  $\downarrow$  – Internal pulldown

LEGEND: 1 – Internal pullup

(Typical active pullup/pulldown value is  $\pm 16 \ \mu$ A.)



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
	ADDR	ESS, DATA	, AND MEM		ROL SIGNALS (CONTINUED)
READY	120				READY is pulled low to add wait states for external accesses. READY indicates that an external device is prepared for a bus transaction to be completed. If the device is not ready, it pulls the READY pin low. The processor waits one cycle and checks READY again. Note that the processor performs READY-detection if at least one software wait state is programmed. To meet the external READY timings, the wait-state generator control register (WSGR) should be programmed for at least one wait state. ( $\uparrow$ )
MP/MC	118				Microprocessor/Microcomputer mode select. If this pin is low during reset, the device is put in microcomputer mode and program execution begins at 0000h of internal program memory (Flash EEPROM). A high value during reset puts the device in microprocessor mode and program execution begins at 0000h of external program memory. This line sets the MP/MC bit (bit 2 in the SCSR2 register). $(\downarrow)$
ENA_144	122				Active high to enable external interface signals. If pulled low, the 2407A behaves like the 2406A/2404A—i.e., it has no external memory and generates an illegal address if $\overline{\text{DS}}$ is asserted. This pin has an internal pulldown. ( $\downarrow$ )
VIS_OE	97				Visibility output enable (active when data bus is output). This pin is active (low) whenever the external data bus is driving as an output during visibility mode. Can be used by external decode logic to prevent data bus contention while running in visibility mode.
A0	80				Bit 0 of the 16-bit address bus
A1	78				Bit 1 of the 16-bit address bus
A2	74				Bit 2 of the 16-bit address bus
A3	71				Bit 3 of the 16-bit address bus
A4	68				Bit 4 of the 16-bit address bus
A5	64				Bit 5 of the 16-bit address bus
A6	61				Bit 6 of the 16-bit address bus
A7	57				Bit 7 of the 16-bit address bus
A8	53				Bit 8 of the 16-bit address bus
A9	51				Bit 9 of the 16-bit address bus
A10	48				Bit 10 of the 16-bit address bus

<sup>†</sup> Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

¶ Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### pin functions (continued)

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
	ADDR	ESS, DATA	, AND MEM	ORY CONTR	ROL SIGNALS (CONTINUED)
A11	45				Bit 11 of the 16-bit address bus
A12	43				Bit 12 of the 16-bit address bus
A13	39				Bit 13 of the 16-bit address bus
A14	34				Bit 14 of the 16-bit address bus
A15	31				Bit 15 of the 16-bit address bus
D0	127				Bit 0 of 16-bit data bus (1)
D1	130				Bit 1 of 16-bit data bus (1)
D2	132				Bit 2 of 16-bit data bus (1)
D3	134				Bit 3 of 16-bit data bus (1)
D4	136				Bit 4 of 16-bit data bus (1)
D5	138				Bit 5 of 16-bit data bus (1)
D6	143				Bit 6 of 16-bit data bus (1)
D7	5				Bit 7 of 16-bit data bus (1)
D8	9				Bit 8 of 16-bit data bus (1)
D9	13				Bit 9 of 16-bit data bus (1)
D10	15				Bit 10 of 16-bit data bus (↑)
D11	17				Bit 11 of 16-bit data bus (↑)
D12	20				Bit 12 of 16-bit data bus (↑)
D13	22				Bit 13 of 16-bit data bus (↑)
D14	24				Bit 14 of 16-bit data bus (1)
D15	27				Bit 15 of 16-bit data bus (↑)

<sup>†</sup> Bold, italicized pin names indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

 $\P$  Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

<sup>#</sup> No power supply pin (V<sub>DD</sub>, V<sub>DDO</sub>, V<sub>SS</sub>, or V<sub>SSO</sub>) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.





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### pin functions (continued)

PIN NAME	LF2407A (144-PGE)	2406A (100-PZ)	LC2404A (100-PZ)	2402A (64-PG)	DESCRIPTION
	•	1	PC	WER SUPP	LY
	29	20	20	6	
N#	50	35	35	27	
V <sub>DD</sub> #	86	59	59	56	Core supply +3.3 V. Digital logic supply voltage.
	129	91	91		
	4	4	4	10	
	42	30	30	35	
v #	67	47	47	52	I/O buffer supply +3.3 V. Digital logic and buffer supply
V <sub>DDO</sub> #	77	54	54		voltage.
	95	64	64		
	141	98	98		
	28	19	19	5	
N#	49	34	34	26	
VSS#	85	58	58	55	Core ground. Digital logic ground reference.
	128	90	90		
	3	3	3	9	
	41	29	29	34	
	66	46	46	51	]
VSSO <sup>#</sup>	76	53	53		I/O buffer ground. Digital logic and buffer ground reference.
	94	63	63		
	125	97	97		
	140				]

### Table 2. LF240xA and LC240xA Pin List and Package Options<sup>†‡</sup> (Continued)

*† Bold, italicized pin names* indicate pin function after reset.

<sup>‡</sup>GPIO – General-purpose input/output pin. All GPIOs come up as input after reset.

§ It is highly recommended that V<sub>CCA</sub> be isolated from the digital supply voltage (and V<sub>SSA</sub> from digital ground) to maintain the specified accuracy and improve the noise immunity of the ADC.

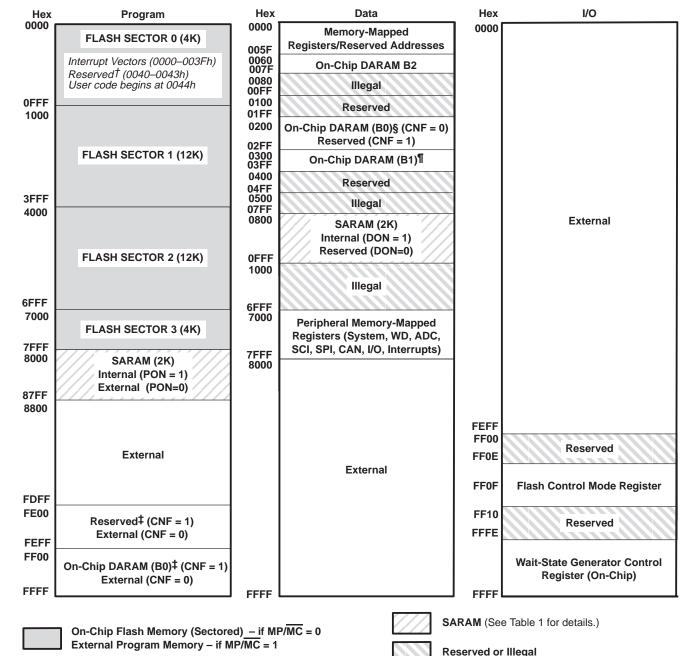
¶ Only when all of the following conditions are met: EMU1/OFF is low, TRST is low, and EMU0 is high

# No power supply pin (VDD, VDDO, VSS, or VSSO) should be left unconnected. All power supply pins must be connected appropriately for proper device operation.



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### memory maps – LF2407A



NOTE A: Boot ROM: If the boot ROM is enabled, then addresses 0000-00FF in the program space will be occupied by boot ROM.

<sup>†</sup> Addresses 0040h–0043h in on-chip program memory are reserved for code security passwords.

When CNF = 1, addresses FE00h-FEFFh and FF00h-FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h-FEFFh are referred to as reserved when CNF = 1.

§ When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example, a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h–01FFh are referred to as reserved.

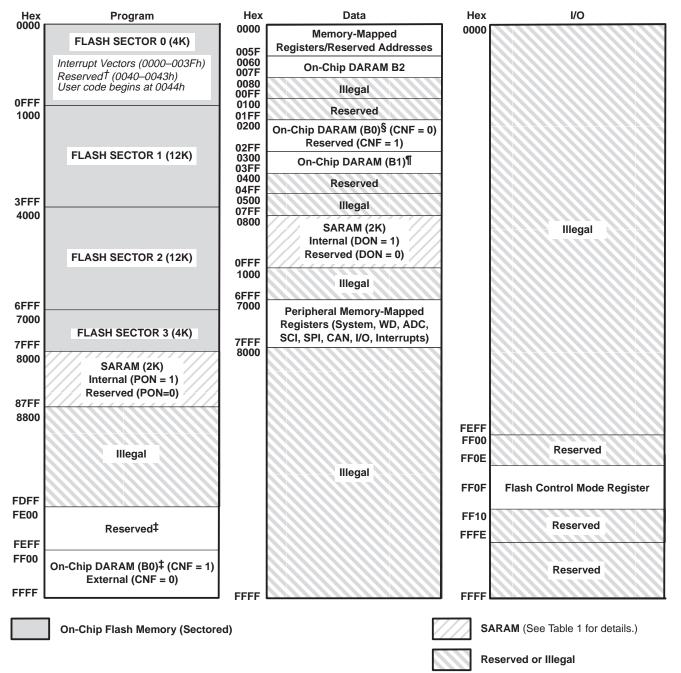
Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

### Figure 1. TMS320LF2407A Memory Map



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### memory maps (continued) - LF2406A



NOTE A: Boot ROM: If the boot ROM is enabled, then addresses 0000–00FF in the program space will be occupied by boot ROM. † Addresses 0040h–0043h in program memory are reserved for code security passwords.

<sup>‡</sup> When CNF = 1, addresses FE00h–FEFFh and FF00h–FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h–FEFFh are referred to as reserved.
§ When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example,

a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h-01FFh are referred to as reserved.

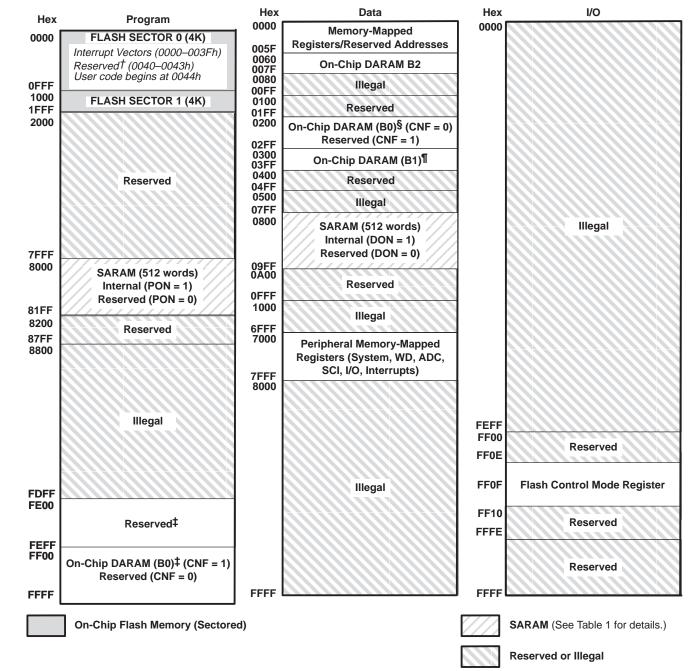
Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

Figure 2. TMS320LF2406A Memory Map



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### memory maps (continued) – LF2402A



NOTE A: Boot ROM: If the boot ROM is enabled, then addresses 0000–00FF in the program space will be occupied by boot ROM. † Addresses 0040h–0043h in program memory are reserved for code security passwords.

<sup>‡</sup> When CNF = 1, addresses FE00h–FEFFh and FF00h–FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h–FEFFh are referred to as reserved. § When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example,

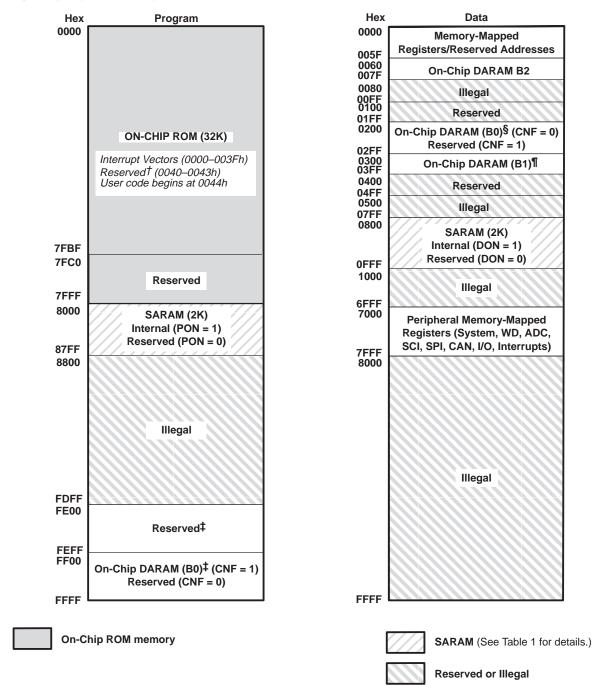
a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h–01FFh are referred to as reserved. Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

### Figure 3. TMS320LF2402A Memory Map



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### memory maps (continued) - LC2406A

<sup>†</sup>Addresses 0040h–0043h in program memory are reserved for code security passwords.

When CNF = 1, addresses FE00h–FEFFh and FF00h–FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h–FEFFh are referred to as reserved.
 When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example,

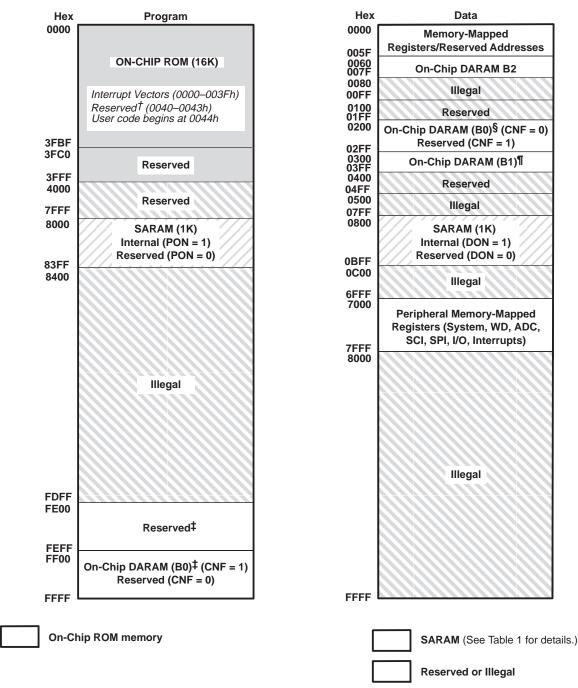
a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h–01FFh are referred to as reserved. Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

### Figure 4. TMS320LC2406A Memory Map



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### memory maps (continued) - LC2404A



<sup>†</sup> Addresses 0040h–0043h in program memory are reserved for code security passwords.

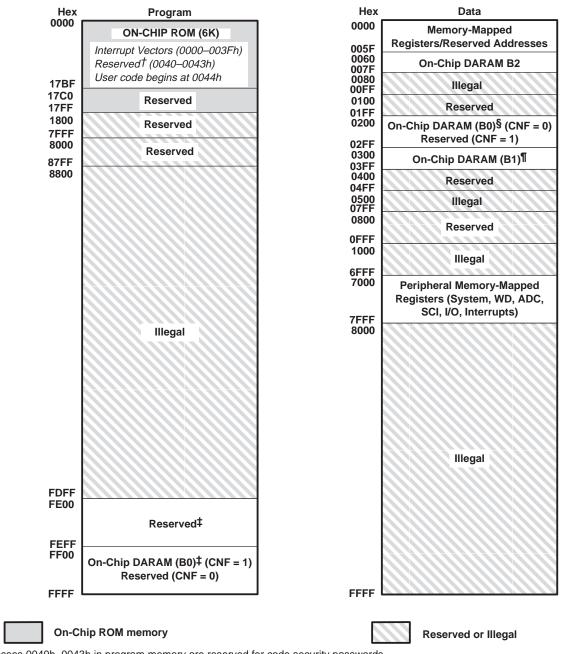
When CNF = 1, addresses FE00h–FEFFh and FF00h–FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h–FEFFh are referred to as reserved.
 When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example,

a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h–01FFh are referred to as reserved. Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

### Figure 5. TMS320LC2404A Memory Map



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### memory maps (continued) - LC2402A

<sup>†</sup> Addresses 0040h–0043h in program memory are reserved for code security passwords.

When CNF = 1, addresses FE00h–FEFFh and FF00h–FFFFh are mapped to the same physical block (B0) in program-memory space. For example, a write to FE00h has the same effect as a write to FF00h. For simplicity, addresses FE00h–FEFFh are referred to as reserved.
 When CNF = 0, addresses 0100h–01FFh and 0200h–02FFh are mapped to the same physical block (B0) in data-memory space. For example,

a write to 0100h has the same effect as a write to 0200h. For simplicity, addresses 0100h–01FFh are referred to as reserved. Addresses 0300h–03FFh and 0400h–04FFh are mapped to the same physical block (B1) in data-memory space. For example, a write to 0400h

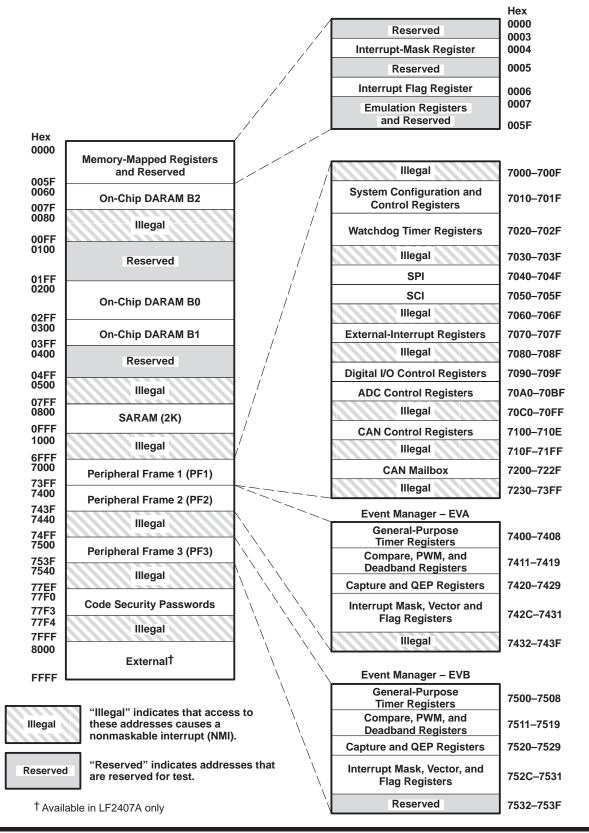
has the same effect as a write to 0300h. For simplicity, addresses 0400h–04FFh are referred to as reserved.

Figure 6. TMS320LC2402A Memory Map



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### peripheral memory map of the 2407A/2406A





### device reset and interrupts

The TMS320x240xA software-programmable interrupt structure supports flexible on-chip and external interrupt configurations to meet real-time interrupt-driven application requirements. The LF240xA recognizes three types of interrupt sources.

• **Reset** (hardware- or software-initiated) is unarbitrated by the CPU and takes immediate priority over any other executing functions. All maskable interrupts are disabled until the reset service routine enables them.

The LF240xA devices have two sources of reset: an external reset pin and a watchdog timer time-out (reset).

- Hardware-generated interrupts are requested by external pins or by on-chip peripherals. There are two types:
  - External interrupts are generated by one of four external pins corresponding to the interrupts XINT1, XINT2, PDPINTA, and PDPINTB. These four can be masked both by dedicated enable bits and by the CPU's interrupt mask register (IMR), which can mask each maskable interrupt line at the DSP core.
  - Peripheral interrupts are initiated internally by these on-chip peripheral modules: event manager A, event manager B, SPI, SCI, CAN, and ADC. They can be masked both by enable bits for each event in each peripheral and by the CPU's IMR, which can mask each maskable interrupt line at the DSP core.
- **Software-generated interrupts** for the LF240xA devices include:
  - The INTR instruction. This instruction allows initialization of any LF240xA interrupt with software. Its
    operand indicates the interrupt vector location to which the CPU branches. This instruction globally
    disables maskable interrupts (sets the INTM bit to 1).
  - The NMI instruction. This instruction forces a branch to interrupt vector location 24h. This instruction globally disables maskable interrupts. 240xA devices do not have the NMI hardware signal, only software activation is provided.
  - The TRAP instruction. This instruction forces the CPU to branch to interrupt vector location 22h. The TRAP instruction does *not* disable maskable interrupts (INTM is not set to 1); therefore, when the CPU branches to the interrupt service routine, that routine can be interrupted by the maskable hardware interrupts.
  - An emulator trap. This interrupt can be generated with either an INTR instruction or a TRAP instruction.

Six core interrupts (INT1–INT6) are expanded using a peripheral interrupt expansion (PIE) module identical to the F24x devices. The PIE manages all the peripheral interrupts from the 240xA peripherals and are grouped to share the six core level interrupts. Figure 7 shows the PIE block diagram for hardware-generated interrupts.

The PIE block diagram (Figure 7) and the interrupt table (Table 3) explain the grouping and interrupt vector maps. LF240xA devices have interrupts identical to those of the F24x devices and should be completely code-compatible. 240xA devices also have peripheral interrupts identical to those of the F24x – plus additional interrupts for new peripherals such as event manager B. Though the new interrupts share the 24x interrupt grouping, they all have a unique vector to differentiate among the interrupts. See Table 3 for details.



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### device reset and interrupts (continued)

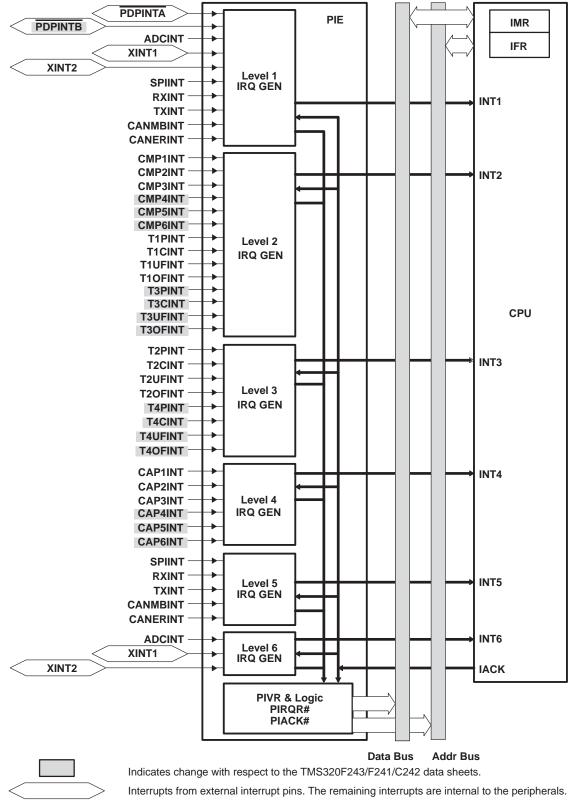


Figure 7. Peripheral Interrupt Expansion (PIE) Module Block Diagram for Hardware-Generated Interrupts



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### interrupt request structure

INTERRUPT NAME	OVERALL PRIORITY	- AND POSITION IN		AND VECTOR A		SOURCE PERIPHERAL MODULE	DESCRIPTION		
Reset	1	RSN 0000h		N/A	N	RS pin, Watchdog	Reset from pin, watchdog timeout		
Reserved	2	_ 0026h		N/A	N	CPU	Emulator trap		
NMI	3	NMI 0024h		N/A	N	Nonmaskable Interrupt	Nonmaskable interrupt, software interrupt only		
PDPINTA	4		0.0	0020h	Y	EVA	Power device protection		
PDPINTB	5		2.0	0019h	Y	EVB	interrupt pins		
ADCINT	6		0.1	0004h	Y	ADC	ADC interrupt in high-priority mode		
XINT1	7		0.2	0001h	Y	External Interrupt Logic	External interrupt pins in high		
XINT2	8		0.3	0011h	Y	External Interrupt Logic	priority		
SPIINT	9	INT1 0002h	0.4	0005h	Y	SPI	SPI interrupt pins in high priority		
RXINT	10		0.5	0006h	Y	SCI	SCI receiver interrupt in high-priority mode		
TXINT	11	]	0.6	0007h	Y	SCI	SCI transmitter interrupt in high-priority mode		
CANMBINT	12		0.7	0040	Y	CAN	CAN mailbox in high-priority mode		
CANERINT	13		0.8	0041	Y	CAN	CAN error interrupt in high-priority mode		
CMP1INT	14		0.9	0021h	Y	EVA	Compare 1 interrupt		
CMP2INT	15		0.10	0022h	Y	EVA	Compare 2 interrupt		
CMP3INT	16		0.11	0023h	Y	EVA	Compare 3 interrupt		
T1PINT	17	INT2	0.12	0027h	Y	EVA	Timer 1 period interrupt		
T1CINT	18	0004h	0.13	0028h	Y	EVA	Timer 1 compare interrupt		
T1UFINT	19		0.14	0029h	Y	EVA	Timer 1 underflow interrupt		
T10FINT	20		0.15	002Ah	Y	EVA	Timer 1 overflow interrupt		
CMP4INT	21		2.1	0024h	Y	EVB	Compare 4 interrupt		
CMP5INT	22		2.2	0025h	Y	EVB	Compare 5 interrupt		
CMP6INT	23		2.3	0026h	Y	EVB	Compare 6 interrupt		
T3PINT	24		2.4	002Fh	Y	EVB	Timer 3 period interrupt		
T3CINT	25		2.5	0030h	Y	EVB	Timer 3 compare interrupt		
T3UFINT	26		2.6	0031h	Y	EVB	Timer 3 underflow interrupt		
T3OFINT	27		2.7	0032h	Y	EVB	Timer 3 overflow interrupt		

### Table 3. LF240xA/LC240xA Interrupt Source Priority and Vectors

† Refer to the TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals (literature number SPRU357A) for more information.

NOTE: Some interrupts may not be available in a particular device due to the absence of a peripheral. See Table 1 for more details.

New peripheral interrupts and vectors with respect to the F243/F241 devices.



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### interrupt request structure (continued)

### Table 3. LF240xA/LC240xA Interrupt Source Priority and Vectors (Continued)

INTERRUPT NAME	AND		POSITION IN PIRQRx AND	SITION ININTERRUPTQRx ANDVECTOR		SOURCE PERIPHERAL MODULE	DESCRIPTION	
T2PINT	28		1.0	002Bh	Y	EVA	Timer 2 period interrupt	
T2CINT	29	1	1.1	002Ch	Y	EVA	Timer 2 compare interrupt	
T2UFINT	30	1	1.2	002Dh	Y	EVA	Timer 2 underflow interrupt	
T2OFINT	31	INT3	1.3	002Eh	Y	EVA	Timer 2 overflow interrupt	
T4PINT	32	0006h	2.8	0039h	Y	EVB	Timer 4 period interrupt	
T4CINT	33	1	2.9	003Ah	Y	EVB	Timer 4 compare interrupt	
T4UFINT	34	1	2.10	003Bh	Y	EVB	Timer 4 underflow interrupt	
T4OFINT	35	1	2.11	003Ch	Y	EVB	Timer 4 overflow interrupt	
CAP1INT	36		1.4	0033h	Y	EVA	Capture 1 interrupt	
CAP2INT	37		1.5	0034h	Y	EVA	Capture 2 interrupt	
CAP3INT	38	INT4	1.6	0035h	Y	EVA	Capture 3 interrupt	
CAP4INT	39	0008h	2.12	0036h	Y	EVB	Capture 4 interrupt	
CAP5INT	40	1	2.13	0037h	Y	EVB	Capture 5 interrupt	
CAP6INT	41		2.14	0038h	Y	EVB	Capture 6 interrupt	
SPIINT	42		1.7	0005h	Y	SPI	SPI interrupt (low priority)	
RXINT	43		1.8	0006h	Y	SCI	SCI receiver interrupt (low-priority mode)	
TXINT	44	INT5	1.9	0007h	Y	SCI	SCI transmitter interrupt (low-priority mode)	
CANMBINT	45	000Ah	1.10	0040h	Y	CAN	CAN mailbox interrupt (low-priority mode)	
CANERINT	46		1.11	0041h	Y	CAN	CAN error interrupt (low-priority mode)	
ADCINT	47		1.12	0004h	Y	ADC	ADC interrupt (low priority)	
XINT1	48	INT6 000Ch	1.13	0001h	Y	External Interrupt Logic	External interrupt pins	
XINT2	NT2 49		1.14	0011h	Y	External Interrupt Logic	(low-priority mode)	
Reserved		000Eh		N/A	Y	CPU	Analysis interrupt	
TRAP	N/A	0022h		N/A	N/A	CPU	TRAP instruction	
Phantom Interrupt Vector	N/A	N/A		0000h	N/A	CPU	Phantom interrupt vector	
INT8–INT16	N/A	0010h-0020h		N/A	N/A	CPU	<b>o</b> <i>t</i>	
INT20–INT31	N/A	00028h-0603Fh		N/A	N/A	CPU	Software interrupt vectors	

<sup>†</sup> Refer to the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A) for more information.

NOTE: Some interrupts may not be available in a particular device due to the absence of a peripheral. See Table 1 for more details.

New peripheral interrupts and vectors with respect to the F243/F241 devices.



### **DSP CPU core**

The TMS320x240xA devices use an advanced Harvard-type architecture that maximizes processing power by maintaining two separate memory bus structures — program and data — for full-speed execution. This multiple bus structure allows data and instructions to be read simultaneously. Instructions support data transfers between program memory and data memory. This architecture permits coefficients that are stored in program memory to be read in RAM, thereby eliminating the need for a separate coefficient ROM. This, coupled with a four-deep pipeline, allows the LF240xA/LC240xA devices to execute most instructions in a single cycle. See the functional block diagram of the 240xA DSP CPU for more information.

### TMS320x240xA instruction set

The x240xA microprocessor implements a comprehensive instruction set that supports both numeric-intensive signal-processing operations and general-purpose applications, such as multiprocessing and high-speed control.

For maximum throughput, the next instruction is prefetched while the current one is being executed. Because the same data lines are used to communicate to external data, program, or I/O space, the number of cycles an instruction requires to execute varies, depending upon whether the next data operand fetch is from internal or external memory. Highest throughput is achieved by maintaining data memory on chip and using either internal or fast external program memory.

### addressing modes

The TMS320x240xA instruction set provides four basic memory-addressing modes: direct, indirect, immediate, and register.

In direct addressing, the instruction word contains the lower seven bits of the data memory address. This field is concatenated with the nine bits of the data memory page pointer (DP) to form the 16-bit data memory address. Therefore, in the direct-addressing mode, data memory is paged effectively with a total of 512 pages, with each page containing 128 words.

Indirect addressing accesses data memory through the auxiliary registers. In this addressing mode, the address of the instruction operand is contained in the currently selected auxiliary register. Eight auxiliary registers (AR0–AR7) provide flexible and powerful indirect addressing. To select a specific auxiliary register, the auxiliary register pointer (ARP) is loaded with a value from 0 to 7 for AR0 through AR7, respectively.

### scan-based emulation

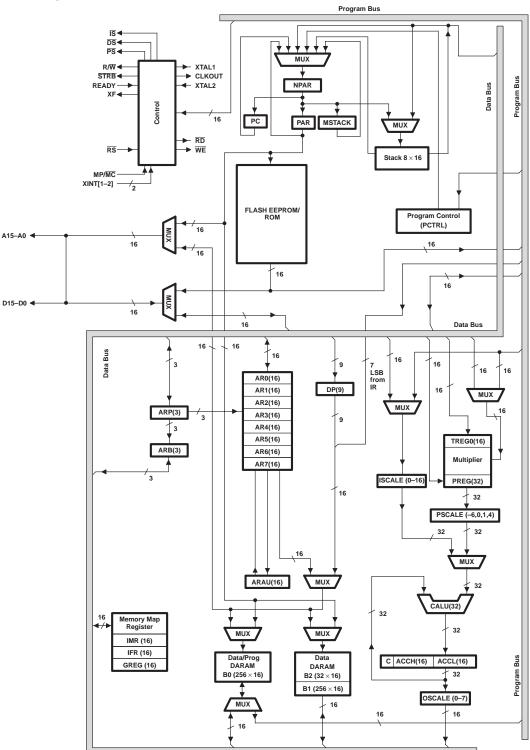
TMS320x2xx devices incorporate scan-based emulation logic for code-development and hardwaredevelopment support. Scan-based emulation allows the emulator to control the processor in the system without the use of intrusive cables to the full pinout of the device. The scan-based emulator communicates with the x2xx by way of the IEEE 1149.1-compatible (JTAG) interface. The x240xA DSPs do not include boundary scan. The scan chain of these devices is useful for emulation function only.





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### functional block diagram of the 240xA DSP CPU



- NOTES: A. See Table 4 for symbol descriptions.
  - B. For clarity, the data and program buses are shown as single buses although they include address and data bits.
  - C. Refer to the TMS320F/C24x DSP Controllers Reference Guide: CPU and Instruction Set (literature number SPRU160) for CPU instruction set information.



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### 240xA legend for the internal hardware

SYMBOL	NAME	DESCRIPTION
ACC	Accumulator	32-bit register that stores the results and provides input for subsequent CALU operations. Also includes shift and rotate capabilities
ARAU	Auxiliary Register Arithmetic Unit	An unsigned, 16-bit arithmetic unit used to calculate indirect addresses using the auxiliary registers as inputs and outputs
AUX REGS	Auxiliary Registers 0–7	These 16-bit registers are used as pointers to anywhere within the data space address range. They are operated upon by the ARAU and are selected by the auxiliary register pointer (ARP). AR0 can also be used as an index value for AR updates of more than one and as a compare value to AR.
С	Carry	Register carry output from CALU. C is fed back into the CALU for extended arithmetic operation. The C bit resides in status register 1 (ST1), and can be tested in conditional instructions. C is also used in accumulator shifts and rotates.
CALU	Central Arithmetic Logic Unit	32-bit-wide main arithmetic logic unit for the TMS320C2xx core. The CALU executes 32-bit operations in a single machine cycle. CALU operates on data coming from ISCALE or PSCALE with data from ACC, and provides status results to PCTRL.
DARAM	Dual-Access RAM	If the on-chip RAM configuration control bit (CNF) is set to 0, the reconfigurable data dual-access RAM (DARAM) block B0 is mapped to data space; otherwise, B0 is mapped to program space. Blocks B1 and B2 are mapped to data memory space only, at addresses 0300–03FF and 0060–007F, respectively. Blocks 0 and 1 contain 256 words, while block 2 contains 32 words.
DP	Data Memory Page Pointer	The 9-bit DP register is concatenated with the seven least significant bits (LSBs) of an instruction word to form a direct memory address of 16 bits. DP can be modified by the LST and LDP instructions.
GREG	Global Memory Allocation Register	GREG specifies the size of the global data memory space. Since the global memory space is not used in the 240xA devices, this register is reserved.
IMR	Interrupt Mask Register	IMR individually masks or enables the seven interrupts.
IFR	Interrupt Flag Register	The 7-bit IFR indicates that the TMS320C2xx has latched an interrupt from one of the seven maskable interrupts.
INT#	Interrupt Traps	A total of 32 interrupts by way of hardware and/or software are available.
ISCALE	Input Data-Scaling Shifter	16- to 32-bit barrel left-shifter. ISCALE shifts incoming 16-bit data 0 to16 positions left, relative to the 32-bit output within the fetch cycle; therefore, no cycle overhead is required for input scaling operations.
MPY	Multiplier	$16 \times 16$ -bit multiplier to a 32-bit product. MPY executes multiplication in a single cycle. MPY operates either signed or unsigned 2s-complement arithmetic multiply.
MSTACK	Micro Stack	MSTACK provides temporary storage for the address of the next instruction to be fetched when program address-generation logic is used to generate sequential addresses in data space.
MUX	Multiplexer	Multiplexes buses to a common input
NPAR	Next Program Address Register	NPAR holds the program address to be driven out on the PAB in the next cycle.
OSCALE	Output Data-Scaling Shifter	16- to 32-bit barrel left-shifter. OSCALE shifts the 32-bit accumulator output 0 to 7 bits left for quantization management and outputs either the 16-bit high- or low-half of the shifted 32-bit data to the data-write data bus (DWEB).
PAR	Program Address Register	PAR holds the address currently being driven on PAB for as many cycles as it takes to complete all memory operations scheduled for the current bus cycle.
PC	Program Counter	PC increments the value from NPAR to provide sequential addresses for instruction-fetching and sequential data-transfer operations.
PCTRL	Program Controller	PCTRL decodes instruction, manages the pipeline, stores status, and decodes conditional operations.



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### 240xA legend for the internal hardware (continued)

### Table 4. Legend for the 240xA DSP CPU Internal Hardware (Continued)

SYMBOL	NAME	DESCRIPTION
PREG	Product Register	32-bit register holds results of $16 \times 16$ multiply
PSCALE	Product-Scaling Shifter	0-, 1-, or 4-bit left shift, or 6-bit right shift of multiplier product. The left-shift options are used to manage the additional sign bits resulting from the 2s-complement multiply. The right-shift option is used to scale down the number to manage overflow of product accumulation in the CALU. PSCALE resides in the path from the 32-bit product shifter and from either the CALU or the data-write data bus (DWEB), and requires no cycle overhead.
STACK	Stack	STACK is a block of memory used for storing return addresses for subroutines and interrupt-service routines, or for storing data. The C2xx stack is 16 bits wide and 8 levels deep.
TREG	Temporary Register	16-bit register holds one of the operands for the multiply operations. TREG holds the dynamic shift count for the LACT, ADDT, and SUBT instructions. TREG holds the dynamic bit position for the BITT instruction.

### status and control registers

Two status registers, ST0 and ST1, contain the status of various conditions and modes. These registers can be stored into data memory and loaded from data memory, thus allowing the status of the machine to be saved and restored for subroutines.

The load status register (LST) instruction is used to write to ST0 and ST1. The store status register (SST) instruction is used to read from ST0 and ST1 — except for the INTM bit, which is not affected by the LST instruction. The individual bits of these registers can be set or cleared when using the SETC and CLRC instructions. Figure 8 shows the organization of status registers ST0 and ST1, indicating all status bits contained in each. Several bits in the status registers are reserved and are read as logic 1s. Table 5 lists status register field definitions.

	15		13	12	11	10	9	8								0
ST0		ARP		٥V	OVM	1	INTM					DP				
	15		13	12	11	10	9	8	7	6	5	4	3	2	1	0
ST1		ARB		CNF	тс	SXM	С	1	1	1	1	XF	1	1	PN	I

### Figure 8. Organization of Status Registers ST0 and ST1

### Table 5. Status Register Field Definitions

FIELD	FUNCTION
ARB	Auxiliary register pointer buffer. When the ARP is loaded into ST0, the old ARP value is copied to the ARB except during an LST instruction. When the ARB is loaded by way of an LST #1 instruction, the same value is also copied to the ARP.
ARP	Auxiliary register (AR) pointer. ARP selects the AR to be used in indirect addressing. When the ARP is loaded, the old ARP value is copied to the ARB register. ARP can be modified by memory-reference instructions when using indirect addressing, and by the LARP, MAR, and LST instructions. The ARP is also loaded with the same value as ARB when an LST #1 instruction is executed.
с	Carry bit. C is set to 1 if the result of an addition generates a carry, or reset to 0 if the result of a subtraction generates a borrow. Otherwise, C is reset after an addition or set after a subtraction, except if the instruction is ADD or SUB with a 16-bit shift. In these cases, ADD can only set and SUB can only reset the carry bit, but cannot affect it otherwise. The single-bit shift and rotate instructions also affect C, as well as the SETC, CLRC, and LST #1 instructions. Branch instructions have been provided to branch on the status of C. C is set to 1 on a reset.
CNF	On-chip RAM configuration control bit. If CNF is set to 0, the reconfigurable data dual-access RAM blocks are mapped to data space; otherwise, they are mapped to program space. The CNF can be modified by the SETC CNF, CLRC CNF, and LST #1 instructions. RS sets the CNF to 0.



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### status and control registers (continued)

### Table 5. Status Register Field Definitions (Continued)

FIELD	FUNCTION
DP	Data memory page pointer. The 9-bit DP register is concatenated with the 7 LSBs of an instruction word to form a direct memory address of 16 bits. DP can be modified by the LST and LDP instructions.
INTM	Interrupt mode bit. When INTM is set to 0, all unmasked interrupts are enabled. When set to 1, all maskable interrupts are disabled. INTM is set and reset by the SETC INTM and CLRC INTM instructions. RS also sets INTM. INTM has no effect on the unmaskable RS and NMI interrupts. Note that INTM is unaffected by the LST instruction. This bit is set to 1 by reset. It is also set to 1 when a maskable interrupt trap is taken.
OV	Overflow flag bit. As a latched overflow signal, OV is set to 1 when overflow occurs in the arithmetic logic unit (ALU). Once an overflow occurs, the OV remains set until a reset, BCND/D on OV/NOV, or LST instruction clears OV.
OVM	Overflow mode bit. When OVM is set to 0, overflowed results overflow normally in the accumulator. When set to 1, the accumulator is set to either its most positive or negative value upon encountering an overflow. The SETC and CLRC instructions set and reset this bit, respectively. LST can also be used to modify the OVM.
РМ	Product shift mode. If these two bits are 00, the multiplier's 32-bit product is loaded into the ALU with no shift. If $PM = 01$ , the PREG output is left-shifted one place and loaded into the ALU, with the LSB zero-filled. If $PM = 10$ , the PREG output is left-shifted by 4 bits and loaded into the ALU, with the LSBs zero-filled. PM = 11 produces a right shift of 6 bits, sign-extended. Note that the PREG contents remain unchanged. The shift takes place when transferring the contents of the PREG to the ALU. PM is loaded by the SPM and LST #1 instructions. PM is cleared by RS.
SXM	Sign-extension mode bit. SXM = 1 produces sign extension on data as it is passed into the accumulator through the scaling shifter. SXM = 0 suppresses sign extension. SXM does not affect the definitions of certain instructions; for example, the ADDS instruction suppresses sign extension regardless of SXM. SXM is set by the SETC SXM instruction and reset by the CLRC SXM instruction and can be loaded by the LST #1 instruction. SXM is set to 1 by reset.
тс	Test/control flag bit. TC is affected by the BIT, BITT, CMPR, LST #1, and NORM instructions. TC is set to a 1 if a bit tested by BIT or BITT is a 1, if a compare condition tested by CMPR exists between AR (ARP) and AR0, if the exclusive-OR function of the 2 most significant bits (MSBs) of the accumulator is true when tested by a NORM instruction. The conditional branch, call, and return instructions can execute based on the condition of TC.
XF	XF pin status bit. XF indicates the state of the XF pin, a general-purpose output pin. XF is set by the SETC XF instruction and reset by the CLRC XF instruction. XF is set to 1 by reset.

### central processing unit

The TMS320x240xA central processing unit (CPU) contains a 16-bit scaling shifter, a 16 x 16-bit parallel multiplier, a 32-bit central arithmetic logic unit (CALU), a 32-bit accumulator, and additional shifters at the outputs of both the accumulator and the multiplier. This section describes the CPU components and their functions. The functional block diagram shows the components of the CPU.

### input scaling shifter

The TMS320x240xA provides a scaling shifter with a 16-bit input connected to the data bus and a 32-bit output connected to the CALU. This shifter operates as part of the path of data coming from program or data space to the CALU and requires no cycle overhead. It is used to align the 16-bit data coming from memory to the 32-bit CALU. This is necessary for scaling arithmetic as well as aligning masks for logical operations.

The scaling shifter produces a left shift of 0 to 16 on the input data. The LSBs of the output are filled with zeros; the MSBs can either be filled with zeros or sign-extended, depending upon the value of the SXM bit (sign-extension mode) of status register ST1. The shift count is specified by a constant embedded in the instruction word or by a value in TREG. The shift count in the instruction allows for specific scaling or alignment operations specific to that point in the code. The TREG base shift allows the scaling factor to be adaptable to the system's performance.



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### multiplier

The TMS320x240xA devices use a 16 x 16-bit hardware multiplier that is capable of computing a signed or an unsigned 32-bit product in a single machine cycle. All multiply instructions, except the MPYU (multiply unsigned) instruction, perform a signed multiply operation. That is, two numbers being multiplied are treated as 2s-complement numbers, and the result is a 32-bit 2s-complement number. There are two registers associated with the multiplier, as follow:

- 16-bit temporary register (TREG) that holds one of the operands for the multiplier
- 32-bit product register (PREG) that holds the product

Four product-shift modes (PM) are available at the PREG output (PSCALE). These shift modes are useful for performing multiply/accumulate operations, performing fractional arithmetic, or justifying fractional products. The PM field of status register ST1 specifies the PM shift mode, as shown in Table 6.

### Table 6. PSCALE Product-Shift Modes

РМ	SHIFT	DESCRIPTION
00	No shift	Product feed to CALU or data bus with no shift
01	Left 1	Removes the extra sign bit generated in a 2s-complement multiply to produce a Q31 product
10	Left 4	Removes the extra 4 sign bits generated in a 16x13 2s-complement multiply to a produce a Q31 product when using the multiply-by-a-13-bit constant
11	Right 6	Scales the product to allow up to 128 product accumulation without the possibility of accumulator overflow

The product can be shifted one bit to compensate for the extra sign bit gained in multiplying two 16-bit 2s-complement numbers (MPY instruction). A four-bit shift is used in conjunction with the MPY instruction with a short immediate value (13 bits or less) to eliminate the four extra sign bits gained in multiplying a 16-bit number by a 13-bit number. Finally, the output of PREG can be right-shifted 6 bits to enable the execution of up to 128 consecutive multiply/accumulates without the possibility of overflow.

The LT (load TREG) instruction normally loads TREG to provide one operand (from the data bus), and the MPY (multiply) instruction provides the second operand (also from the data bus). A multiplication also can be performed with a 13-bit immediate operand when using the MPY instruction. Then, a product is obtained every two cycles. When the code is executing multiple multiplies and product sums, the CPU supports the pipelining of the TREG load operations with CALU operations using the previous product. The pipeline operations that run in parallel with loading the TREG include: load ACC with PREG (LTP); add PREG to ACC (LTA); add PREG to ACC and shift TREG input data (DMOV) to next address in data memory (LTD); and subtract PREG from ACC (LTS).

Two multiply/accumulate instructions (MAC and MACD) fully utilize the computational bandwidth of the multiplier, allowing both operands to be processed simultaneously. The data for these operations can be transferred to the multiplier each cycle by way of the program and data buses. This facilitates single-cycle multiply/accumulates when used with the repeat (RPT) instruction. In these instructions, the coefficient addresses are generated by program address generation (PAGEN) logic, while the data addresses are generated by data address generation (DAGEN) logic. This allows the repeated instruction to access the values from the coefficient table sequentially and step through the data in any of the indirect addressing modes.

The MACD instruction, when repeated, supports filter constructs (weighted running averages) so that as the sum-of-products is executed, the sample data is shifted in memory to make room for the next sample and to throw away the oldest sample.



### multiplier (continued)

The MPYU instruction performs an unsigned multiplication, which greatly facilitates extended-precision arithmetic operations. The unsigned contents of TREG are multiplied by the unsigned contents of the addressed data memory location, with the result placed in PREG. This process allows the operands of greater than 16 bits to be broken down into 16-bit words and processed separately to generate products of greater than 32 bits. The SQRA (square/add) and SQRS (square/subtract) instructions pass the same value to both inputs of the multiplier for squaring a data memory value.

After the multiplication of two 16-bit numbers, the 32-bit product is loaded into the 32-bit product register (PREG). The product from PREG can be transferred to the CALU or to data memory by way of the SPH (store product high) and SPL (store product low) instructions. Note: the transfer of PREG to either the CALU or data bus passes through the PSCALE shifter, and therefore is affected by the product shift mode defined by PM. This is important when saving PREG in an interrupt-service-routine context save as the PSCALE shift effects cannot be modeled in the restore operation. PREG can be cleared by executing the MPY #0 instruction. The product register can be restored by loading the saved low half into TREG and executing a MPY #1 instruction. The high half, then, is loaded using the LPH instruction.

### central arithmetic logic unit

The TMS320x240xA central arithmetic logic unit (CALU) implements a wide range of arithmetic and logical functions, the majority of which execute in a single clock cycle. This ALU is referred to as central to differentiate it from a second ALU used for indirect-address generation called the auxiliary register arithmetic unit (ARAU). Once an operation is performed in the CALU, the result is transferred to the accumulator (ACC) where additional operations, such as shifting, can occur. Data that is input to the CALU can be scaled by ISCALE when coming from one of the data buses (DRDB or PRDB) or scaled by PSCALE when coming from the multiplier.

The CALU is a general-purpose ALU that operates on 16-bit words taken from data memory or derived from immediate instructions. In addition to the usual arithmetic instructions, the CALU can perform Boolean operations, facilitating the bit-manipulation ability required for a high-speed controller. One input to the CALU is always provided from the accumulator, and the other input can be provided from the product register (PREG) of the multiplier or the output of the scaling shifter (that has been read from data memory or from the ACC). After the CALU has performed the arithmetic or logical operation, the result is stored in the accumulator.

The TMS320x240xA devices support floating-point operations for applications requiring a large dynamic range. The NORM (normalization) instruction is used to normalize fixed-point numbers contained in the accumulator by performing left shifts. The four bits of the TREG define a variable shift through the scaling shifter for the LACT/ADDT/SUBT (load/add to/subtract from accumulator with shift specified by TREG) instructions. These instructions are useful in floating-point arithmetic where a number needs to be denormalized — that is, floating-point to fixed-point conversion. They are also useful in the execution of an automatic gain control (AGC) going into a filter. The BITT (bit test) instruction provides testing of a single bit of a word in data memory based on the value contained in the four LSBs of TREG.

The CALU overflow saturation mode can be enabled/disabled by setting/resetting the OVM bit of ST0. When the CALU is in the overflow saturation mode and an overflow occurs, the overflow flag is set and the accumulator is loaded with either the most positive or the most negative value representable in the accumulator, depending on the direction of the overflow. The value of the accumulator at saturation is 07FFFFFFF (positive) or 080000000h (negative). If the OVM (overflow mode) status register bit is reset and an overflow occurs, the overflowed results are loaded into the accumulator with modification. (Note that logical operations cannot result in overflow.)

The CALU can execute a variety of branch instructions that depend on the status of the CALU and the accumulator. These instructions can be executed conditionally based on any meaningful combination of these status bits. For overflow management, these conditions include OV (branch on overflow) and EQ (branch on accumulator equal to zero). In addition, the BACC (branch to address in accumulator) instruction provides the ability to branch to an address specified by the accumulator (computed goto). Bit test instructions (BIT and BITT), which do not affect the accumulator, allow the testing of a specified bit of a word in data memory.



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### central arithmetic logic unit (continued)

The CALU also has an associated carry bit that is set or reset depending on various operations within the device. The carry bit allows more efficient computation of extended-precision products and additions or subtractions. It is also useful in overflow management. The carry bit is affected by most arithmetic instructions as well as the single-bit shift and rotate instructions. It is not affected by loading the accumulator, logical operations, or other such non-arithmetic or control instructions.

The ADDC (add to accumulator with carry) and SUBB (subtract from accumulator with borrow) instructions use the previous value of carry in their addition/subtraction operation.

The one exception to the operation of the carry bit is in the use of ADD with a shift count of 16 (add to high accumulator) and SUB with a shift count of 16 (subtract from high accumulator) instructions. This case of the ADD instruction can set the carry bit only if a carry is generated, and this case of the SUB instruction can reset the carry bit only if a borrow is generated; otherwise, neither instruction affects it.

Two conditional operands, C and NC, are provided for branching, calling, returning, and conditionally executing, based upon the status of the carry bit. The SETC, CLRC, and LST #1 instructions also can be used to load the carry bit. The carry bit is set to one on a hardware reset.

### accumulator

The 32-bit accumulator is the registered output of the CALU. It can be split into two 16-bit segments for storage in data memory. Shifters at the output of the accumulator provide a left shift of 0 to 7 places. This shift is performed while the data is being transferred to the data bus for storage. The contents of the accumulator remain unchanged. When the postscaling shifter is used on the high word of the accumulator (bits 16–31), the MSBs are lost and the LSBs are filled with bits shifted in from the low word (bits 0–15). When the postscaling shifter is used on the low word, the LSBs are zero-filled.

The SFL and SFR (in-place one-bit shift to the left/right) instructions and the ROL and ROR (rotate to the left/right) instructions implement shifting or rotating of the contents of the accumulator through the carry bit. The SXM bit affects the definition of the SFR (shift accumulator right) instruction. When SXM = 1, SFR performs an arithmetic right shift, maintaining the sign of the accumulator data. When SXM = 0, SFR performs a logical shift, shifting out the LSBs and shifting in a zero for the MSB. The SFL (shift accumulator left) instruction is not affected by the SXM bit and behaves the same in both cases, shifting out the MSB and shifting in a zero. Repeat (RPT) instructions can be used with the shift and rotate instructions for multiple-bit shifts.

### auxiliary registers and auxiliary-register arithmetic unit (ARAU)

The 240xA provides a register file containing eight auxiliary registers (AR0–AR7). The auxiliary registers are used for indirect addressing of the data memory or for temporary data storage. Indirect auxiliary-register addressing allows placement of the data memory address of an instruction operand into one of the auxiliary registers. These registers are referenced with a 3-bit auxiliary register pointer (ARP) that is loaded with a value from 0 through 7, designating AR0 through AR7, respectively. The auxiliary registers and the ARP can be loaded from data memory, the ACC, the product register, or by an immediate operand defined in the instruction. The contents of these registers also can be stored in data memory or used as inputs to the CALU.

The auxiliary register file (AR0–AR7) is connected to the ARAU. The ARAU can autoindex the current auxiliary register while the data memory location is being addressed. Indexing either by  $\pm 1$  or by the contents of the AR0 register can be performed. As a result, accessing tables of information does not require the CALU for address manipulation; therefore, the CALU is free for other operations in parallel.



## internal memory

The TMS320x240xA devices are configured with the following memory modules:

- Dual-access random-access memory (DARAM)
- Single-access random-access memory (SARAM)
- Flash
- ROM
- Boot ROM

### dual-access RAM (DARAM)

There are 544 words  $\times$  16 bits of DARAM on the 240xA devices. The 240xA DARAM allows writes to and reads from the RAM in the same cycle. The DARAM is configured in three blocks: block 0 (B0), block 1 (B1), and block 2 (B2). Block 1 contains 256 words and Block 2 contains 32 words, and both blocks are located only in data memory space. Block 0 contains 256 words, and can be configured to reside in either data or program memory space. The SETC CNF (configure B0 as program memory) and CLRC CNF (configure B0 as data memory) instructions allow dynamic configuration of the memory maps through software.

When using on-chip RAM, the 240xA runs at full speed with no wait states. The ability of the DARAM to allow two accesses to be performed in one cycle, coupled with the parallel nature of the 240xA architecture, enables the device to perform three concurrent memory accesses in any given machine cycle. Externally, the READY line or on-chip software wait-state generator can be used to interface the 2407A to slower, less expensive external memory.

### single-access RAM (SARAM)

There are 2K words  $\times$  16 bits of SARAM on some of the 240xA devices.<sup>†</sup> The PON and DON bits select SARAM (2K) mapping in program space, data space, or both. See Table 18 for details on the SCSR2 register and the PON and DON bits. At reset, these bits are 11, and the on-chip SARAM is mapped in both the program and data spaces. The SARAM (starting at 8000h in program memory) is accessible in external memory space (for 2407A only), if the on-chip SARAM is not enabled.

### flash EEPROM

Flash EEPROM provides an attractive alternative to masked program ROM. Like ROM, Flash is nonvolatile. However, it has the advantage of "in-target" reprogrammability. The LF2407A incorporates one  $32K \times 16$ -bit Flash EEPROM module in program space. The Flash module has multiple sectors that can be individually protected while erasing or programming. The sector size is non-uniform and partitioned as 4K/12K/12K/4K sectors.

Unlike most discrete Flash memory, the LF240xA Flash does not require a dedicated state machine, because the algorithms for programming and erasing the Flash are executed by the DSP core. This enables several advantages, including: reduced chip size and sophisticated, adaptive algorithms. For production programming, the IEEE Standard 1149.1<sup>‡</sup> (JTAG) scan port provides easy access to the on-chip RAM for downloading the algorithms and Flash code. This Flash requires 5 V for programming (at V<sub>CCP</sub> pin only) the array. The Flash runs at zero wait state while the device is powered at 3.3 V.

<sup>†</sup> See Table 1 for device-specific features.

<sup>‡</sup>IEEE Standard 1149.1–1990, IEEE Standard Test Access Port.



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## ROM

The LC240xA devices contain mask-programmable ROM located in program memory space. Customers can arrange to have this ROM programmed with contents unique to any particular application. See Table 1 for the ROM memory capacity of each LC240xA device.

### boot ROM (LF240xA only)

Boot ROM is a 256-word ROM memory mapped in program space 0000–00FF. This ROM will be enabled if the BOOT\_EN pin is low during reset. The BOOT\_EN bit (bit 3 of the SCSR2 register) will be set to 0 if the BOOT\_EN pin is low at reset. Boot ROM can also be enabled by writing 0 to the SCSR2.3 bit and disabled by writing 1 to this bit.

The boot ROM has a generic bootloader to transfer code through SCI or SPI ports. The incoming code should disable the BOOT\_ROM bit by writing 1 to bit 3 of the SCSR2 register, or else, the whole Flash array will not be enabled.

The boot ROM code sets the PLL to x2 or x4 option based on the condition of the SCITXD pin during reset. The SCITXD pin should be pulled high/low to select the PLL multiplication factor. The choices made are as follows:

- If the SCITXD pin is pulled low, the PLL multiplier is set to 2.
- If the SCITXD pin is pulled high, the PLL multiplier is set to 4. (Default)
- If the SCITXD pin is not driven at reset, the internal pullup selects the default multiplier of 4.

Care should be taken such that a combination of CLKIN and the PLL multiplication factor should not result in a CPU clock speed of greater than 40 MHz, the maximum rated speed.

Furthermore, when the bootloader is used, only specific values of CLKIN would result in a baud-lock for the SCI. Refer to the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A) for more details about the bootloader operation.

### flash/ROM security

240xA devices incorporate a security feature that prevents external access to program memory. This feature is useful in preventing unauthorized duplication of proprietary code.

If access to Flash/ROM contents are desired for debugging purposes, two actions need to be taken:

- 1. A "dummy" read of locations 40h, 41h, 42h and 43h (of program memory space) is necessary. The word "dummy" indicates that the destination address of this read is insignificant.
- 2. A 64-bit password (split as four 16-bit words) must be written to the data-memory locations 77F0h, 77F1h, 77F2h, and 77F3h. The four 16-bit words written to these locations must match the four words stored in 40h, 41h, 42h, and 43h (of program memory space), respectively. The device becomes "unsecured" one cycle after the last instruction that unsecures the part.



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# PERIPHERALS

The integrated peripherals of the TMS320x240xA are described in the following subsections:

- Two event-manager modules (EVA, EVB)
- Enhanced analog-to-digital converter (ADC) module
- Controller area network (CAN) module
- Serial communications interface (SCI) module
- Serial peripheral interface (SPI) module
- PLL-based clock module
- Digital I/O and shared pin functions
- External memory interfaces (LF2407A only)
- Watchdog (WD) timer module

## event manager modules (EVA, EVB)

The event-manager modules include general-purpose (GP) timers, full-compare/PWM units, capture units, and quadrature-encoder pulse (QEP) circuits. EVA's and EVB's timers, compare units, and capture units function identically. However, timer/unit names differ for EVA and EVB. Table 7 shows the module and signal names used. Table 7 shows the features and functionality available for the event-manager modules and highlights EVA nomenclature.

Event managers A and B have identical peripheral register sets with EVA starting at 7400h and EVB starting at 7500h. The paragraphs in this section describe the function of GP timers, compare units, capture units, and QEPs using EVA nomenclature. These paragraphs are applicable to EVB with regard to function—however, module/signal names would differ.

	E\	/Α	E\	/В
EVENT MANAGER MODULES	MODULE	SIGNAL	MODULE	SIGNAL
GP Timers	Timer 1	T1PWM/T1CMP	Timer 3	T3PWM/T3CMP
	Timer 2	T2PWM/T2CMP	Timer 4	T4PWM/T4CMP
Compare Units	Compare 1	PWM1/2	Compare 4	PWM7/8
	Compare 2	PWM3/4	Compare 5	PWM9/10
	Compare 3	PWM5/6	Compare 6	PWM11/12
Capture Units	Capture 1	CAP1	Capture 4	CAP4
	Capture 2	CAP2	Capture 5	CAP5
	Capture 3	CAP3	Capture 6	CAP6
QEP	QEP1	QEP1	QEP3	QEP3
	QEP2	QEP2	QEP4	QEP4
External Inputs	Direction	TDIRA	Direction	TDIRB
	External Clock	TCLKINA	External Clock	TCLKINB

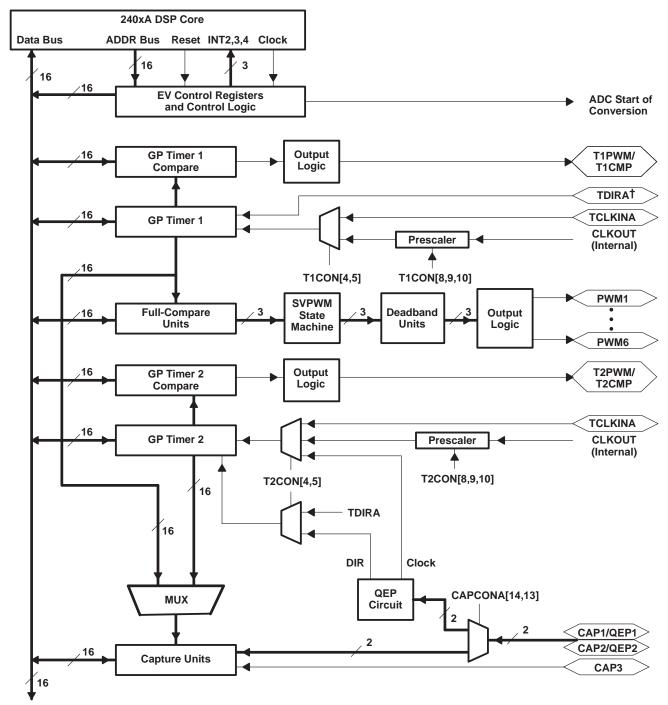
### Table 7. Module and Signal Names for EVA and EVB





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## event manager modules (EVA, EVB) (continued)



<sup>†</sup>2402A devices do not support external direction control. TDIR is not available.

Figure 9. Event Manager A Block Diagram



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## general-purpose (GP) timers

There are two GP timers. The GP timer x (x = 1 or 2 for EVA; x = 3 or 4 for EVB) includes:

- A 16-bit timer, up-/down-counter, TxCNT, for reads or writes
- A 16-bit timer-compare register, TxCMPR (double-buffered with shadow register), for reads or writes
- A 16-bit timer-period register, TxPR (double-buffered with shadow register), for reads or writes
- A 16-bit timer-control register, TxCON, for reads or writes
- Selectable internal or external input clocks
- A programmable prescaler for internal or external clock inputs
- Control and interrupt logic, for four maskable interrupts: *underflow, overflow, timer compare, and period interrupts*
- A selectable direction input pin (TDIRx) (to count up or down when directional up-/down-count mode is selected)

The GP timers can be operated independently or synchronized with each other. The compare register associated with each GP timer can be used for compare function and PWM-waveform generation. There are three continuous modes of operations for each GP timer in up- or up/down-counting operations. Internal or external input clocks with programmable prescaler are used for each GP timer. GP timers also provide the time base for the other event-manager submodules: GP timer 1 for all the compares and PWM circuits, GP timer 2/1 for the capture units and the quadrature-pulse counting operations. Double-buffering of the period and compare registers allows programmable change of the timer (PWM) period and the compare/PWM pulse width as needed.

#### full-compare units

There are three full-compare units on each event manager. These compare units use GP timer1 as the time base and generate six outputs for compare and PWM-waveform generation using programmable deadband circuit. The state of each of the six outputs is configured independently. The compare registers of the compare units are double-buffered, allowing programmable change of the compare/PWM pulse widths as needed.

### programmable deadband generator

The deadband generator circuit includes three 8-bit counters and an 8-bit compare register. Desired deadband values (from 0 to 16  $\mu$ s) can be programmed into the compare register for the outputs of the three compare units. The deadband generation can be enabled/disabled for each compare unit output individually. The deadband-generator circuit produces two outputs (with or without deadband zone) for each compare unit output signal. The output states of the deadband generator are configurable and changeable as needed by way of the double-buffered ACTR register.

### PWM waveform generation

Up to eight PWM waveforms (outputs) can be generated simultaneously by each event manager: three independent pairs (six outputs) by the three full-compare units with *programmable deadbands*, and two independent PWMs by the GP-timer compares.



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### **PWM** characteristics

Characteristics of the PWMs are as follows:

- 16-bit registers
- Programmable deadband for the PWM output pairs, from 0 to 12 μs
- Minimum deadband width of 25 ns
- Change of the PWM carrier frequency for PWM frequency wobbling as needed
- Change of the PWM pulse widths within and after each PWM period as needed
- External-maskable power and drive-protection interrupts
- Pulse-pattern-generator circuit, for programmable generation of asymmetric, symmetric, and four-space vector PWM waveforms
- Minimized CPU overhead using auto-reload of the compare and period registers
- The PWM pins are driven to a high-impedance state when the PDPINTx pin is driven low and after PDPINTx signal qualification. The PDPINTx pin (after qualification) is reflected in bit 8 of the COMCONx register.
  - PDPINTA pin status is reflected in bit 8 of COMCONA register.
  - PDPINTB pin status is reflected in bit 8 of COMCONB register.

#### capture unit

The capture unit provides a logging function for different events or transitions. The values of the selected GP timer counter is captured and stored in the two-level-deep FIFO stacks when selected transitions are detected on capture input pins, CAPx (x = 1, 2, or 3 for EVA; and x = 4, 5, or 6 for EVB). The capture unit consists of three capture circuits.

- Capture units include the following features:
  - One 16-bit capture control register, CAPCONx (R/W)
  - One 16-bit capture FIFO status register, CAPFIFOx
  - Selection of GP timer 1/2 (for EVA) or 3/4 (for EVB) as the time base
  - Three 16-bit 2-level-deep FIFO stacks, one for each capture unit
  - Three capture input pins (CAP1/2/3 for EVA, CAP4/5/6 for EVB)—one input pin per capture unit. [All inputs are synchronized with the device (CPU) clock. In order for a transition to be captured, the input must hold at its current level to meet two rising edges of the device clock. The input pins CAP1/2 and CAP4/5 can also be used as QEP inputs to the QEP circuit.]
  - User-specified transition (rising edge, falling edge, or both edges) detection
  - Three maskable interrupt flags, one for each capture unit

### quadrature-encoder pulse (QEP) circuit

Two capture inputs (CAP1 and CAP2 for EVA; CAP4 and CAP5 for EVB) can be used to interface the on-chip QEP circuit with a quadrature encoder pulse. Full synchronization of these inputs is performed on-chip. Direction or leading-quadrature pulse sequence is detected, and GP timer 2/4 is incremented or decremented by the rising and falling edges of the two input signals (four times the frequency of either input pulse).



## input qualifier circuitry

An input-qualifier circuitry qualifies the input signal to the CAP1–6, XINT1/2, ADCSOC and PDPINTA/B pins in the 240xA devices. (The I/O functions of these pins do not use the input-qualifier circuitry). The state of the internal input signal will change only after the pin is high/low for 6(12) clock edges. This ensures that a glitch smaller than 5(11) CLKOUT cycles wide will not change the internal pin input state. The user must hold the pin high/low for 6(12) cycles to ensure the device will see the level change. Bit 6 of the SCSR2 register controls whether 6 clock edges (bit 6 = 0) or 12 clock edges (bit 6 = 1) are used to block 5- or 11-cycle glitches. On the LC2402A, input qualification is for the CAP1, CAP2, CAP3, PDPINTA, and XINT2/ADCSOC pins.

## enhanced analog-to-digital converter (ADC) module

A simplified functional block diagram of the ADC module is shown in Figure 10. The ADC module consists of a 10-bit ADC with a built-in sample-and-hold (S/H) circuit. Functions of the ADC module include:

- 10-bit ADC core with built-in S/H
- Fast conversion time (S/H + Conversion) of 500 ns
- 16-channel, muxed inputs
- Autosequencing capability provides up to 16 "autoconversions" in a single session. Each conversion can be programmed to select any 1 of 16 input channels
- Sequencer can be operated as two independent 8-state sequencers or as one large 16-state sequencer (i.e., two cascaded 8-state sequencers)
- Sixteen result registers (individually addressable) to store conversion values
  - The digital value of the input analog voltage is derived by:

$$\label{eq:Digital Value} \begin{array}{rcl} \text{Digital Value} &=& 1023 \times \frac{\text{Input Analog Voltage} - V_{\text{REFLO}}}{V_{\text{REFHI}} - V_{\text{REFLO}}} \end{array}$$

- Multiple triggers as sources for the start-of-conversion (SOC) sequence
  - S/W software immediate start
  - EVA Event manager A (multiple event sources within EVA)
  - EVB Event manager B (multiple event sources within EVB)
  - Ext External pin (ADCSOC)
- Flexible interrupt control allows interrupt request on every end-of-sequence (EOS) or every other EOS
- Sequencer can operate in "start/stop" mode, allowing multiple "time-sequenced triggers" to synchronize conversions
- EVA and EVB triggers can operate independently in dual-sequencer mode
- Sample-and-hold (S/H) acquisition time window has separate prescale control
- Built-in calibration mode
- Built-in self-test mode



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# enhanced analog-to-digital converter (ADC) module (continued)

The ADC module in the 240xA has been enhanced to provide flexible interface to event managers A and B. The ADC interface is built around a fast, 10-bit ADC module with total conversion time of 500 ns (S/H + conversion). The ADC module has 16 channels, configurable as two independent 8-channel modules to service event managers A and B. The two independent 8-channel modules can be cascaded to form a 16-channel module. Although there are multiple input channels and two sequencers, there is only one converter in the ADC module. Figure 10 shows the block diagram of the 240xA ADC module.

The two 8-channel modules have the capability to autosequence a series of conversions, each module has the choice of selecting any one of the respective eight channels available through an analog mux. In the cascaded mode, the autosequencer functions as a single 16-channel sequencer. On each sequencer, once the conversion is complete, the selected channel value is stored in its respective RESULT register. Autosequencing allows the system to convert the same channel multiple times, allowing the user to perform oversampling algorithms. This gives increased resolution over traditional single-sampled conversion results.

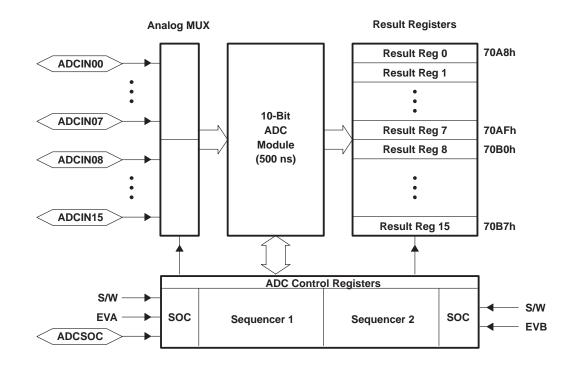


Figure 10. Block Diagram of the 240xA ADC Module

To obtain the specified accuracy of the ADC, proper board layout is very critical. To the best extent possible, traces leading to the ADCINn pins should not run in close proximity to the digital signal paths. This is to minimize switching noise on the digital lines from getting coupled to the ADC inputs. Furthermore, proper isolation techniques must be used to isolate the ADC module power pins (such as  $V_{CCA}$ ,  $V_{REFHI}$ , and  $V_{SSA}$ ) from the digital supply.



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## controller area network (CAN) module

The CAN module is a full-CAN controller designed as a 16-bit peripheral module and supports the following features:

- CAN specification 2.0B (active)
  - Standard data and remote frames
  - Extended data and remote frames
- Six mailboxes for objects of 0- to 8-byte data length
  - Two receive mailboxes, two transmit mailboxes
  - Two configurable transmit/receive mailboxes
- Local acceptance mask registers for mailboxes 0 and 1 and mailboxes 2 and 3
- Configurable standard or extended message identifier
- Programmable bit rate
- Programmable interrupt scheme
- Readable error counters
- Self-test mode
  - In this mode, the CAN module operates in a loop-back fashion, receiving its own transmitted message.

The CAN module is a 16-bit peripheral. The accesses are split into the control/status-registers accesses and the mailbox-RAM accesses.

CAN peripheral registers: The CPU can access the CAN peripheral registers only using 16-bit write accesses. The CAN peripheral always presents full 16-bit data to the CPU bus during read cycles.



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## controller area network (CAN) module (continued)

## CAN controller architecture

Figure 11 shows the basic architecture of the CAN controller through this block diagram of the CAN Peripherals.

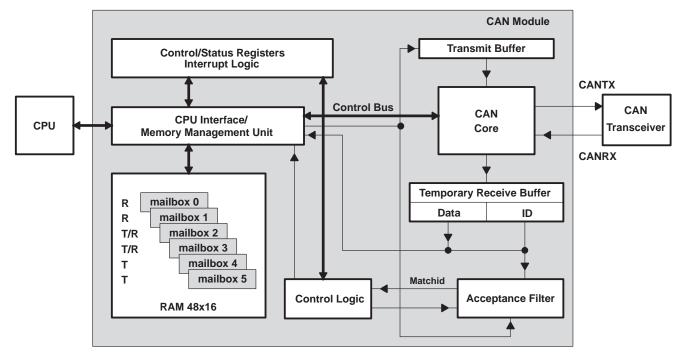


Figure 11. CAN Module Block Diagram

The mailboxes are situated in one 48-word x 16-bit RAM. It can be written to or read by the CPU or the CAN. The CAN write or read access, as well as the CPU read access, needs one clock cycle. The CPU write access needs two clock cycles. In these two clock cycles, the CAN performs a read-modify-write cycle and, therefore, inserts one wait state for the CPU.

Address bit 0 of the address bus used when accessing the RAM decides if the lower (0) or the higher (1) 16-bit word of the 32-bit word is taken. The RAM location is determined by the upper bits 5 to 1 of the address bus.

Table 8 shows the mailbox locations in RAM. One half-word has 16 bits.

## CAN interrupt logic

There are two interrupt requests from the CAN module to the peripheral interrupt expansion (PIE) controller: the mailbox interrupt and the error interrupt. Both interrupts can assert either a high-priority request or a low-priority request to the CPU. Since CAN mailboxes can generate multiple interrupts, the software should read the CAN\_IFR register for every interrupt and prioritize the interrupt service, or else, these multiple interrupts will not be recognized by the CPU and PIE hardware logic. Each interrupt routine should service all the interrupt bits that are set and clear them after service.



# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

## CAN memory map

Table 8 shows the mailbox locations in the CAN module.

ADDRESS OFFSET [5:0]	NAME	DESCRIPTION UPPER HALF-WORD ADDRESS BIT 0 = 1	DESCRIPTION LOWER HALF-WORD ADDRESS BIT 0 = 0			
00h	MSGID0	Message ID for mailbox 0	Message ID for mailbox 0			
02h	MSGCTRL0	Unused	RTR and DLC (bits 4 to 0)			
0.41	Dataland	Databyte 0, Databyte 1 (DBO = 1)	Databyte 2, Databyte 3 (DBO = 1)			
04h	Datalow0	Databyte 3, Databyte 2 (DBO = 0)	Databyte 1, Databyte 0 (DBO = 0)			
0.01	Detablaho	Databyte 4, Databyte 5 (DBO = 1)	Databyte 6, Databyte 7 (DBO = 1)			
06h	Datahigh0	Databyte 7, Databyte 6 (DBO = 0)	Databyte 5, Databyte 4 (DBO = 0)			
08h	MSGID1	Message ID for mailbox 1	Message ID for mailbox 1			
0Ah	MSGCTRL1	Unused	RTR and DLC (bits 4 to 0)			
		Databyte 0, Databyte 1 (DBO = 1)	Databyte 2, Databyte 3 (DBO = 1)			
0Ch	Datalow1	Databyte 3, Databyte 2 (DBO = 0)	Databyte 1, Databyte 0 (DBO = 0)			
0Eh	Datahigh1	Databyte 4, Databyte 5 (DBO = 1)	Databyte 6, Databyte 7 (DBO = 1)			
28h	MSGID5	Message ID for mailbox 5	Message ID for mailbox 5			
2Ah	MSGCTRL5	Unused	RTR and DLC (bits 4 to 0)			
		Databyte 0, Databyte 1 (DBO = 1)	Databyte 2, Databyte 3 (DBO = 1)			
2Ch	Datalow5	Databyte 3, Databyte 2 (DBO = 0)	Databyte 3, Databyte 2 (DBO = 0)			
054	Detabish	Databyte 4, Databyte 5 (DBO = 1)	Databyte 6, Databyte 7 (DBO = 1)			
2Eh	Datahigh5	Databyte 7, Databyte 6 (DBO = 0)	Databyte 5, Databyte 4 (DBO = 0)			

Table 8. Mailbox Addresses<sup>†</sup>

<sup>†</sup> The DBO (data byte order) bit is located in the MCR register and is used to define the order in which the data bytes are stored in the mailbox when received and the order in which the data bytes are transmitted. Byte 0 is the first byte in the message and Byte 7 is the last one shown in the CAN message.



## serial communications interface (SCI) module

The 240xA devices include a serial communications interface (SCI) module. The SCI module supports digital communications between the CPU and other asynchronous peripherals that use the standard non-return-to-zero (NRZ) format. The SCI receiver and transmitter are double-buffered, and each has its own separate enable and interrupt bits. Both can be operated independently or simultaneously in the full-duplex mode. To ensure data integrity, the SCI checks received data for break detection, parity, overrun, and framing errors. The bit rate is programmable to over 65000 different speeds through a 16-bit baud-select register. Features of the SCI module include:

- Two external pins:
  - SCITXD: SCI transmit-output pin
  - SCIRXD: SCI receive-input pin
- NOTE: Both pins can be used as GPIO if not used for SCI.
- Baud rate programmable to 64K different rates
  - Up to 2500 Kbps at 40-MHz CPUCLK
- Data-word format
  - One start bit
  - Data-word length programmable from one to eight bits
  - Optional even/odd/no parity bit
  - One or two stop bits
- Four error-detection flags: parity, overrun, framing, and break detection
- Two wake-up multiprocessor modes: idle-line and address bit
- Half- or full-duplex operation
- Double-buffered receive and transmit functions
- Transmitter and receiver operations can be accomplished through interrupt-driven or polled algorithms with status flags.
  - Transmitter: TXRDY flag (transmitter-buffer register is ready to receive another character) and TX EMPTY flag (transmitter-shift register is empty)
  - Receiver: RXRDY flag (receiver-buffer register is ready to receive another character), BRKDT flag (break condition occurred), and RX ERROR flag (monitoring four interrupt conditions)
- Separate enable bits for transmitter and receiver interrupts (except BRKDT)
- NRZ (non-return-to-zero) format
- Ten SCI module control registers located in the control register frame beginning at address 7050h NOTE: All registers in this module are 8-bit registers that are connected to the 16-bit peripheral bus. When a register is accessed, the register data is in the lower byte (7–0), and the upper byte (15–8) is read as zeros. Writing to the upper byte has no effect.

Figure 12 shows the SCI module block diagram.



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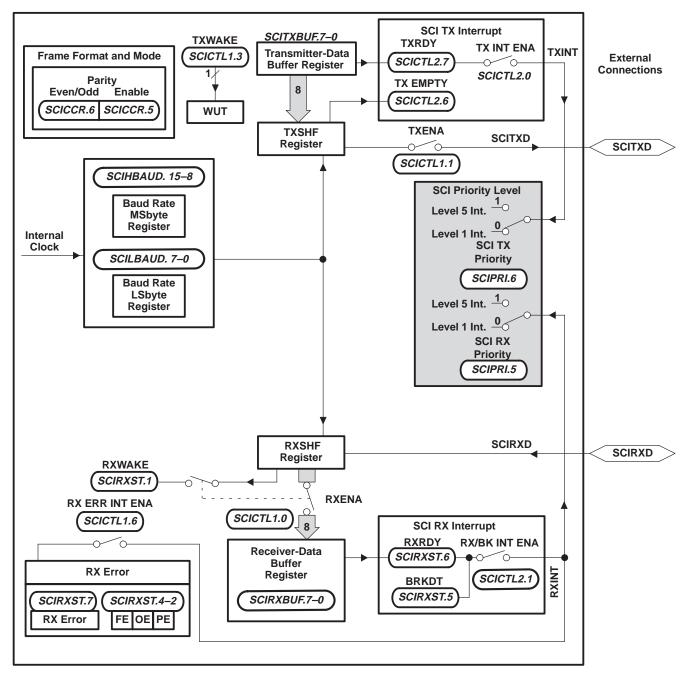


Figure 12. Serial Communications Interface (SCI) Module Block Diagram



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## serial peripheral interface (SPI) module

Some 240xA devices include the four-pin serial peripheral interface (SPI) module. The SPI is a high-speed, synchronous serial I/O port that allows a serial bit stream of programmed length (one to sixteen bits) to be shifted into and out of the device at a programmable bit-transfer rate. Normally, the SPI is used for communications between the DSP controller and external peripherals or another processor. Typical applications include external I/O or peripheral expansion through devices such as shift registers, display drivers, and ADCs. Multidevice communications are supported by the master/slave operation of the SPI.

The SPI module features include:

- Four external pins:
  - SPISOMI: SPI slave-output/master-input pin
  - SPISIMO: SPI slave-input/master-output pin
  - SPISTE: SPI slave transmit-enable pin
  - SPICLK: SPI serial-clock pin

NOTE: All four pins can be used as GPIO, if the SPI module is not used.

- Two operational modes: master and slave
- Baud rate: 125 different programmable rates/10 Mbps at 40-MHz CPUCLK
- Data word length: one to sixteen data bits
- Four clocking schemes (controlled by clock polarity and clock phase bits) include:
  - Falling edge without phase delay: SPICLK active high. SPI transmits data on the falling edge of the SPICLK signal and receives data on the rising edge of the SPICLK signal.
  - Falling edge with phase delay: SPICLK active high. SPI transmits data one half-cycle ahead of the falling edge of the SPICLK signal and receives data on the falling edge of the SPICLK signal.
  - Rising edge without phase delay: SPICLK inactive low. SPI transmits data on the rising edge of the SPICLK signal and receives data on the falling edge of the SPICLK signal.
  - Rising edge with phase delay: SPICLK inactive low. SPI transmits data one half-cycle ahead of the falling edge of the SPICLK signal and receives data on the rising edge of the SPICLK signal.
- Simultaneous receive and transmit operation (transmit function can be disabled in software)
- Transmitter and receiver operations are accomplished through either interrupt-driven or polled algorithms.

• Nine SPI module control registers: Located in control register frame beginning at address 7040h. NOTE: All registers in this module are 16-bit registers that are connected to the 16-bit peripheral bus. When a register is accessed, the register data is in the lower byte (7–0), and the upper byte (15–8) is read as zeros. Writing to the upper byte has no effect.



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# serial peripheral interface (SPI) module (continued)

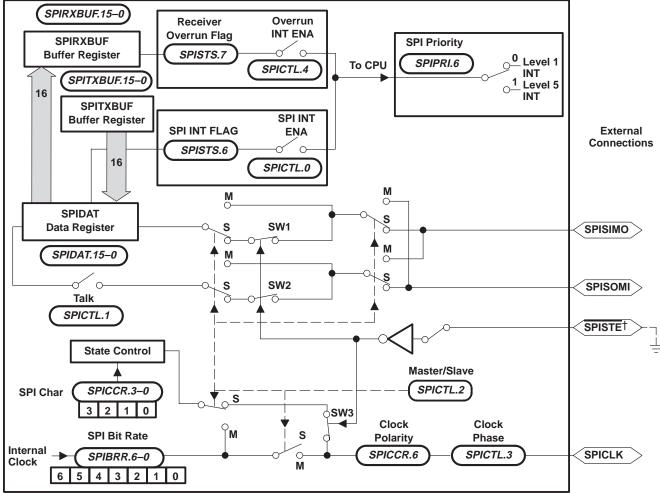


Figure 13 is a block diagram of the SPI in slave mode.

NOTE A: The diagram is shown in the slave mode.

<sup>†</sup> The SPISTE pin is driven low externally. Note that SW1, SW2, and SW3 are closed in this configuration.

Figure 13. Four-Pin Serial Peripheral Interface Module Block Diagram



## PLL-based clock module

The 240xA has an on-chip, PLL-based clock module. This module provides all the necessary clocking signals for the device, as well as control for low-power mode entry. The PLL has a 3-bit ratio control to select different CPU clock rates. See Figure 14 for the PLL Clock Module Block Diagram, Table 9 for clock rates, and Table 10 for the loop filter component values.

The PLL-based clock module provides two modes of operation:

- Crystal-operation
   This mode allows the use of an external crystal/resonator to provide the time base to the device.
- External clock source operation
   This mode allows the internal oscillator to be bypassed. The device clocks are generated from an external clock source input on the XTAL1/CLKIN pin. In this case, an external oscillator clock is connected to the XTAL1/CLKIN pin.

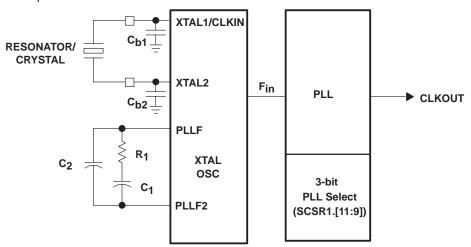


Figure 14. PLL Clock Module Block Diagram

Table 9. PLI	- Clock Selection	Through Bits	(11–9) in	SCSR1 Register
--------------	-------------------	--------------	-----------	----------------

CLK PS2	CLK PS1	CLK PS0	CLKOUT
0	0	0	$4 \times F_{in}$
0	0	1	$2 \times F_{in}$
0	1	0	$1.33  imes F_{in}$
0	1	1	$1 \times F_{in}$
1	0	0	$0.8 \times F_{in}$
1	0	1	$0.66 \times F_{in}$
1	1	0	$0.57  imes F_{in}$
1	1	1	$0.5  imes F_{in}$

Default multiplication factor after reset is (1,1,1), i.e.,  $0.5 \times F_{\mbox{in}}.$ 

#### CAUTION:

The bootloader sets the PLL to x2 or x4 option. If the bootloader is used, the value of CLKIN used should not force CLKOUT to exceed the maximum rated device speed. See the "Boot ROM" section for more details.

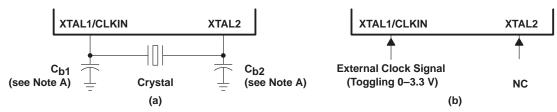


#### external reference crystal clock option

The internal oscillator is enabled by connecting a crystal across the XTAL1/CLKIN and XTAL2 pins as shown in Figure 15a. The crystal should be in fundamental operation and parallel resonant, with an effective series resistance of 30  $\Omega$ –150  $\Omega$  and a power dissipation of 1 mW; it should be specified at a load capacitance of 20 pF.

## external reference oscillator clock option

The internal oscillator is disabled by connecting a clock signal to XTAL1/CLKIN and leaving the XTAL2 input pin unconnected as shown in Figure 15b.



NOTE A: TI recommends that customers have the resonator/crystal vendor characterize the operation of their device with the DSP chip. The resonator/crystal vendor has the equipment and expertise to tune the tank circuit. The vendor can also advise the customer regarding the proper tank component values that will ensure start-up and stability over the entire operating range.

### Figure 15. Recommended Crystal/Clock Connection

### loop filter

The PLL module uses an external loop filter circuit for jitter minimization. The components for the loop filter circuit are R1, C1, and C2. The capacitors (C1 and C2) must be non-polarized. This loop filter circuit is connected between the PLLF and PLLF2 pins (see Figure 14). For examples of component values of R1, C1, and C2 at a specified oscillator frequency (XTAL1), see Table 10.

XTAL1/CLKIN FREQUENCY (MHz)	R1 (Ω) (±5% TOLERANCE)	C1 (μF) (±20% TOLERANCE)	C2 (μF) (±20% TOLERANCE)
4	4.7	3.9	0.082
5	5.6	2.7	0.056
6	6.8	1.8	0.039
7	8.2	1.5	0.033
8	9.1	1	0.022
9	10	0.82	0.015
10	11	0.68	0.015
11	12	0.56	0.012
12	13	0.47	0.01
13	15	0.39	0.0082
14	15	0.33	0.0068
15	16	0.33	0.0068
16	18	0.27	0.0056
17	18	0.22	0.0047
18	20	0.22	0.0047
19	22	0.18	0.0039
20	24	0.15	0.0033

Table 10. Loop Filter Component Values With Damping Factor = 2.0



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#### low-power modes

The 240xA has an IDLE instruction. When executed, the IDLE instruction stops the clocks to all circuits in the CPU, but the clock output from the CPU continues to run. With this instruction, the CPU clocks can be shut down to save power while the peripherals (clocked with CLKOUT) continue to run. The CPU exits the IDLE state if it is reset, or, if it receives an interrupt request.

#### clock domains

All 240xA-based devices have two clock domains:

- 1. CPU clock domain consists of the clock for most of the CPU logic
- 2. System clock domain consists of the peripheral clock (which is derived from CLKOUT of the CPU) and the clock for the interrupt logic in the CPU.

When the CPU goes into IDLE mode, the CPU clock domain is stopped while the system clock domain continues to run. This mode is also known as IDLE1 mode. The 240xA CPU also contains support for a second IDLE mode, IDLE2. By asserting IDLE2 to the 240xA CPU, both the CPU clock domain and the system clock domain are stopped, allowing further power savings. A third low-power mode, HALT mode, the deepest, is possible if the oscillator and WDCLK are also shut down when in IDLE2 mode.

Two control bits, LPM1 and LPM0, specify which of the three possible low-power modes is entered when the IDLE instruction is executed (see Table 11). These bits are located in the System Control and Status Register 1 (SCSR1), and they are described in the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A).

LOW-POWER MODE	LPMx BITS SCSR1 [13:12]	CPU CLOCK DOMAIN	SYSTEM CLOCK DOMAIN	WDCLK STATUS	PLL STATUS	OSC STATUS	FLASH POWER	EXIT CONDITION
CPU running normally	XX	On	On	On	On	On	On	—
IDLE1 – (LPM0)	00	Off	On	On	On	On	On	Peripheral Interrupt, External Interrupt, <u>Reset,</u> PDPINTA/B
IDLE2 – (LPM1)	01	Off	Off	On	On	On	On	Wakeup Interrupts, External Interrupt, Reset, PDPINTA/B
HALT – (LPM2) [PLL/OSC power down]	1X	Off	Off	Off	Off	Off	Off	Reset, PDPINTA/B

Table 11. Low-Power Modes Summary

### other power-down options

240xA devices have clock-enable bits to the following on-chip peripherals: ADC, SCI, SPI, CAN, EVB, and EVA. Clock to these peripherals are disabled after reset; thus, start-up power can be low for the device.

Depending on the application, these peripherals can be turned on/off to achieve low power.

Refer to the SCSR1 register for details on the peripheral clock enable bits.



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## digital I/O and shared pin functions

The 240xA has up to 41 general-purpose, bidirectional, digital I/O (GPIO) pins—most of which are shared between primary functions and I/O. Most I/O pins of the 240xA are shared with other functions. The digital I/O ports module provides a flexible method for controlling both dedicated I/O and shared pin functions. All I/O and shared pin functions are controlled using eight 16-bit registers. These registers are divided into two types:

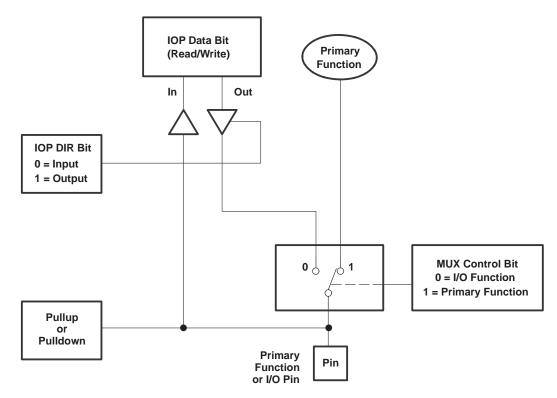
- Output Control Registers used to control the multiplexer selection that chooses between the primary function of a pin or the general-purpose I/O function.
- Data and Control Registers used to control the data and data direction of bidirectional I/O pins.

## description of shared I/O pins

The control structure for shared I/O pins is shown in Figure 16, where each pin has three bits that define its operation:

- Mux control bit this bit selects between the primary function (1) and I/O function (0) of the pin.
- I/O direction bit if the I/O function is selected for the pin (mux control bit is set to 0), this bit determines
  whether the pin is an input (0) or an output (1).
- I/O data bit if the I/O function is selected for the pin (mux control bit is set to 0) and the direction selected is an input, data is read from this bit; if the direction selected is an output, data is written to this bit.

The mux control bit, I/O direction bit, and I/O data bit are in the I/O control registers.



## Figure 16. Shared Pin Configuration

A summary of shared pin configurations and associated bits is shown in Table 12.



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#### description of shared I/O pins (continued)

PIN FUNCTION	SELECTED	MUX	MUX CONTROL	I/O PO	ORT DATA AND DIR	ECTION <sup>‡</sup>
(MCRx.n = 1) Primary Function	(MCRX.N = 0) I/O	CONTROL REGISTER (name.bit #)	VALUE AT RESET (MCRx.n)	REGISTER	DATA BIT NO.§	DIR BIT NO.¶
				PORT A		
SCITXD	IOPA0	MCRA.0	0	PADATDIR	0	8
SCIRXD	IOPA1	MCRA.1	0	PADATDIR	1	9
XINT1	IOPA2	MCRA.2	0	PADATDIR	2	10
CAP1/QEP1	IOPA3	MCRA.3	0	PADATDIR	3	11
CAP2/QEP2	IOPA4	MCRA.4	0	PADATDIR	4	12
CAP3	IOPA5	MCRA.5	0	PADATDIR	5	13
PWM1	IOPA6	MCRA.6	0	PADATDIR	6	14
PWM2	IOPA7	MCRA.7	0	PADATDIR	7	15
				PORT B		
PWM3	IOPB0	MCRA.8	0	PBDATDIR	0	8
PWM4	IOPB1	MCRA.9	0	PBDATDIR	1	9
PWM5	IOPB2	MCRA.10	0	PBDATDIR	2	10
PWM6	IOPB3	MCRA.11	0	PBDATDIR	3	11
T1PWM/T1CMP	IOPB4	MCRA.12	0	PBDATDIR	4	12
T2PWM/T2CMP	IOPB5	MCRA.13	0	PBDATDIR	5	13
TDIRA	IOPB6	MCRA.14	0	PBDATDIR	6	14
TCLKINA	IOPB7	MCRA.15	0	PBDATDIR	7	15
				PORT C		
W/R <sup>#</sup>	IOPC0	MCRB.0	1	PCDATDIR	0	8
BIO	IOPC1	MCRB.1	1	PCDATDIR	1	9
SPISIMO	IOPC2	MCRB.2	0	PCDATDIR	2	10
SPISOMI	IOPC3	MCRB.3	0	PCDATDIR	3	11
SPICLK	IOPC4	MCRB.4	0	PCDATDIR	4	12
SPISTE	IOPC5	MCRB.5	0	PCDATDIR	5	13
CANTX	IOPC6	MCRB.6	0	PCDATDIR	6	14
CANRX	IOPC7	MCRB.7	0	PCDATDIR	7	15
				PORT D		
XINT2/ADCSOC	IOPD0	MCRB.8	0	PDDATDIR	0	8
EMUO	Reserved	MCRB.9	1	PDDATDIR	1	9
EMU1	Reserved	MCRB.10	1	PDDATDIR	2	10
ТСК	Reserved	MCRB.11	1	PDDATDIR	3	11
TDI	Reserved	MCRB.12	1	PDDATDIR	4	12
TDO	Reserved	MCRB.13	1	PDDATDIR	5	13
TMS	Reserved	MCRB.14	1	PDDATDIR	6	14
TMS2	Reserved	MCRB.15	1	PDDATDIR	7	15

#### Table 12. Shared Pin Configurations<sup>†</sup>

<sup>†</sup>Bold, italicized pin names indicate pin functions at reset.

<sup>‡</sup> Valid only if the I/O function is selected on the pin

§ If the GPIO pin is configured as an output, these bits can be written to. If the pin is configured as an input, these bits are read from.

¶ If the DIR bit is 0, the GPIO pin functions as an input. For a value of 1, the pin is configured as an output.

# At reset, all LF240xA devices come up with the W/R/IOPC0 pin in W/R mode. On devices that lack an external memory interface (e.g., LF2406A and LF2402A), W/R mode is not functional and MCRB.0 must be set to a 0 if the IOPC0 pin is to be used.

|| Note that bits 15 through 9 of the MCRB register **must** be written as 1 only. Writing a 0 to any of these bits will cause unpredictable operation of the device.



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## description of shared I/O pins (continued)

 Table 12. Shared Pin Configurations<sup>†</sup> (Continued)

PIN FUNCTION	SELECTED	MUX MUX CONTROL		I/O P(	ORT DATA AND DIR	ECTION <sup>‡</sup>
(MCRx.n = 1) Primary Function	(MCRX.N = 0) I/O	CONTROL REGISTER (name.bit #)	VALUE AT RESET (MCRx.n)	REGISTER	DATA BIT NO.§	DIR BIT NO.¶
				PORT E		
CLKOUT	IOPE0	MCRC.0	1	PEDATDIR	0	8
PWM7	IOPE1	MCRC.1	0	PEDATDIR	1	9
PWM8	IOPE2	MCRC.2	0	PEDATDIR	2	10
PWM9	IOPE3	MCRC.3	0	PEDATDIR	3	11
PWM10	IOPE4	MCRC.4	0	PEDATDIR	4	12
PWM11	IOPE5	MCRC.5	0	PEDATDIR	5	13
PWM12	IOPE6	MCRC.6	0	PEDATDIR	6	14
CAP4/QEP3	IOPE7	MCRC.7	0	PEDATDIR	7	15
				PORT F		
CAP5/QEP4	IOPF0	MCRC.8	0	PFDATDIR	0	8
CAP6	IOPF1	MCRC.9	0	PFDATDIR	1	9
T3PWM/T3CMP	IOPF2	MCRC.10	0	PFDATDIR	2	10
T4PWM/T4CMP	IOPF3	MCRC.11	0	PFDATDIR	3	11
TDIRB	IOPF4	MCRC.12	0	PFDATDIR	4	12
TCLKINB	IOPF5	MCRC.13	0	PFDATDIR	5	13

<sup>†</sup>Bold, italicized pin names indicate pin functions at reset.

<sup>‡</sup> Valid only if the I/O function is selected on the pin

§ If the GPIO pin is configured as an output, these bits can be written to. If the pin is configured as an input, these bits are read from.

If the DIR bit is 0, the GPIO pin functions as an input. For a value of 1, the pin is configured as an output.

<sup>#</sup> At reset, all LF240xA devices come up with the W/R/IOPC0 pin in W/R mode. On devices that lack an external memory interface (e.g., LF2406A and LF2402A), W/R mode is not functional and MCRB.0 must be set to a 0 if the IOPC0 pin is to be used.

|| Note that bits 15 through 9 of the MCRB register **must** be written as 1 only. Writing a 0 to any of these bits will cause unpredictable operation of the device.

#### digital I/O control registers

Table 13 lists the registers available in the digital I/O module. As with other 240xA peripherals, these registers are memory-mapped to the data space.

ADDRESS	DECISTED	NAME
ADDRE55	REGISTER	NAME
7090h	MCRA	I/O mux control register A
7092h	MCRB	I/O mux control register B
7094h	MCRC	I/O mux control register C
7095h	PEDATDIR	I/O port E data and direction register
7096h	PFDATDIR	I/O port F data and direction register
7098h	PADATDIR	I/O port A data and direction register
709Ah	PBDATDIR	I/O port B data and direction register
709Ch	PCDATDIR	I/O port C data and direction register
709Eh	PDDATDIR	I/O port D data and direction register

#### Table 13. Addresses of Digital I/O Control Registers



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## external memory interface (LF2407A)

The TMS320LF2407A can address up to  $64K \times 16$  words of memory (or registers) in each of the program, data, and I/O spaces. On-chip memory, when enabled, occupies some of this off-chip range.

The CPU of the TMS320LF2407A schedules a program fetch, data read, and data write on the same machine cycle. This is because from on-chip memory, the CPU can execute all three of these operations in the same cycle. However, the external interface multiplexes the internal buses to one address bus and one data bus. The external interface sequences these operations to complete first the data write, then the data read, and finally the program read.

The LF2407A supports a wide range of system interfacing requirements. Program, data, and I/O address spaces provide interface to memory and I/O, thereby maximizing system throughput. The full 16-bit address and data buses, along with the PS, DS, and IS space-select signals, allow addressing of 64K 16-bit words in program, data, and I/O space. Since on-chip peripheral registers occupy positions of data-memory space (7000–7FFF), the externally addressable data-memory space is 32K 16-bit words (8000–FFFF). Note that the global memory space of the C2xx core is not used for 240xA DSP devices. Therefore, the global memory allocation register (GREG) is reserved for all these devices.

Input/output (I/O) design is simplified by having I/O space treated the same way as memory. I/O devices are accessed in the I/O address space using the processor's external address and data buses in the same manner as memory-mapped devices.

The LF2407A external parallel interface provides various control signals to facilitate interfacing to the device. The R/W output signal is provided to indicate whether the current cycle is a read or a write. The STRB output signal provides a timing reference for all external cycles. For convenience, the device also provides the RD and the WE output signals, which indicate a read cycle and a write cycle, respectively, along with timing information for those cycles. The availability of these signals minimizes external gating necessary for interfacing external devices to the LF2407A.

The 2407A provides  $\overline{RD}$  and  $W/\overline{R}$  signals to help the zero-wait-state external memory interface. At higher CLKOUT speeds,  $\overline{RD}$  may not meet the slow memory device's timing. In such instances, the  $W/\overline{R}$  signal could be used as an alternative signal with some tradeoffs. See the timings for details.

The TMS320LF2407A supports zero-wait-state reads on the external interface. However, to avoid bus conflicts, writes take two cycles. This allows the TMS320LF2407A to buffer the transition of the data bus from input to output (or from output to input) by a half cycle. In most systems, the TMS320LF2407A ratio of reads to writes is significantly large to minimize the overhead of the extra cycle on writes.

### wait-state generation (LF2407A only)

Wait-state generation is incorporated in the LF2407A without any external hardware for interfacing the LF2407A with slower off-chip memory and I/O devices. Adding wait states lengthens the time the CPU waits for external memory or an external I/O port to respond when the CPU reads from or writes to that external memory or I/O port. Specifically, the CPU waits one extra cycle (one CLKOUT cycle) for every wait state. The wait states operate on CLKOUT cycle boundaries.

To avoid bus conflicts, writes from the LF2407A always take at least two CLKOUT cycles. The LF2407A offers two options for generating wait states:

- READY Signal. With the READY signal, you can externally generate any number of wait states. The READY pin has no effect on accesses to *internal* memory.
- On-Chip Wait-State Generator. With this generator, you can generate zero to seven wait states.



#### generating wait states with the READY signal

When the READY signal is low, the LF2407A waits one CLKOUT cycle and then checks READY again. The LF2407A does not continue executing until the READY signal is driven high; therefore, if the READY signal is not used, it should be pulled high.

The READY pin can be used to generate any number of wait states. However, when the LF2407A operates at full speed, it may not respond fast enough to provide a READY-based wait state for the first cycle. For extended wait states using external READY logic, the on-chip wait-state generator should be programmed to generate at least one wait state.

#### generating wait states with the LF2407A on-chip software wait-state generator

The software wait-state generator can be programmed to generate zero to seven wait states for a given off-chip memory space (program, data, or I/O), regardless of the state of the READY signal. These zero to seven wait states are controlled by the wait-state generator register (WSGR) (I/O FFFFh). For more detailed information on the WSGR and associated bit functions, refer to the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A).

#### watchdog (WD) timer module

The x240xA devices include a watchdog (WD) timer module. The WD function of this module monitors software and hardware operation by generating a system reset if it is not periodically serviced by software by having the correct key written. The WD timer operates independently of the CPU. It does not need any CPU initialization to function. When a system reset occurs, the WD timer defaults to the fastest WD timer rate available (WDCLK signal = CLKOUT/512). As soon as reset is released internally, the CPU starts executing code, and the WD timer begins incrementing. This means that, to avoid a premature reset, WD setup should occur early in the power-up sequence. See Figure 17 for a block diagram of the WD module. The WD module features include the following:

- WD Timer
  - Seven different WD overflow rates
  - A WD-reset key (WDKEY) register that clears the WD counter when a correct value is written, and generates a system reset if an incorrect value is written to the register
  - WD check bits that initiate a system reset if an incorrect value is written to the WD control register (WDCR)
- Automatic activation of the WD timer, once system reset is released
  - Three WD control registers located in control register frame beginning at address 7020h.
- NOTE: All registers in this module are 8-bit registers. When a register is accessed, the register data is in the lower byte, the upper byte is read as zeros. Writing to the upper byte has no effect.

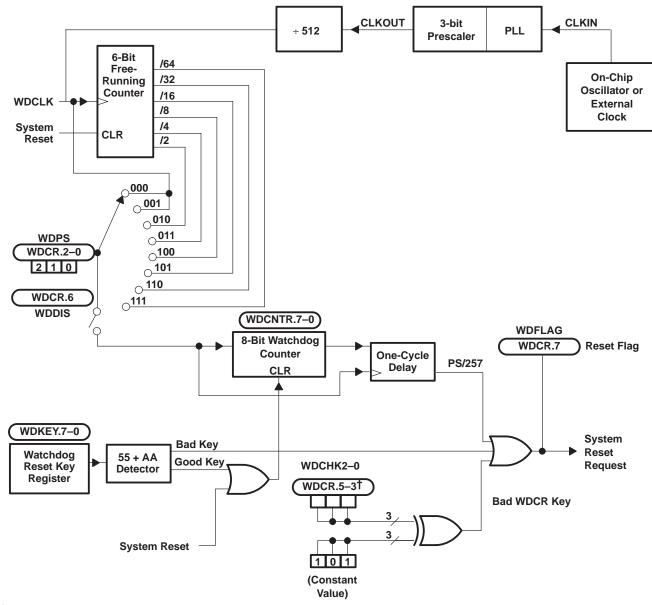
Figure 17 shows the WD block diagram. Table 14 shows the different WD overflow (time-out) selections.

The watchdog can be disabled in software by writing '1' to bit 6 of the WDCR register (WDCR.6) while bit 5 of the SCSR2 register (SCSR2.5) is 1. If SCSR2.5 is 0, the watchdog will not be disabled. SCSR2.5 is equivalent to the WDDIS pin of the TMS320F243/241 devices.



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## watchdog (WD) timer module (continued)



<sup>†</sup> Writing to bits WDCR.5–3 with anything but the correct pattern (101) generates a system reset.

Figure 17. Block Diagram of the WD Module



# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

# watchdog (WD) timer module (continued)

#### Table 14. WD Overflow (Time-out) Selections

v	WD PRESCALE SELECT BITS		WDCLK DIVIDER	WATCHDOG CLOCK RATE <sup>†</sup>
WDPS2	WDPS1	WDPS0		FREQUENCY (Hz)
0	0	X‡	1	WDCLK/1
0	1	0	2	WDCLK/2
0	1	1	4	WDCLK/4
1	0	0	8	WDCLK/8
1	0	1	16	WDCLK/16
1	1	0	32	WDCLK/32
1	1	1	64	WDCLK/64

<sup>†</sup>WDCLK = CLKOUT/512

‡X = Don't care



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#### development support

Texas Instruments (TI) offers an extensive line of development tools for the x240xA generation of DSPs, including tools to evaluate the performance of the processors, generate code, develop algorithm implementations, and fully integrate and debug software and hardware modules.

The following products support development of x240xA-based applications:

### Software Development Tools:

Assembler/linker Simulator Optimizing ANSI C compiler Application algorithms C/Assembly debugger and code profiler

#### Hardware Development Tools:

Emulator XDS510<sup>™</sup> (supports x24x multiprocessor system debug) TMS320LF2407 EVM (Evaluation module for 2407 DSP)

The *TMS320 DSP Development Support Reference Guide* (literature number SPRU011) contains information about development support products for all TMS320<sup>™</sup> DSP family member devices, including documentation. Refer to this document for further information about TMS320<sup>™</sup> DSP documentation or any other TMS320<sup>™</sup> DSP support products from Texas Instruments. There is also an additional document, the *TMS320 Third-Party Support Reference Guide* (literature number SPRU052), which contains information from other companies in the industry regarding products related to the TMS320<sup>™</sup> DSPs. To receive copies of TMS320<sup>™</sup> DSP literature, contact the Literature Response Center at 800-477-8924.

See Table 15 and Table 16 for complete listings of development support tools for the x240xA. For information on pricing and availability, contact the nearest TI field sales office or authorized distributor.

DEVELOPMENT TOOL	PLATFORM	PART NUMBER			
Software – Code Generation Tools					
Assembler/Linker	PC <sup>™</sup> , OS/2 <sup>™</sup> , Windows 3.x/Windows <sup>™</sup> 95	TMDS3242850-02			
C Compiler/Assembler/Linker	PC, OS/2, Windows 3.x/Windows 95	TMDS3242855-02			
C Compiler/Assembler/Linker	Open Windows, HP9000, SPARC™	TMDS3242555-08			
	Software – Simulation				
C2xx Simulator	SPARC, Open Windows	TMDX324x551-09			
	Software – Emulation Debug Tools				
Code Composer 4.10, Code Generation 7.0	PC, Windows 3.x, OS/2	TMDS324012xx			
TI C Source Debugger – WS	SPARC, SunOS™	TMDX32406A2xx			
I	Hardware – Emulation Debug Tools				
XDS510XL™ Board (ISA card), w/JTAG cable	PC	TMDS00510			
XDS510PP™ Pod (Parallel Port) w/JTAG cable	PC	TMDS00510PP			
XDS510WS™ Box w/JTAG cable	SPARC	TMDS00510WS			

#### Table 15. Development Support Tools

SPARC is a trademark of SPARC International, Inc. PC and OS/2 are trademarks of International Business Machines Corp. Windows is a registered trademark of Microsoft Corporation.

SunOS is a trademark of Sun Microsystems, Inc.

XDS510XL, XDS510PP, and XDS510WS are trademarks of Texas Instruments.



## development support (continued)

#### Table 16. TMS320x24x-Specific Development Tools

DEVELOPMENT TOOL	PLATFORM	PART NUMBER
	Hardware – Evaluation/Starter Kits	
TMS320LF2407A EVM	PC, Windows 95, Windows™ 98	TMDX3P701016
TMS320F240 EVM	PC, Windows 3.x	TMDX326P124X
TMS320F243 EVM	PC, Windows 95	TMDS3P604030

The LF2407 Evaluation Module (EVM) provide designers of motor and motion control applications with a complete and cost-effective way to take their designs from concept to production. These tools offer both a hardware and software development environment and include:

- Flash-based LF240xA evaluation board
- Code Generation Tools
- Assembler/Linker
- C Compiler
- Source code debugger
- C24x<sup>™</sup> Debugger
- Code Composer IDE
- XDS510PP™ JTAG-based emulator
- Sample applications code
- Universal 5-V DC power supply
- Documentation and cables

#### device and development support tool nomenclature

To designate the stages in the product development cycle, Texas Instruments assigns prefixes to the part numbers of all TMS320<sup>™</sup> DSP devices and support tools. Each TMS320<sup>™</sup> DSP member has one of three prefixes: TMX, TMP, or TMS. Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMX/TMDX) through fully qualified production devices/tools (TMS/TMDS). This development flow is defined below.

Support tool development evolutionary flow:

TMDX Development support product that has not completed TI's internal qualification testing

**TMDS** Fully qualified development support product

TMX and TMP devices and TMDX development support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

TMS devices and TMDS development support tools have been fully characterized, and the quality and reliability of the device have been fully demonstrated. TI's standard warranty applies.

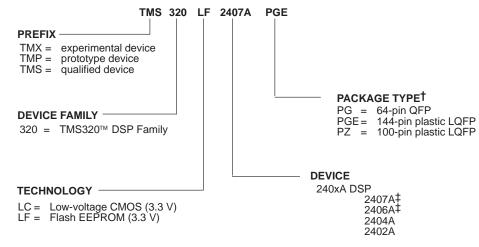


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#### device and development support tool nomenclature (continued)

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, PG, PGE, and PZ) and temperature range (for example, S). Figure 18 and Figure 19 provide a legend for reading the complete device name for any TMS320x240xA family member. Refer to the timing section for specific options that are available on 240xA devices.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

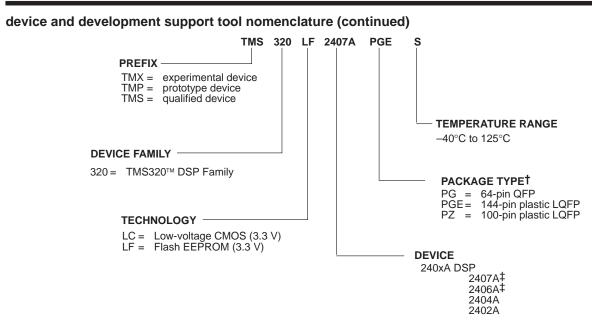


†QFP = Quad Flatpack

- LQFP = Low-Profile Quad Flatpack
- <sup>‡</sup> The package dimensions of the 2407A and 2406A devices correspond to the LQFP package. These devices were stated to be in TQFP packaging in the TMX data sheets. The package dimensions have *not* changed; only the package designation has changed.

Figure 18. TMS320x240xA Device Nomenclature (Industrial Temperature Range: -40°C to 85°C)

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†QFP = Quad Flatpack

LQFP = Low-Profile Quad Flatpack

<sup>‡</sup> The package dimensions of the 2407A and 2406A devices correspond to the LQFP package. These devices were stated to be in TQFP packaging in the TMX data sheets. The package dimensions have *not* changed; only the package designation has changed.

Figure 19. TMS320x240xA Device Nomenclature (Extended Temperature Range: -40°C to 125°C)



#### documentation support

Extensive documentation supports all of the TMS320<sup>™</sup> DSP family generations of devices from product announcement through applications development. The types of documentation available include: data sheets, such as this document, with design specifications; complete user's guides for all devices and development support tools; and hardware and software applications. Useful reference documentation includes:

- User Guides
  - TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals (literature number SPRU357A)
  - TMS320C240 DSP Controllers CPU, System, and Instruction Set Reference Guide (literature number SPRU160)
- Data Sheets
  - TMS320LF2407, TMS320LF2406, TMS320LF2402 DSP Controllers (literature number SPRS094)
  - TMS320LF2407A, TMS320LF2406A, TMS320LF2402A, TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP Controllers (literature number SPRS145)
- Application Reports
  - 3.3V DSP for Digital Motor Control (literature number SPRA550)

A series of DSP textbooks is published by Prentice-Hall and John Wiley & Sons to support digital signal processing research and education. The TMS320<sup>TM</sup> DSP newsletter, *Details on Signal Processing*, is published quarterly and distributed to update TMS320<sup>TM</sup> DSP customers on product information.

Updated information on the TMS320<sup>™</sup> DSP controllers can be found on the worldwide web at: http://www.ti.com.

To send comments regarding the 240xA data sheet (SPRS145), use the *comments@books.sc.ti.com* email address, which is a repository for feedback. For questions and support, contact the Product Information Center listed at the http://www.ti.com/sc/docs/pic/home.htm site.

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absolute maximum ratings over operating free-air temperature ranges (unless otherwise noted) <sup>†</sup>
Supply voltage range, V <sub>DD</sub> , PLLV <sub>CCA</sub> , V <sub>DDO</sub> , and V <sub>CCA</sub> (see Note 1)
V <sub>CCP</sub> range
Input voltage range, V <sub>IN</sub> – 0.3 V to 4.6 V
Output voltage range, V <sub>O</sub> LF240xA – 0.3 V to 4.6 V
Output voltage range, V <sub>O</sub> LC240xA – 0.3 V to 4.6 V
Input clamp current, I <sub>IK</sub> (V <sub>IN</sub> < 0 or V <sub>IN</sub> > V <sub>CC</sub> )
Output clamp current, I <sub>OK</sub> (V <sub>O</sub> < 0 or V <sub>O</sub> > V <sub>CC</sub> )
Operating free-air temperature ranges, T <sub>A</sub> : A version
S version – 40°C to 125°C
Junction temperature range, T <sub>J</sub> – 40°C to 150°C
Storage temperature range, T <sub>stg</sub> – 65°C to 150°C

<sup>†</sup> Clamp current stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. NOTE 1: All voltage values are with respect to V<sub>SS</sub>.

### recommended operating conditions<sup>‡</sup>

				MIN	NOM	MAX	UNIT
VDD/VDDO	Supply voltage		V <sub>DDO</sub> = V <sub>DD</sub> ± 0.3 V	3	3.3	3.6	V
V <sub>SS</sub>	Supply ground			0	0	0	V
PLLV <sub>CCA</sub>	PLL supply voltage			3	3.3	3.6	V
VCCA§	ADC supply voltage			3	3.3	3.6	V
VCCP	Flash programming supply voltage			4.75	5	5.25	V
<b>fCLKOUT</b>	Device clock frequency (system clock)			4		40	MHz
VIH	High-level input voltage	High-level input voltage		2			V
VIL	_ow-level input voltage		All inputs			0.8	V
			Output pins Group 1			0.8 - 2	mA
IOH	High-level output source current, $V_{OH}$ :	= 2.4 V	Output pins Group 2¶			- 4	mA
			Output pins Group 3			- 8	mA
			Output pins Group 1			2	mA
IOL	Low-level output sink current, $V_{OL} = V_{OL}$	OL MAX	Output pins Group 2¶			4	mA
			Output pins Group 3			8	mA
-	Free states and the	A version		- 40		85	<u></u>
Τ <sub>Α</sub>	Free-air temperature	S version		- 40		125	°C
TJ	Junction temperature			- 40	25	150	°C
N <sub>f</sub>	Flash endurance for the array (Write/era	ase cycles)	- 40°C to 85°C		10K		cycles

<sup>‡</sup>Refer to the mechanical data package page for thermal resistance values,  $\Theta_{JA}$  (junction-to-ambient) and  $\Theta_{JC}$  (junction-to-case). § V<sub>CCA</sub> should not exceed V<sub>DD</sub> by 0.3 V.

Primary signals and their groupings:

Group 1: PWM1-PWM6, CAP1-CAP6, TCLKINA, RS, IOPF6, IOPC1, TCK, TDI, TMS, XF, A0-A15

Group 2: PS/DS/IS, RD, W/R, STRB, R/W, VIS\_OE, D0–D15, T1PWM–T4PWM, PWM7–PWM12, CANTX, CANRX, SPICLK, SPISOMI, SPISIMO, SPISTE, EMU0, EMU1, TDO, TMS2

Group 3: TDIRA, TDIRB, SCIRXD, SCITXD, XINT1, XINT2, CLKOUT



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# electrical characteristics over recommended operating free-air temperature ranges (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	V <sub>OH</sub> High-level output voltage		$V_{DD}$ = 3.0 V, $I_{OH}$ = $I_{OH}MAX$	2.4			V
⊻он			All outputs at 50 µA	V <sub>DDO</sub> - 0.2			v
VOL	Low-level output voltage		$I_{OL} = I_{OL}MAX$			0.4	V
	I <sub>IL</sub> Input current (low level) With pullup V <sub>DI</sub>		-9	-9 -16 -25	•		
١L		With pulldown	V <sub>DD</sub> = 3.3 V, V <sub>IN</sub> = 0 V			±2	μA
	Input ourreast (high loval)	With pullup				<u>±2</u>	A
ΊН	Input current (high level)	With pulldown	V <sub>DD</sub> = 3.3 V, V <sub>IN</sub> = V <sub>DD</sub>	9	16	25	μΑ
IOZ	Output current, high-impedance	state (off-state)	$V_{O} = V_{DD} \text{ or } 0 V$			±2	μA
Ci	Input capacitance				2		pF
Co	Output capacitance				3		pF

# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LF2407A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		85	120	mA
ICCA	ADC module current			5	15	mA

<sup>†</sup> IDD is the current flowing into the VDD, VDDO, and PLLVCCA pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LF2406A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		85	120	mA
ICCA	ADC module current			5	15	mA

<sup>†</sup> IDD is the current flowing into the VDD, VDDO, and PLLVCCA pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LF2402A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		75	110	mA
ICCA	ADC module current			5	15	mA

<sup>†</sup>  $I_{DD}$  is the current flowing into the  $V_{DD}$ ,  $V_{DDO}$ , and  $PLLV_{CCA}$  pins.



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# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LC2406A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		55	65	mA
ICCA	ADC module current			5	15	mA

<sup> $\dagger$ </sup> I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LC2404A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		55	65	mA
ICCA	ADC module current			5	15	mA

 $^{\dagger}$  I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges at 40-MHz CLOCKOUT (TMS320LC2402A)<sup>†</sup>

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	Clock to all peripherals is enabled. CPU is running a simple loop code and no I/O pins are switching.		40	50	mA
ICCA	ADC module current			5	15	mA

 $^{\dagger}$  I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.



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# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LF2407A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current		Clock to all peripherals is enabled.		72	96	mA
ICCA	ADC module current	LPM0	No I/O pins are switching.		0.2	1	mA
IDD	Operational Current	LPM1	Clock to all peripherals is disabled. No I/O pins are switching.		48	72	mA
ICCA	ADC module current				0	0	mA
I <sub>DD</sub>	Operational Current	LPM2	Clock to all peripherals is disabled.		200		μΑ
ICCA	ADC module current				0	0	mA

<sup> $\dagger$ </sup> I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LF2406A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	LPM0	Clock to all peripherals is enabled. No I/O pins are switching.		72	96	mA
ICCA	ADC module current				0.2	1	mA
IDD	Operational Current		Clock to all peripherals is disabled.		48	72	mA
ICCA	ADC module current	LPM1	No I/O pins are switching.		0	0	mA
IDD	Operational Current	LPM2	Clock to all peripherals is disabled.		200		μΑ
ICCA	ADC module current				0	0	mA

<sup>†</sup>  $I_{DD}$  is the current flowing into the  $V_{DD}$ ,  $V_{DDO}$ , and  $PLLV_{CCA}$  pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LF2402A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>DD</sub>	Operational Current		Clock to all peripherals is enabled.		60	84	mA
ICCA	ADC module current	LPM0	No I/O pins are switching.		0.2	1	mA
IDD	Operational Current		Clock to all peripherals is disabled.		48	72	mA
ICCA	ADC module current	LPM1	No I/O pins are switching.		0	0	mA
IDD	Operational Current	LPM2	Clock to all peripherals is disabled.		200		μΑ
ICCA	ADC module current				0	0	mA

 $^{\dagger}$  I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.



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# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LC2406A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	LPM0	Clock to all peripherals is enabled. No I/O pins are switching.		30	55	mA
ICCA	ADC module current				0.2	1	mA
IDD	Operational Current	LPM1	Clock to all peripherals is disabled. No I/O pins are switching.		20	42	mA
ICCA	ADC module current				0	0	mA
IDD	Operational Current	LPM2	Clock to all peripherals is disabled.		100		μΑ
ICCA	ADC module current				0	0	mA

 $^{\dagger}$  I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LC2404A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	LPM0	Clock to all peripherals is enabled. No I/O pins are switching.		30	55	mA
ICCA	ADC module current	LPIVIO			0.2	1	mA
I <sub>DD</sub>	Operational Current		Clock to all peripherals is disabled. No I/O pins are switching.		20	42	mA
ICCA	ADC module current	LPM1			0	0	mA
IDD	Operational Current	LPM2	Clock to all peripherals is disabled.		100		μΑ
ICCA	ADC module current				0	0	mA

 $^{\dagger}$  IDD is the current flowing into the VDD, VDDO, and PLLV<sub>CCA</sub> pins.

# current consumption by power-supply pins over recommended operating free-air temperature ranges during low-power modes at 40-MHz CLOCKOUT (TMS320LC2402A)<sup>†</sup>

	PARAMETER	MODE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
IDD	Operational Current	LPM0	Clock to all peripherals is enabled. No I/O pins are switching.		25	50	mA
ICCA	ADC module current				0.2	1	mA
IDD	Operational Current	LPM1	Clock to all peripherals is disabled. No I/O pins are switching.		17	40	mA
ICCA	ADC module current				0	0	mA
IDD	Operational Current	LPM2	Clock to all peripherals is disabled.		100		μΑ
ICCA	ADC module current				0	0	mA

 $^{\dagger}$  I<sub>DD</sub> is the current flowing into the V<sub>DD</sub>, V<sub>DDO</sub>, and PLLV<sub>CCA</sub> pins.



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current consumption graphs

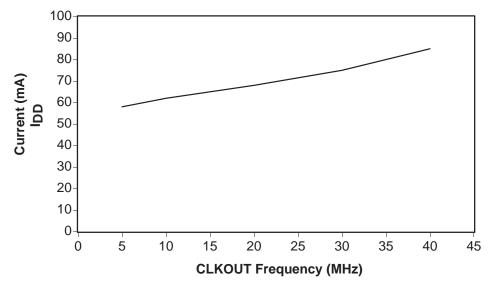


Figure 20. LF2407A Typical Current Consumption (With Peripheral Clocks Enabled)

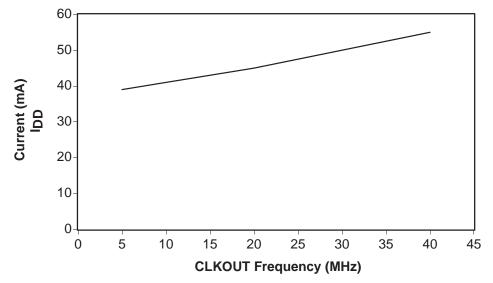


Figure 21. LC2406A Typical Current Consumption (With Peripheral Clocks Enabled)



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### reducing current consumption

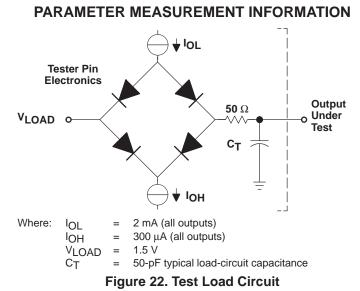
240x DSPs incorporate a unique method to reduce the device current consumption. A reduction in current consumption can be achieved by turning off the clock to any peripheral module which is not used in a given application. Table 17 indicates the typical reduction in current consumption achieved by turning off the clocks to various peripherals. Refer to the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A) for further information on how to turn off the clock to the peripherals.

Table 17. Typical Current Consumption by	Various Peripherals (at 40 MHz)
--	---------------------------------

PERIPHERAL MODULE	CURRENT REDUCTION (mA)
T EKII MEKAE MODOLE	
CAN	8.4
EVA	6.1
EVB	6.1
ADC <sup>†</sup>	2.8†
SCI	1.9
SPI	1.3

<sup>+</sup>ADC current shown is at 30 MHz.

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### signal transition levels

The data in this section is shown for the 3.3-V version. Note that some of the signals use different reference voltages, see the recommended operating conditions table. Output levels are driven to a minimum logic-high level of 2.4 V and to a maximum logic-low level of 0.8 V.

Figure 23 shows output levels.

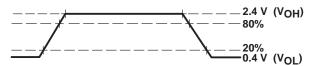


Figure 23. Output Levels

Output transition times are specified as follows:

- For a *high-to-low transition*, the level at which the output is said to be no longer high is below 80% of the total voltage range and lower and the level at which the output is said to be low is 20% of the total voltage range and lower.
- For a *low-to-high transition*, the level at which the output is said to be no longer low is 20% of the total voltage range and higher and the level at which the output is said to be high is 80% of the total voltage range and higher.

Figure 24 shows the input levels.

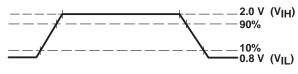


Figure 24. Input Levels

Input transition times are specified as follows:

- For a *high-to-low transition* on an input signal, the level at which the input is said to be no longer high is 90% of the total voltage range and lower and the level at which the input is said to be low is 10% of the total voltage range and lower.
- For a *low-to-high transition* on an input signal, the level at which the input is said to be no longer low is 10% of the total voltage range and higher and the level at which the input is said to be high is 90% of the total voltage range and higher.



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### PARAMETER MEASUREMENT INFORMATION

### timing parameter symbology

Timing parameter symbols used are created in accordance with JEDEC Standard 100. To shorten the symbols, some of the pin names and other related terminology have been abbreviated as follows:

А	A[15:0]	MS	Memory strobe pins $\overline{IS}$ , $\overline{DS}$ , or $\overline{PS}$
CI	XTAL1/CLKIN	R	READY
CO	CLKOUT	RD	Read cycle or RD
D	D[15:0]	RS	RESET pin RS
INT	XINT1, XINT2	W	Write cycle or WE

Lowercase subscripts and their meanings:

а	access time
С	cycle time (period)
d	delay time
f	fall time
h	hold time
r	rise time
su	setup time
t	transition time
v	valid time
W	pulse duration (width)

Letters and symbols and their meanings:

н	High
L	Low
V	Valid
Х	Unknown, changing, or don't care level
Z	High impedance

### general notes on timing parameters

All output signals from the 240xA devices (including CLKOUT) are derived from an internal clock such that all output transitions for a given half-cycle occur with a minimum of skewing relative to each other.

The signal combinations shown in the following timing diagrams may not necessarily represent actual cycles. For actual cycle examples, refer to the appropriate cycle description section of this data sheet.



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### external reference crystal/clock with PLL circuit enabled

### timings with the PLL circuit enabled

PARAMETER		MIN	MAX	UNIT	
		Resonator	4	13	
f <sub>X</sub> Input clock frequency <sup>†</sup>	Crystal	4	20	MHz	
	CLKIN	4	20		

<sup>†</sup> Input frequency should be adjusted (CLK PS bits in SCSR1 register) such that CLKOUT = 40 MHz maximum, 4 MHz minimum.

### switching characteristics over recommended operating conditions [H = 0.5 t<sub>c(CO)</sub>] (see Figure 25)

				· · ·		
	PARAMETER	PLL MODE	MIN	TYP	MAX	UNIT
<sup>t</sup> c(CO)	Cycle time, CLKOUT	×4 mode <sup>†</sup>	33			ns
<sup>t</sup> f(CO)	Fall time, CLKOUT			4		ns
t <sub>r(CO)</sub>	Rise time, CLKOUT			4		ns
<sup>t</sup> w(COL)	Pulse duration, CLKOUT low		H–3	Н	H+3	ns
<sup>t</sup> w(COH)	Pulse duration, CLKOUT high		H–3	Н	H+3	ns
tt	Transition time, PLL synchronized after $\overline{RS}$ pin high				4096t <sub>C(CI)</sub>	ns

<sup>†</sup> Input frequency should be adjusted (CLK PS bits in SCSR1 register) such that CLKOUT = 40 MHz maximum, 4 MHz minimum.

### timing requirements (see Figure 25)

		MIN	MAX	UNIT
<sup>t</sup> c(CI)	Cycle time, XTAL1/CLKIN		250	ns
<sup>t</sup> f(CI)	Fall time, XTAL1/CLKIN		5	ns
<sup>t</sup> r(Cl)	Rise time, XTAL1/CLKIN		5	ns
<sup>t</sup> w(CIL)	Pulse duration, XTAL1/CLKIN low as a percentage of t <sub>C(CI)</sub>	40	60	%
<sup>t</sup> w(CIH)	Pulse duration, XTAL1/CLKIN high as a percentage of $t_{C(CI)}$	40	60	%

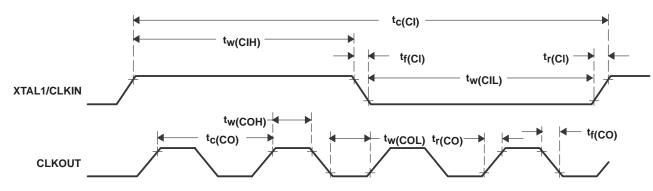


Figure 25. CLKIN-to-CLKOUT Timing with PLL and External Clock in  $\times$ 4 Mode



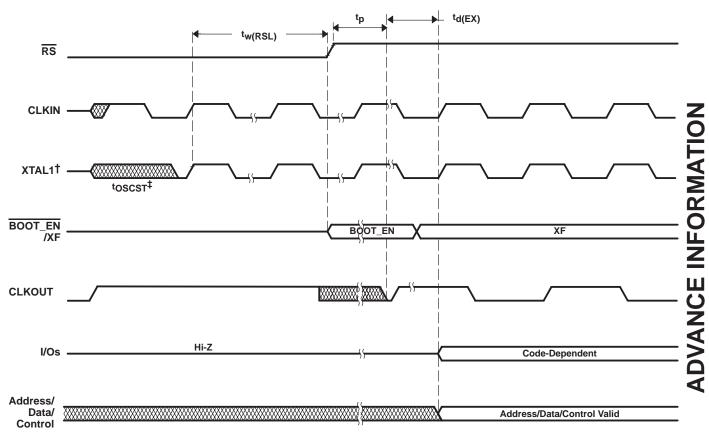
### TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

### **RS** timings

### timing requirements for a reset [H = $0.5t_{c(CO)}$ ] (see Figure 26 and Figure 27)

		MIN	NOM MAX	UNIT
<sup>t</sup> w(RSL)	Pulse duration, stable CLKIN to RS high	<sup>8t</sup> c(CI)		cycles
<sup>t</sup> w(RSL2)	Pulse duration, RS low	<sup>8t</sup> c(CI)		cycles
tp	PLL lock-up time		4096t <sub>c(CI)</sub>	cycles
<sup>t</sup> d(EX)	Delay time, reset vector executed after PLL lock time		36H	ns

V<sub>DD</sub>/V<sub>DDO</sub>



<sup>†</sup> XTAL1 refers to internal oscillator clock if on-chip oscillator is used.

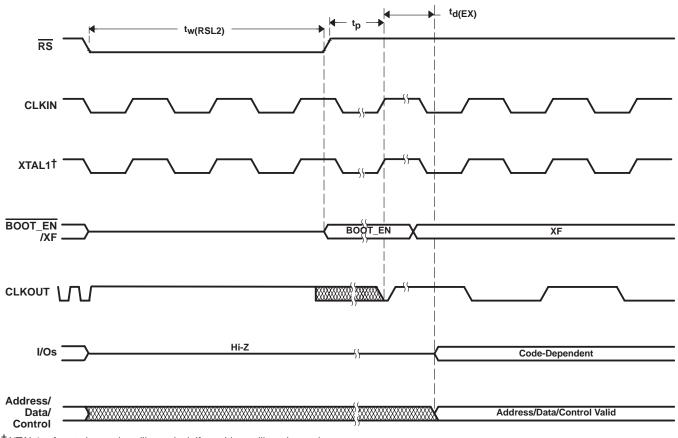
 $\pm$  toscst is the oscillator start-up time, which is dependent on crystal/resonator and board design.

### Figure 26. Power-on Reset



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### **RS** timings (continued)



<sup>†</sup> XTAL1 refers to internal oscillator clock if on-chip oscillator is used.

Figure 27. Warm Reset

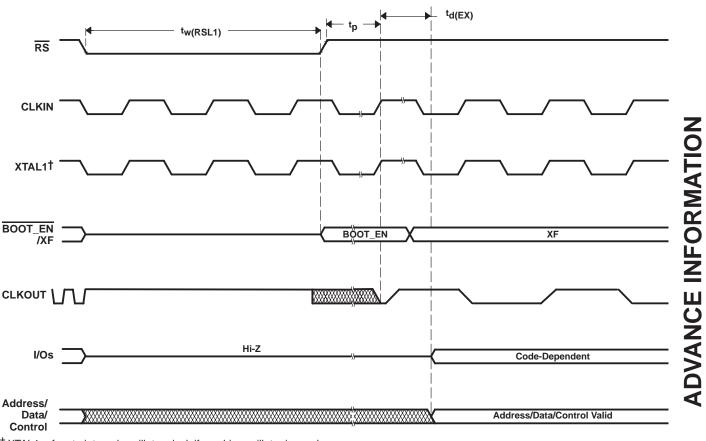


### **RS** timings (continued)

# switching characteristics over recommended operating conditions for a reset $[H = 0.5t_{c(CO)}]$ (see Figure 28)

	PARAMETER	MIN	MAX	UNIT
<sup>t</sup> w(RSL1)	Pulse duration, RS low <sup>†</sup>	128t <sub>c(CI)</sub>		ns
<sup>t</sup> d(EX)	Delay time, reset vector executed after PLL lock time	36H		ns
tp	PLL lock time (input cycles)	4096	<sup>6t</sup> c(CI)	ns

<sup>†</sup> The parameter  $t_{W(RSL1)}$  refers to the time  $\overline{RS}$  is an output.



<sup>†</sup> XTAL1 refers to internal oscillator clock if on-chip oscillator is used.





### TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLÉRS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

### low-power mode timings

# switching characteristics over recommended operating conditions $[H = 0.5t_{c(CO)}]$ (see Figure 29, Figure 30, and Figure 31)

	PARAMETER	LOW-POWER MOD	ES	MIN TYP	MAX	UNIT
	Delay time, CLKOUT switching to	IDLE1	LPM0	$12 \times t_{c(CO)}$		
d(WAKE-A)	program execution resume	IDLE2	LPM1	$15 \times t_{C(CO)}$		ns
d(IDLE-COH)	Delay time, Idle instruction executed to CLKOUT high	IDLE2	LPM1	<sup>4t</sup> c(CO)		ns
d(WAKE-OSC)	Delay time, wakeup interrupt asserted to oscillator running	HALT	LPM2	OSC start-up and PLL lock time		ms
<sup>t</sup> d(IDLE-OSC)	Delay time, Idle instruction executed to oscillator power off	{PLL/OSC power down}		<sup>4t</sup> c(CO)		ns
<sup>t</sup> d(EX)	Delay time, reset vector executed	after RS high		36H		ns
WAKE INT	······			<u> </u>		
WAKE INT can be	e any valid interrupt or RESET.					
		LE1 Entry and Exit 1	Timing	– LPM0		
A0-A15	Figure 29. ID	-	[iming	- LPM0 		×
A0–A15 CLKOUT	Figure 29. ID	-	Fiming	- LPM0		$\sim$
CLKOUT WAKE INT <sup>†</sup>	Figure 29. ID		t <sub>c</sub>	;,;,/ ,/,/,/// i(wake-a)		× ~~
CLKOUT WAKE INT <sup>†</sup>	Figure 29. ID	-	t <sub>c</sub>	; , ; , ; , ; , ; , , , ,		x ~~
CLKOUT WAKE INT <sup>†</sup>	Figure 29. ID	H)		, , , , , , , , , , , , , , , , , , ,		
CLKOUT WAKE INT† WAKE INT can be	Figure 29. ID	H)		; , ; , ; , ; , ; , , , ,		



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### LPM2 wakeup timings

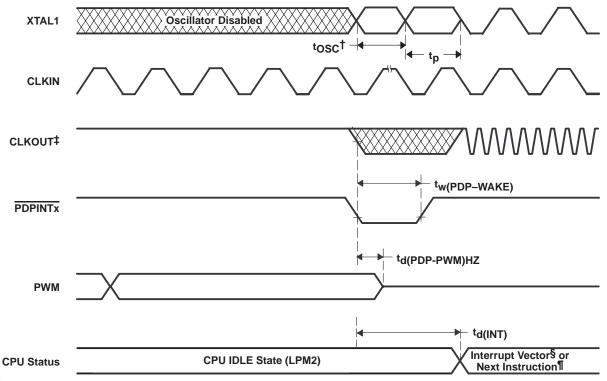
### switching characteristics over recommended operating conditions (see Figure 32)

PARAMETER		MIN	MAX	UNIT
<sup>t</sup> d(PDP-PWM)HZ	Delay time, PDPINTx low to PWM high-impedance state		12	ns
<sup>t</sup> d(INT)	Delay time, INT low/high to interrupt-vector fetch	10t <sub>C(CO)</sub>		ns

### timing requirements (see Figure 32)

		MIN	MAX	UNIT
	if bit 6 of SCSR2 = 0	6t <sub>C</sub> (CO)		
ruise duration, PDPINTX input low	if bit 6 of SCSR2 = 1	12t <sub>c(CO)</sub>		ns
PLL lock-up time		2	<sup>4096t</sup> c(CI)	cycles
	Pulse duration, PDPINTx input low	Pulse duration, PDPINTx input low if bit 6 of SCSR2 = 1	Pulse duration, $\overrightarrow{PDPINTx}$ input lowif bit 6 of SCSR2 = 0 $6t_{C(CO)}$ if bit 6 of SCSR2 = 1 $12t_{C(CO)}$	Pulse duration, $\overrightarrow{PDPINTx}$ input lowif bit 6 of SCSR2 = 0 $6t_{C}(CO)$ if bit 6 of SCSR2 = 1 $12t_{C}(CO)$

<sup>†</sup> This is different from 240x devices.



<sup>†</sup> tOSC is the oscillator start-up time.

‡CLKOUT frequency after LPM2 wakeup will be the same as that upon entering LPM2 (x4 shown as an example).

§ PDPINTx interrupt vector, if PDPINTx interrupt is enabled.

¶ If PDPINTx interrupt is disabled.

Figure 32. LPM2 Wakeup Using PDPINTx



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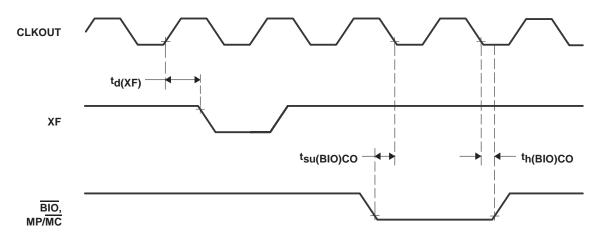
### XF, BIO, and MP/MC timings

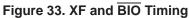
### switching characteristics over recommended operating conditions (see Figure 33)

PARAMETER		MIN	MAX	UNIT
<sup>t</sup> d(XF)	Delay time, CLKOUT high to XF high/low	-3	7	ns

### timing requirements (see Figure 33)

		MIN	MAX	UNIT
t <sub>su(BIO)</sub> CO	Setup time, BIO or MP/MC low before CLKOUT low	0		ns
<sup>t</sup> h(BIO)CO	Hold time, BIO or MP/MC low after CLKOUT low	19		ns





TIMING EVENT MANAGER INTERFACE

### **PWM timings**

PWM refers to all PWM outputs on EVA and EVB.

# switching characteristics over recommended operating conditions for PWM timing $[H = 0.5t_{c(CO)}]$ (see Figure 34)

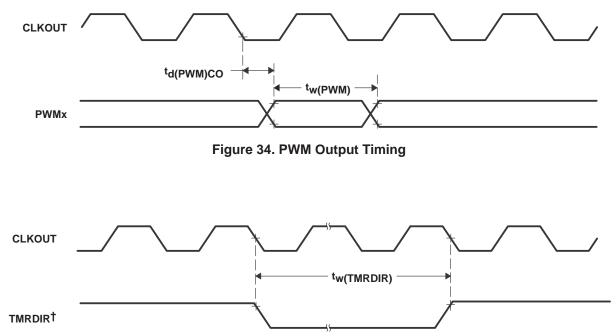
PARAMETER				UNIT
<sup>t</sup> w(PWM) <sup>†</sup>	Pulse duration, PWMx output high/low	2H+5		ns
<sup>t</sup> d(PWM)CO	Delay time, CLKOUT low to PWMx output switching		15	ns

<sup>†</sup> PWM outputs may be 100%, 0%, or increments of t<sub>c(CO)</sub> with respect to the PWM period.

### timing requirements<sup>‡</sup> [H = 0.5t<sub>c(CO)</sub>] (see Figure 35)

	MIN	MAX	UNIT
Pulse duration, TMRDIR low/high	4H+5		ns
Pulse duration, TMRCLK low as a percentage of TMRCLK cycle time	40	60	%
Pulse duration, TMRCLK high as a percentage of TMRCLK cycle time	40	60	%
Cycle time, TMRCLK	$4 \times t_{c(CO)}$		ns
	Pulse duration, TMRCLK low as a percentage of TMRCLK cycle time Pulse duration, TMRCLK high as a percentage of TMRCLK cycle time	Pulse duration, TMRDIR low/high       4H+5         Pulse duration, TMRCLK low as a percentage of TMRCLK cycle time       40         Pulse duration, TMRCLK high as a percentage of TMRCLK cycle time       40	Pulse duration, TMRDIR low/high       4H+5         Pulse duration, TMRCLK low as a percentage of TMRCLK cycle time       40       60         Pulse duration, TMRCLK high as a percentage of TMRCLK cycle time       40       60

 $\ddagger$  Parameter TMRDIR is equal to the pin TDIRx, and parameter TMRCLK is equal to the pin TCLKINx.



<sup>†</sup> Parameter TMRDIR is equal to the pin TDIRx.

Figure 35. Capture/TMRDIR Timing



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### capture and QEP timings

CAP refers to all QEP and capture input pins.

### timing requirements (see Figure 36)

			MIN	MAX	UNIT
t (ata)t	Dulas duration CADy input low/high	if bit 6 of SCSR2 = $0$	6t <sub>C</sub> (CO)		ns
'w(CAP)'	tw(CAP) <sup>†</sup> Pulse duration, CAPx input low/high	if bit 6 of SCSR2 = 1	12t <sub>c(CO)</sub>		115

<sup>†</sup>This is different from 240x devices.

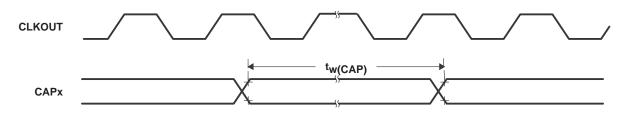


Figure 36. Capture Input and QEP Timing



### interrupt timings

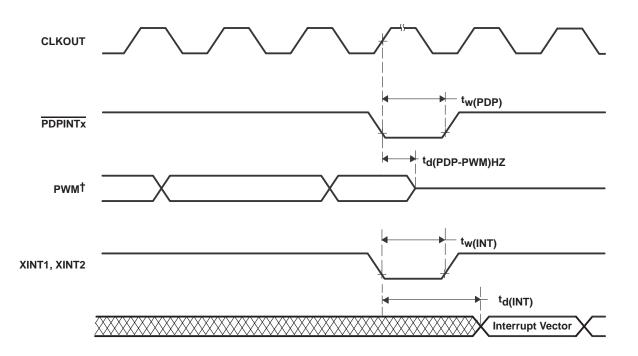
INT refers to XINT1 and XINT2. PDP refers to PDPINTx.

### switching characteristics over recommended operating conditions (see Figure 37)

PARAMETER			MAX	UNIT
<sup>t</sup> d(PDP-PWM)HZ	d(PDP-PWM)HZ Delay time, PDPINTx low to PWM high-impedance state			
<sup>t</sup> d(INT)	Delay time, INT low/high to interrupt-vector fetch	10t <sub>c(CO)</sub>		ns

### timing requirements (see Figure 37)

			MIN	MAX	UNIT
		if bit 6 of SCSR2 = 0	6t <sub>c(CO)</sub>		50
<sup>t</sup> w(INT)	Pulse duration, in Finput low/high	if bit 6 of SCSR2 = 1	12t <sub>c(CO)</sub>		ns
		if bit 6 of SCSR2 = 0	6t <sub>c(CO)</sub>		
<sup>t</sup> w(PDP)	Pulse duration, PDPINTx input low	if bit 6 of SCSR2 = 1	12t <sub>c(CO)</sub>		ns



<sup>†</sup> PWM refers to **all** the PWM pins in the device (i.e., PWMn and TnPWM pins). The state of the PWM pins after PDPINTx is taken high depends on the state of the FCOMPOE bit.

### Figure 37. Power Drive Protection Interrupt Timing

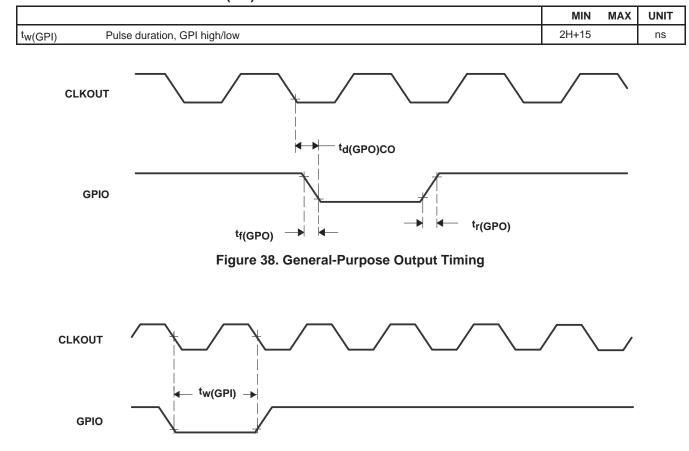
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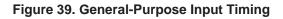
### general-purpose input/output timings

### switching characteristics over recommended operating conditions (see Figure 38)

PARAMETER			MIN MAX	UNIT
<sup>t</sup> d(GPO)CO	Delay time, CLKOUT low to GPIO low/high	All GPIOs	9	ns
<sup>t</sup> r(GPO)	Rise time, GPIO switching low to high	All GPIOs	8	ns
<sup>t</sup> f(GPO)	Fall time, GPIO switching high to low	All GPIOs	6	ns

### timing requirements $[H = 0.5t_{C(CO)}]$ (see Figure 39)







### SPI MASTER MODE TIMING PARAMETERS

SPI master mode timing information is listed in the following tables.

### SPI master mode external timing parameters (clock phase = 0)<sup>†‡</sup> (see Figure 40)

NO.			SPI WHEN (SPIBRF OR SPIBRR =		SPI WHEN (SI IS ODD AND S		UNIT
			MIN	MAX	MIN	MAX	
1	<sup>t</sup> c(SPC)M	Cycle time, SPICLK	4t <sub>c</sub> (CO)	128t <sub>c(CO)</sub>	5t <sub>c(CO)</sub>	127t <sub>c(CO)</sub>	ns
2§	<sup>t</sup> w(SPCH)M	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>c</sub> (SPC)M <sup>-0.5t</sup> c(CO) <sup>-10</sup>	$0.5t_{C}(SPC)M - 0.5t_{C}(CO)$	
28	<sup>t</sup> w(SPCL)M	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>c</sub> (SPC)M-0.5t <sub>c</sub> (CO)-10	$0.5t_{C}(SPC)M - 0.5t_{C}(CO)$	ns
3§	<sup>t</sup> w(SPCL)M	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>C</sub> (SPC)M+0.5t <sub>C</sub> (CO)-10	$0.5t_{C}(SPC)M + 0.5t_{C}(CO)$	
33	<sup>t</sup> w(SPCH)M	Pulse duration, SPICLK high (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>C</sub> (SPC)M+0.5t <sub>C</sub> (CO)-10	$0.5t_{C}(SPC)M + 0.5t_{C}(CO)$	ns
3.	<sup>t</sup> d(SPCH-SIMO)M	Delay time, SPICLK high to SPISIMO valid (clock polarity = 0)	- 10	10	- 10	10	
4§	<sup>t</sup> d(SPCL-SIMO)M	Delay time, SPICLK low to SPISIMO valid (clock polarity = 1)	- 10	10	- 10	10	ns
-6	<sup>t</sup> v(SPCL-SIMO)M	Valid time, SPISIMO data valid after SPICLK low (clock polarity =0)	0.5t <sub>c(SPC)M</sub> -10		0.5t <sub>C</sub> (SPC)M+0.5t <sub>C</sub> (CO)-10		
5§	<sup>t</sup> v(SPCH-SIMO)M	Valid time, SPISIMO data valid after SPICLK high (clock polarity =1)	0.5t <sub>c(SPC)M</sub> -10		0.5t <sub>c(SPC)M</sub> +0.5t <sub>c(CO)</sub> -10		ns
-6	<sup>t</sup> su(SOMI-SPCL)M	Setup time, SPISOMI before SPICLK low (clock polarity = 0)	0		0		
8§	<sup>t</sup> su(SOMI-SPCH)M	Setup time, SPISOMI before SPICLK high (clock polarity = 1)	0		0		ns
9§	<sup>t</sup> v(SPCL-SOMI)M	Valid time, SPISOMI data valid after SPICLK low (clock polarity = 0)	0.25t <sub>c(SPC)M</sub> -10		0.5t <sub>c(SPC)M</sub> -0.5t <sub>c(CO)</sub> -10		
98	<sup>t</sup> v(SPCH-SOMI)M	Valid time, SPISOMI data valid after SPICLK high (clock polarity = 1)	0.25t <sub>c(SPC)M</sub> -10		0.5t <sub>c</sub> (SPC)M-0.5t <sub>c</sub> (CO)-10		ns

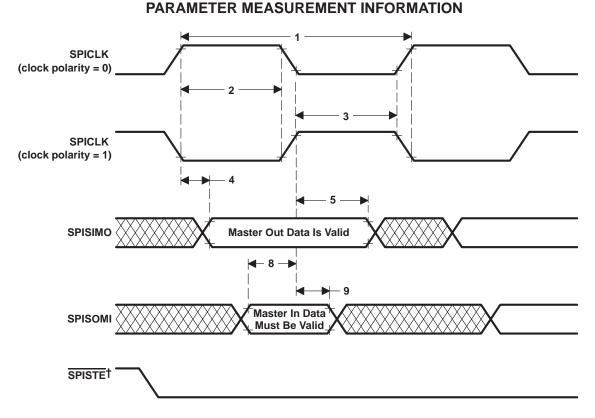
<sup>†</sup> The MASTER/SLAVE bit (SPICTL.2) is set and the CLOCK PHASE bit (SPICTL.3) is cleared. <sup>‡</sup>  $t_c$  = system clock cycle time = 1/CLKOUT =  $t_{c(CO)}$ § The active edge of the SPICLK signal referenced is controlled by the CLOCK POLARITY bit (SPICCR.6).

# TMS320LC2406A, TMS320LF2407A, TMS320LF2406A, TMS320LF2402A **FMS320LC2404A**, SPRS145B DSP CONTROL TMS320LC2402A CONTROLLERS

# **ADVANCE INFORMATION**

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<sup>†</sup> The SPISTE signal must be active before the SPI communication stream starts; the SPISTE signal must remain active until the SPI communication stream is complete.

### Figure 40. SPI Master Mode External Timing (Clock Phase = 0)



NO.			SPI WHEN (SPIBRE OR SPIBRE :		SPI WHEN (S IS ODD AND S		UNIT
			MIN	MAX	MIN	MAX	1
1	<sup>t</sup> c(SPC)M	Cycle time, SPICLK	4t <sub>c</sub> (CO)	128t <sub>C(CO)</sub>	5t <sub>c(CO)</sub>	127t <sub>C(CO)</sub>	ns
2§	<sup>t</sup> w(SPCH)M	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>c</sub> (SPC)M <sup>-0.5t</sup> c(CO) <sup>-10</sup>	0.5t <sub>C</sub> (SPC)M -0.5t <sub>C</sub> (CO)	
28	<sup>t</sup> w(SPCL)M	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> -10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>C</sub> (SPC)M <sup>-0.5t</sup> <sub>C</sub> (CO) <sup>-10</sup>	0.5t <sub>C</sub> (SPC)M -0.5t <sub>C</sub> (CO)	ns
3-5	<sup>t</sup> w(SPCL)M	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>c</sub> (SPC)M-10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>C</sub> (SPC)M+0.5t <sub>C</sub> (CO)-10	$0.5t_{C}(SPC)M + 0.5t_{C}(CO)$	
3§	<sup>t</sup> w(SPCH)M	Pulse duration, SPICLK high (clock polarity = 1)	0.5t <sub>c</sub> (SPC)M-10	0.5t <sub>C</sub> (SPC)M	0.5t <sub>C</sub> (SPC)M+0.5t <sub>C</sub> (CO)-10	$0.5t_{C}(SPC)M + 0.5t_{C}(CO)$	ns
3-0	<sup>t</sup> su(SIMO-SPCH)M	Setup time, SPISIMO data valid before SPICLK high (clock polarity = 0)	0.5t <sub>C(SPC)M</sub> -10		0.5t <sub>C</sub> (SPC)M -10		
6§	<sup>t</sup> su(SIMO-SPCL)M	Setup time, SPISIMO data valid before SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> -10		0.5t <sub>C</sub> (SPC)M -10		ns
3-	<sup>t</sup> v(SPCH-SIMO)M	Valid time, SPISIMO data valid after SPICLK high (clock polarity =0)	0.5t <sub>c(SPC)M</sub> -10		0.5t <sub>c(SPC)M</sub> -10		
7§	<sup>t</sup> v(SPCL-SIMO)M	Valid time, SPISIMO data valid after SPICLK low (clock polarity =1)	0.5t <sub>C(SPC)M</sub> -10		0.5t <sub>C</sub> (SPC)M-10		ns
	<sup>t</sup> su(SOMI-SPCH)M	Setup time, SPISOMI before SPICLK high (clock polarity = 0)	0		0		
10§	<sup>t</sup> su(SOMI-SPCL)M	Setup time, SPISOMI before SPICLK low (clock polarity = 1)	0		0		ns
11§	<sup>t</sup> v(SPCH-SOMI)M	Valid time, SPISOMI data valid after SPICLK high (clock polarity = 0)	0.25t <sub>c(SPC)M</sub> -10		0.5t <sub>c(SPC)M</sub> -10		
113	<sup>t</sup> v(SPCL-SOMI)M	Valid time, SPISOMI data valid after SPICLK low (clock polarity = 1)	0.25t <sub>C(SPC)M</sub> -10		0.5t <sub>c(SPC)M</sub> -10		ns

### SPI master mode external timing parameters (clock phase = 1)<sup> $\dagger$ </sup> (see Figure 41)

<sup>†</sup> The MASTER/SLAVE bit (SPICTL.2) is set and the CLOCK PHASE bit (SPICTL.3) is set. <sup>‡</sup> t<sub>C</sub> = system clock cycle time = 1/CLKOUT = t<sub>C</sub>(CO) § The active edge of the SPICLK signal referenced is controlled by the CLOCK POLARITY bit (SPICCR.6).

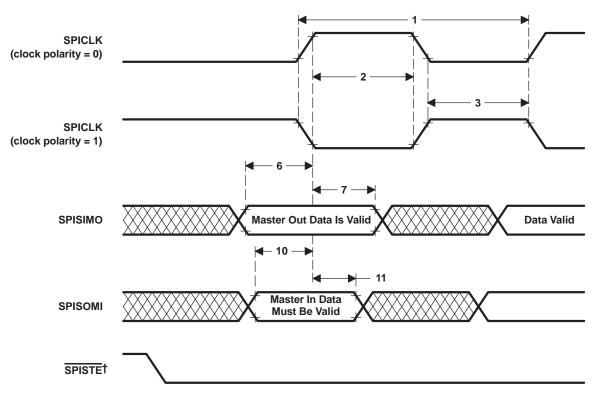
# TMS320LC2406A, TMS320LF2407A, TMS320LF2406A, TMS320LF2402A FMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS SPRS145B - JULY 2000 - REVISED DECEMBER 2000

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## **ADVANCE INFORMATION**

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PARAMETER MEASUREMENT INFORMATION

<sup>†</sup> The SPISTE signal must be active before the SPI communication stream starts; the SPISTE signal must remain active until the SPI communication stream is complete.

Figure 41. SPI Master Mode External Timing (Clock Phase = 1)

**ADVANCE INFORMATION** 



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### SPI SLAVE MODE TIMING PARAMETERS

Slave mode timing information is listed in the following tables.

### SPI slave mode external timing parameters (clock phase = 0)<sup>†‡</sup> (see Figure 42)

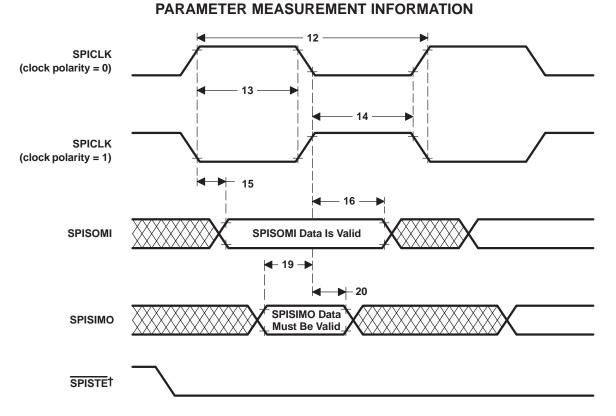
NO.			MIN	MAX	UNIT
12	<sup>t</sup> c(SPC)S	Cycle time, SPICLK	4t <sub>c(CO)</sub> ‡		ns
13§	<sup>t</sup> w(SPCH)S	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)S-10</sub>	0.5t <sub>c(SPC)S</sub>	
138	<sup>t</sup> w(SPCL)S	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>C</sub> (SPC)S-10	0.5t <sub>C</sub> (SPC)S	ns
14§	<sup>t</sup> w(SPCL)S	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>C</sub> (SPC)S-10	0.5t <sub>C</sub> (SPC)S	
148	<sup>t</sup> w(SPCH)S	Pulse duration, SPICLK high (clock polarity = 1)	0.5t <sub>C</sub> (SPC)S-10	0.5t <sub>C</sub> (SPC)S	ns
15§	<sup>t</sup> d(SPCH-SOMI)S	Delay time, SPICLK high to SPISOMI valid (clock polarity = 0)	0.375t <sub>c(SPC)S</sub> -10		ns
	td(SPCL-SOMI)S	Delay time, SPICLK low to SPISOMI valid (clock polarity = 1)	0.375t <sub>C(SPC)S</sub> -10		
201	<sup>t</sup> v(SPCL-SOMI)S	Valid time, SPISOMI data valid after SPICLK low (clock polarity =0)	0.75t <sub>c</sub> (SPC)S		
16§	<sup>t</sup> v(SPCH-SOMI)S	Valid time, SPISOMI data valid after SPICLK high (clock polarity =1)	0.75t <sub>c</sub> (SPC)S		ns
4.05	t <sub>su</sub> (SIMO-SPCL)S	Setup time, SPISIMO before SPICLK low (clock polarity = 0)	0		
19§	t <sub>su</sub> (SIMO-SPCH)S	Setup time, SPISIMO before SPICLK high (clock polarity = 1)	0		ns
20§	tv(SPCL-SIMO)S	Valid time, SPISIMO data valid after SPICLK low (clock polarity = 0)	0.5t <sub>C</sub> (SPC)S		20
208	<sup>t</sup> v(SPCH-SIMO)S	Valid time, SPISIMO data valid after SPICLK high (clock polarity = 1)	0.5t <sub>c</sub> (SPC)S		ns

<sup>†</sup> The MASTER/SLAVE bit (SPICTL.2) is cleared and the CLOCK PHASE bit (SPICTL.3) is cleared.

 $t_{c}$  = system clock cycle time = 1/CLKOUT =  $t_{c}(CO)$ § The active edge of the SPICLK signal referenced is controlled by the CLOCK POLARITY bit (SPICCR.6).



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<sup>†</sup> The SPISTE signal must be active before the SPI communication stream starts; the SPISTE signal must remain active until the SPI communication stream is complete.

### Figure 42. SPI Slave Mode External Timing (Clock Phase = 0)



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NO.			MIN	MAX	UNIT
12	<sup>t</sup> c(SPC)S	Cycle time, SPICLK	8t <sub>C</sub> (CO)		ns
13§	<sup>t</sup> w(SPCH)S	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>C(SPC)S</sub> -10	0.5t <sub>c</sub> (SPC)S	ns
138	<sup>t</sup> w(SPCL)S	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>C(SPC)S</sub> -10		
14§	<sup>t</sup> w(SPCL)S	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>C(SPC)S</sub> -10	0.5t <sub>C</sub> (SPC)S	
148	<sup>t</sup> w(SPCH)S	Pulse duration, SPICLK high (clock polarity = 1)	0.5t <sub>C(SPC)S</sub> -10	0.5t <sub>C</sub> (SPC)S	ns
17§	tsu(SOMI-SPCH)S	Setup time, SPISOMI before SPICLK high (clock polarity = 0)	0.125t <sub>c(SPC)S</sub>		
178	tsu(SOMI-SPCL)S	Setup time, SPISOMI before SPICLK low (clock polarity = 1)	0.125t <sub>c(SPC)S</sub>		ns
6	<sup>t</sup> v(SPCH-SOMI)S	Valid time, SPISOMI data valid after SPICLK high (clock polarity =0)	0.75t <sub>c</sub> (SPC)S		
18§	tv(SPCL-SOMI)S	Valid time, SPISOMI data valid after SPICLK low (clock polarity =1)	0.75t <sub>c(SPC)S</sub>		ns
310	t <sub>su</sub> (SIMO-SPCH)S	Setup time, SPISIMO before SPICLK high (clock polarity = 0)	0		
21§	t <sub>su</sub> (SIMO-SPCL)S	Setup time, SPISIMO before SPICLK low (clock polarity = 1)	0		ns
22§	<sup>t</sup> v(SPCH-SIMO)S	Valid time, SPISIMO data valid after SPICLK high (clock polarity = 0)	0.5t <sub>c</sub> (SPC)S		
223	tv(SPCL-SIMO)S	Valid time, SPISIMO data valid after SPICLK low (clock polarity = 1)	0.5t <sub>C</sub> (SPC)S		ns

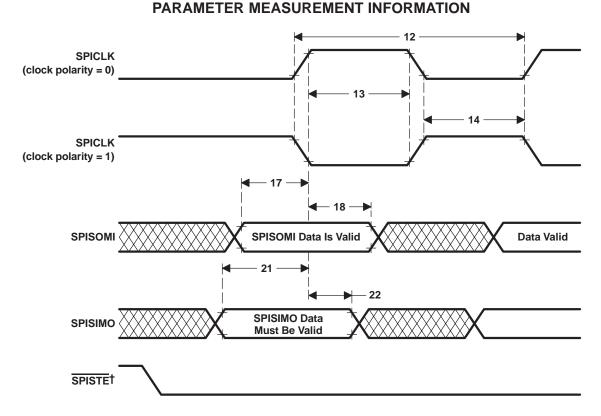
### SPI slave mode external timing parameters (clock phase = 1)<sup> $\dagger$ </sup> (see Figure 43)

<sup>+</sup> The MASTER/SLAVE bit (SPICTL.2) is cleared and the CLOCK PHASE bit (SPICTL.3) is set.

 $t_c$  = system clock cycle time = 1/CLKOUT =  $t_c(CO)$ § The active edge of the SPICLK signal referenced is controlled by the CLOCK POLARITY bit (SPICCR.6).



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<sup>†</sup> The SPISTE signal must be active before the SPI communication stream starts; the SPISTE signal must remain active until the SPI communication stream is complete.

### Figure 43. SPI Slave Mode External Timing (Clock Phase = 1)



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### external memory interface read timings

# switching characteristics over recommended operating conditions for an external memory interface read at 30 MHz [H = $0.5t_{c(CO)}$ ] (see Figure 44)

	PARAMETER	MIN	MAX	UNIT
<sup>t</sup> d(COL–CNTL)	Delay time, CLKOUT low to control valid		6	ns
<sup>t</sup> d(COL–CNTH)	Delay time, CLKOUT low to control inactive		6	ns
<sup>t</sup> d(COL–A)RD	Delay time, CLKOUT low to address valid		8	ns
td(COH-RDL)	Delay time, CLKOUT high to RD strobe active		6	ns
<sup>t</sup> d(COL–RDH)	Delay time, CLKOUT low to RD strobe inactive high	-8	1	ns
td(COL-SL)	Delay time, CLKOUT low to STRB strobe active low		6	ns
td(COL-SH)	Delay time, CLKOUT low to STRB strobe inactive high		6	ns
<sup>t</sup> d(WRN)	Delay time, W/ $\overline{R}$ going low to R/ $\overline{W}$ rising		5	ns
<sup>t</sup> h(A)COL	Hold time, address valid after CLKOUT low	2		ns
<sup>t</sup> su(A)RD	Setup time, address valid before RD strobe active low	H – 7		ns
<sup>t</sup> h(A)RD	Hold time, address valid after RD strobe inactive high	0		ns

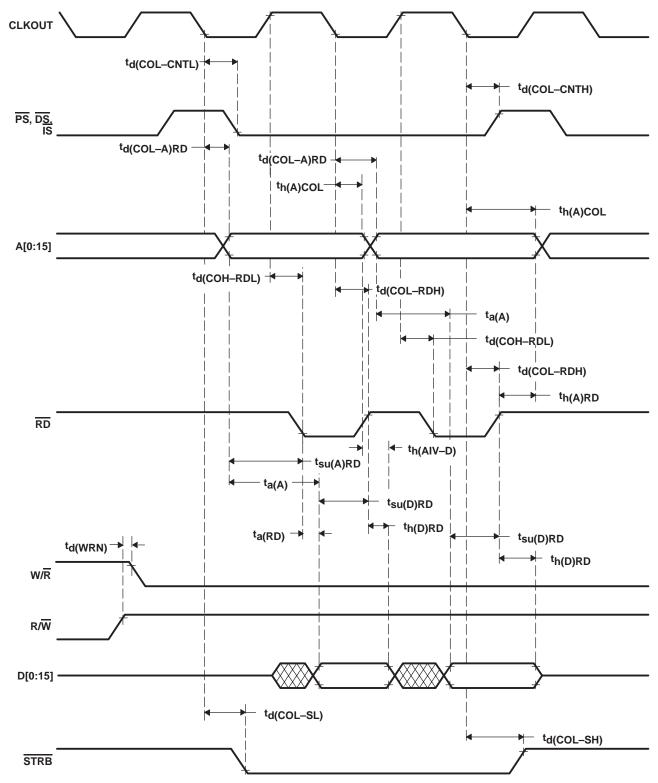
### timing requirements $[H = 0.5t_{c(CO)}]$ (see Figure 44)

		MIN	MAX	UNIT
<sup>t</sup> a(A)	Access time, read data from address valid		2H –1 3	ns
<sup>t</sup> a(RD)	Access time, read data from RD low		H – 7	ns
t <sub>su(D)RD</sub>	Setup time, read data before RD strobe inactive high	10		ns
<sup>t</sup> h(D)RD	Hold time, read data after RD strobe inactive high	0		ns
<sup>t</sup> h(AIV-D)	Hold time, read data after address invalid	0		ns



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### external memory interface read timings (continued)







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### external memory interface write timings

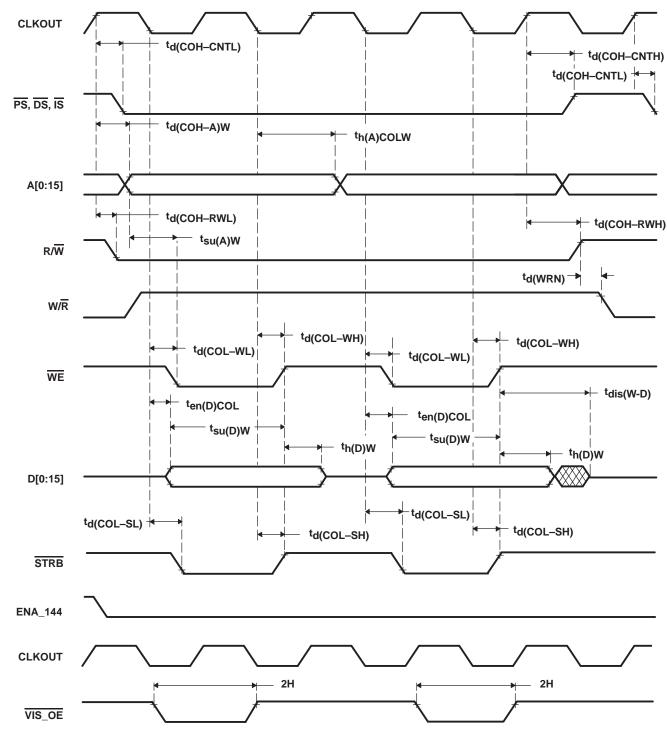
# switching characteristics over recommended operating conditions for an external memory interface write at 30 MHz [H = $0.5t_{c(CO)}$ ] (see Figure 45)

	PARAMETER	MIN	MAX	UNIT
td(COH–CNTL)	Delay time, CLKOUT high to control valid		6	ns
td(COH–CNTH)	Delay time, CLKOUT high to control inactive		6	ns
<sup>t</sup> d(COH–A)W	Delay time, CLKOUT high to address valid		10	ns
<sup>t</sup> d(COH–RWL)	Delay time, CLKOUT high to R/W low		6	ns
<sup>t</sup> d(COH–RWH)	Delay time, CLKOUT high to $R/\overline{W}$ high		6	ns
td(COL–WL)	Delay time, CLKOUT low to WE strobe active low		7	ns
td(COL–WH)	Delay time, CLKOUT low to WE strobe inactive high		7	ns
ten(D)COL	Enable time, data bus driven from CLKOUT low	-3		ns
td(COL–SL)	Delay time, CLKOUT low to STRB active low		6	ns
td(COL–SH)	Delay time, CLKOUT low to STRB inactive high		6	ns
<sup>t</sup> d(WRN)	Delay time, W/R going low to R/W rising		5	ns
<sup>t</sup> h(A)COLW	Hold time, address valid after CLKOUT low	5		ns
t <sub>su(A)W</sub>	Setup time, address valid before WE strobe active low	H–9		ns
<sup>t</sup> su(D)W	Setup time, write data before WE strobe inactive high	2H–17		ns
<sup>t</sup> h(D)W	Hold time, write data after WE strobe inactive high	0		ns
<sup>t</sup> dis(W-D)	Disable time, data bus high impedance from WE high	5		ns



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### external memory interface write timings (continued)



NOTE A: VIS\_OE will be visible at pin 97 of LF2407A when ENA\_144 is low along with BVIS bits (10,9 of WSGR register – FFFh@I/O) set to 10 or 11. CLKOUT and VIS\_OE indicate internal memory write cycles (program/data). During VIS\_OE cycles, the external bus will be driven. CLKOUT is to be used along with VIS\_OE for trace capabilities.

Figure 45. Address Visibility Mode



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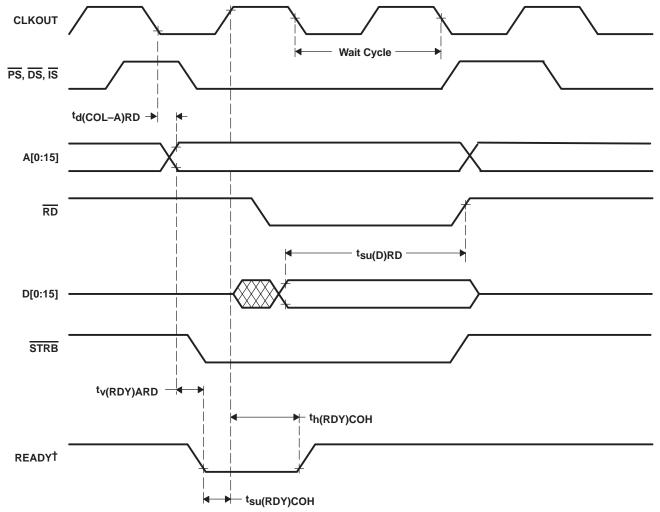
### external memory interface ready-on-read timings

# switching characteristics over recommended operating conditions for an external memory interface ready-on-read (see Figure 46)

	PARAMETER	MIN	MAX	UNIT
<sup>t</sup> d(COL–A)RD	Delay time, CLKOUT low to address valid		8	ns

### timing requirements for an external memory interface ready-on-read (see Figure 46)

		MIN	MAX	UNIT
<sup>t</sup> h(RDY)COH	Hold time, READY after CLKOUT high	-3		ns
t <sub>su</sub> (D)RD	Setup time, read data before RD strobe inactive high	10		ns
<sup>t</sup> v(RDY)ARD	Valid time, READY after address valid on read		-2	ns
t <sub>su</sub> (RDY)COH	Setup time, READY before CLKOUT high	22		ns



<sup>†</sup> The WSGR register must be programmed before the READY pin can be used. See the READY pin description for more details.

Figure 46. Ready-on-Read Timings



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### external memory interface ready-on-read timings (continued)

# timing requirements for an external memory interface ready-on-read with one software wait state and one external wait state (see Figure 47)

		MIN	MAX	UNIT
<sup>t</sup> h(RDY)COH	Hold time, READY after CLKOUT high	14		ns
<sup>t</sup> su(RDY)COH	Setup time, READY before CLKOUT high	7		ns
<sup>t</sup> d(COL–A)RD	Delay time, CLKOUT low to address valid		8	ns

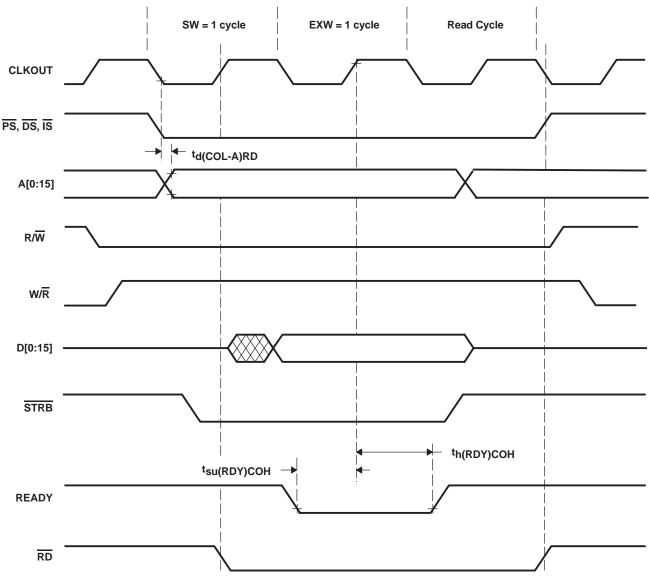


Figure 47. Ready-on-Read Timings With One Software Wait (SW) State and One External Wait (EXW) State



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### external memory interface ready-on-write timings

# switching characteristics over recommended operating conditions for an external memory interface ready-on-write (see Figure 48)

	PARAMETER	MIN	MAX	UNIT
<sup>t</sup> d(COH–A)W	Delay time, CLKOUT high to address valid		10	ns

# timing requirements for an external memory interface ready-on-write $[H = 0.5t_{c(CO)}]$ (see Figure 48)

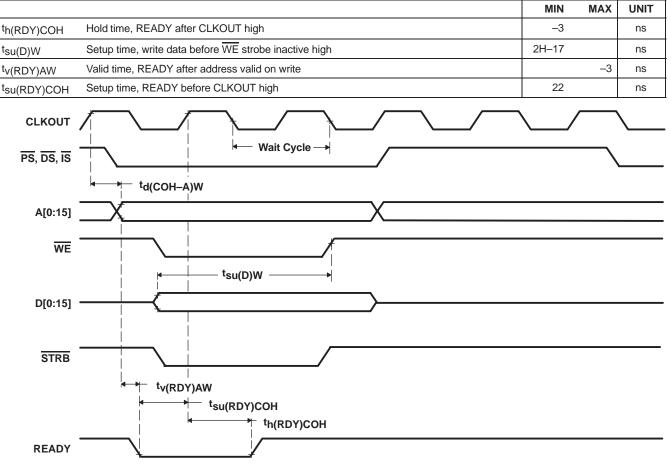


Figure 48. Ready-on-Write Timings



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### external memory interface ready-on-write timings (continued)

### timing requirements for an external memory interface ready-on-write with one software wait state and one external wait state (see Figure 49)

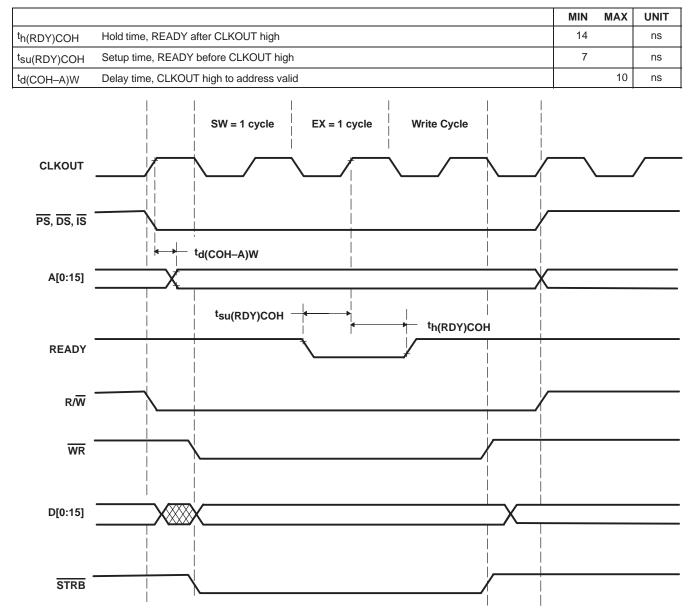


Figure 49. Ready-on-Write Timings With One Software Wait State and One External Wait State



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### 10-bit analog-to-digital converter (ADC)

### recommended operating conditions

		MIN	NOM	MAX	UNIT
VCCA	Analog supply voltage	3.0	3.3	3.6	V
VSSA	Analog ground		0		V
VREFHI	Analog supply reference source <sup>†</sup>	VREFLO		VCCA	V
VREFLO	Analog ground reference source <sup>†</sup>	VSSA		VREFHI	V
VAI	Analog input voltage, ADCIN00–ADCIN07	VREFLO		VREFHI	V

<sup>†</sup> V<sub>REFHI</sub> and V<sub>REFLO</sub> must be stable, within ±1/2 LSB of the required resolution, during the entire conversion time.

### ADC operating frequency

	MIN	MAX	UNIT
ADC operating frequency	4	30	MHz



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### operating characteristics over recommended operating condition ranges<sup>†</sup>

	PARAMETER	DESCRIPTION		MIN	TYP	MAX	UNIT
		V <sub>CCA</sub> = 3.3 V			10	15	mA
ICCA	Analog supply current	V <sub>CCA</sub> = V <sub>REFHI</sub> = 3.3 V	PLL or OSC power down			1	μA
IADREFHI	VREFHI input current				0.75	1.5	mA
IADCIN	Analog input leakage					1	μΑ
~		Typical capacitive load on	Non-sampling		10		
C <sub>ai</sub>	Analog input capacitance analog input pin		Sampling		30		pF
EDNL	Differential nonlinearity error	Difference between the actual step width and the ideal value				±2	LSB
E <sub>INL</sub>	Integral nonlinearity error	Maximum deviation from the best straight line through the ADC transfer characteristics, excluding the quantization error				±2	LSB
<sup>t</sup> d(PU)	Delay time, power-up to ADC valid	Time to stabilize analog stage after power-up				10	μs
Z <sub>AI</sub>	Analog input source impedance	Analog input source impedance needed for conversions to remain within specifications at min <sup>t</sup> w(SH)				10	Ω

<sup>†</sup> Absolute resolution = 3.22 mV. At V<sub>REFHI</sub> = 3.3 V and V<sub>REFLO</sub> = 0 V, this is one LSB. As V<sub>REFHI</sub> decreases, V<sub>REFLO</sub> increases, or both, the LSB size decreases. Therefore, the absolute accuracy and differential/integral linearity errors in terms of LSBs increase.

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### internal ADC module timings<sup>†</sup> (see Figure 50)

		MIN	MAX	UNIT
<sup>t</sup> c(AD)	Cycle time, ADC prescaled clock		33.3	ns
<sup>t</sup> w(SHC)	Pulse duration, total sample/hold and conversion time‡		500	ns
<sup>t</sup> w(SH)	Pulse duration, sample and hold time	2t <sub>c(AD)</sub> §	32t <sub>C(AD)</sub>	ns
<sup>t</sup> w(C)	Pulse duration, total conversion time	10t <sub>c(AD)</sub>		ns
td(SOC-SH)	Delay time, start of conversion to beginning of sample and hold	3t <sub>c(CO)</sub>		ns
<sup>t</sup> d(EOC)	Delay time, end of conversion to data loaded into result register	2t <sub>c(CO)</sub>		ns
<sup>t</sup> d(ADCINT)	Delay time, ADC flag to ADC interrupt	2t <sub>c(CO)</sub>		ns

<sup>†</sup> The ADC timing diagram represents a typical conversion sequence. Refer to the ADC chapter in the *TMS320LF/LC240xA DSP Controllers Reference Guide: System and Peripherals* (literature number SPRU357A) for more details.

<sup>‡</sup> The total sample/hold and conversion time is determined by the summation of  $t_{d(SOC-SH)}$ ,  $t_{w(SH)}$ ,  $t_{w(C)}$ , and  $t_{d(EOC)}$ .

§ Can be varied by ACQ Prescalar bits in the ADCCTRL1 register

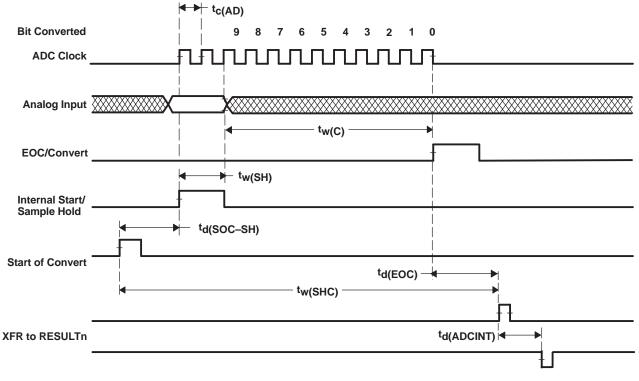


Figure 50. Analog-to-Digital Internal Module Timing



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### Flash parameters @40 MHz CLOCKOUT

	MIN	TYP	MAX	UNIT	
	Time/Word (16-bit)		30		μs
Clear/Programming time <sup>†</sup>	Time/4K Sector		130		ms
	Time/12K Sector		400		ms
Free the t	Time/4K Sector		350		ms
Erase time <sup>†</sup>	Time/12K Sector		1		S
ICCP (VCCP pin current)	> (V <sub>CCP</sub> pin current) Indicates the typical/maximum current consumption during the Clear-Erase-Program (C-E-P) cycle			15	mA

<sup>†</sup> The indicated time does not include the time it takes to load the C-E-P algorithm and the code (to be programmed) onto on-chip RAM. The values specified are when V<sub>DD</sub> = 3.3 V and V<sub>CCP</sub> = 5 V, and any deviation from these values could affect the timing parameters. Aging and process variance could also impact the timing parameters.

### migrating from 240x devices to 240xA devices

This section highlights the new features of the 240xA devices (as compared to the 240x family) and describes the impact these features have on user applications.

### maximum clock speed

240xA devices can operate at a maximum speed of 40 MHz compared to the 30-MHz operation of 240x devices. This change in clock speed warrants a change in the register contents of all the peripherals. For example, to maintain the same baud rate, the divisor values that are loaded to the SPI, SCI, and CAN registers must be recalculated.

### code security module

### input-qualifier circuitry

An input-qualifier circuitry qualifies the input signal to the CAP1–6, XINT1/2, ADCSOC, and PDPINTA/B pins in the LF240xA devices. The state of the internal input signal will change only after these pins are high/low for 6 (12) clock edges. The user must hold the pin high/low for 6 (12) cycles to ensure that the device see the level change. The increase in the pulse width of the signals used to excite these pins must be taken into account while migrating from the 240x to the 240xA family.

Bit 6 of the SCSR2 register controls whether 6 clock edges (bit 6 = 0) or 12 clock edges (bit 6 = 1) are used to block 5- or 11-cycle glitches. This bit is a "reserved" bit in 240x devices.

### status of the PDPINTx pin

The current status of the PDPINTx pins is now reflected in bit 8 of the COMCONx registers. This bit is a "reserved" bit in 240x devices.



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### peripheral register description

Table 18 is a collection of all the programmable registers of the LF240xA/LC240xA and is provided as a quick reference.

	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8		
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0		
ľ		•		DATA MEN	IORY SPACE	•	•	•		
[				CPU STATU	S REGISTERS					
[		ARP		OV	OVM	1	INTM	DP(8)		
[	DP(7)	DP(6)	DP(5)	DP(4)	DP(3)	DP(2)	DP(1)	DP(0)		
ļ		ARB		CNF	TC	SXM	С	1		
	1	1	1	XF	1	1		PM		
			GLOBAL	MEMORY AND C	PU INTERRUPT	REGISTERS				
)4h	_	—	—	—			—	—		
	_	—	INT6 MASK	INT5 MASK	INT4 MASK	INT3 MASK	INT2 MASK	INT1 MASK		
)5h		-		Res	served	-	-	-		
06h	_									
		—	INT6 FLAG	INT5 FLAG	INT4 FLAG	INT3 FLAG	INT2 FLAG	INT1 FLAG		
				SYSTEM	REGISTERS					
10h	IRQ0.15	IRQ0.14	IRQ0.13	IRQ0.12	IRQ0.11	IRQ0.10	IRQ0.9	IRQ0.8		
011	IRQ0.7	IRQ0.6	IRQ0.5	IRQ0.4	IRQ0.3	IRQ0.2	IRQ0.1	IRQ0.0		
11h	IRQ1.15	IRQ1.14	IRQ1.13	IRQ1.12	IRQ1.11	IRQ1.10	IRQ1.9	IRQ1.8		
	IRQ1.7	IRQ1.6	IRQ1.5	IRQ1.4	IRQ1.3	IRQ1.2	IRQ1.1	IRQ1.0		
12h	IRQ2.15	IRQ2.14	IRQ2.13	IRQ2.12	IRQ2.11	IRQ2.10	IRQ2.9	IRQ2.8		
	IRQ2.7	IRQ2.6	IRQ2.5	IRQ2.4	IRQ2.3	IRQ2.2	IRQ2.1	IRQ2.0		
3h					egal					
4h	IAK0.15	IAK0.14	IAK0.13	IAK0.12	IAK0.11	IAK0.10	IAK0.9	IAK0.8		
	IAK0.7	IAK0.6	IAK0.5	IAK0.4	IAK0.3	IAK0.2	IAK0.1	IAK0.0		
5h	IAK1.15	IAK1.14	IAK1.13	IAK1.12	IAK1.11	IAK1.10	IAK1.9	IAK1.8		
	IAK1.7	IAK1.6	IAK1.5	IAK1.4	IAK1.3	IAK1.2	IAK1.1	IAK1.0		
6h	IAK2.15	IAK2.14	IAK2.13	IAK2.12	IAK2.11	IAK2.10	IAK2.9	IAK2.8		
	IAK2.7	IAK2.6	IAK2.5	IAK2.4	IAK2.3	IAK2.2	IAK2.1	IAK2.0		
7h					egal					
18h	_	CLKSRC	LPM1	LPM0	CLK PS2	CLK PS1	CLK PS0	—		
	ADC CLKEN	SCI CLKEN	SPI CLKEN	CAN CLKEN	EVB CLKEN	EVA CLKEN	—	ILLADR		
19h	_	I/P QUALIFIER CLOCKS	WD OVERRIDE	XMIF HI Z	BOOT_EN	MP/MC	DON	PON		
1Ah	Illegal									
IBh	DINIG	DINIA	DINIAG	DINIAG	DINI44	DINIAG	DING	Dillo		
Ch	DIN15	DIN14	DIN13	DIN12	DIN11	DIN10	DIN9	DIN8		
	DIN7	DIN6	DIN5	DIN4	DIN3	DIN2	DIN1	DIN0		
Dh					egal		N/-			
Eh	V15	V14	V13	V12	V11	V10	V9	V8		
	V7	V6	V5	V4	V3 egal	V2	V1	V0		

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description

Indicates change with respect to the F243/F241, C242 device register maps.



# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLÉRS SPRS145B – JULY 2000 – REVISED DECEMBER 2000

### peripheral register description (continued)

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

ADDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	
ADDK	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
		-	-	WD CONTRO	OL REGISTERS	-		-	
07020h to 07022h	Illegal								
07023h	D7	D6	D5	D4	D3	D2	D1	D0	WDCNTR
07024h		I	I		legal	1		1	_
07025h	D7	D6	D5	D4	D3	D2	D1	D0	WDKEY
07026h to 07028h	lliegal								
07029h	WDFLAG	WDDIS	WDCHK2	WDCHK1	WDCHK0	WDPS2	WDPS1	WDPS0	WDCR
0702Ah		•	•	· · · · · · · · · · · · · · · · · · ·	•	-	-	-	
to 0703Fh	Illegal								
		1	PERIPHERAL IN	NTERFACE (SPI)		ON CONTROL R			
07040h	SPI SW RESET	CLOCK POLARITY	—	—	SPI CHAR3	SPI CHAR2	SPI CHAR1	SPI CHAR0	SPICCR
07041h	—	—	—	OVERRUN INT ENA	CLOCK PHASE	MASTER/ SLAVE	TALK	SPI INT ENA	SPICTL
07042h	RECEIVER OVERRUN FLAG	SPI INT FLAG	TX BUF FULL FLAG	_	_	_	_	_	SPISTS
07043h									
07044h	_	SPI BIT RATE 6	SPI BIT RATE 5	SPI BIT RATE 4	SPI BIT RATE 3	SPI BIT RATE 2	SPI BIT RATE 1	SPI BIT RATE 0	SPIBRR
07045h	lllegal								
07046h	ERXB15	ERXB14	ERXB13	ERXB12	ERXB11	ERXB10	ERXB9	ERXB8	SPIRXEMU
0704011	ERXB7	ERXB6	ERXB5	ERXB4	ERXB3	ERXB2	ERXB1	ERXB0	SFIRAEIVIU
070476	RXB15	RXB14	RXB13	RXB12	RXB11	RXB10	RXB9	RXB8	SPIRXBUF
07047h	RXB7	RXB6	RXB5	RXB4	RXB3	RXB2	RXB1	RXB0	
07048h	TXB15	TXB14	TXB13	TXB12	TXB11	TXB10	TXB9	TXB8	SPITXBUF
0704011	TXB7	TXB6	TXB5	TXB4	TXB3	TXB2	TXB1	TXB0	
07049h	SDAT15	SDAT14	SDAT13	SDAT12	SDAT11	SDAT10	SDAT9	SDAT8	SPIDAT
0704311	SDAT7	SDAT6	SDAT5	SDAT4	SDAT3	SDAT2	SDAT1	SDAT0	SLIPKI
0704Ah to 0704Eh					legal				
0704Fh	_	SPI PRIORITY	SPI SUSP SOFT	SPI SUSP FREE	_	_	_	_	SPIPRI

Indicates change with respect to the F243/F241, C242 device register maps.



# peripheral register description (continued)

	BIT 15	DIT 44	DIT 40	DIT 40			DITA	DIT O	7
ADDR	BIT 15 BIT 7	BIT 14 BIT 6	BIT 13 BIT 5	BIT 12 BIT 4	BIT 11 BIT 3	BIT 10 BIT 2	BIT 9 BIT 1	BIT 8 BIT 0	REG
				S INTERFACE (S				BILO	-
07050h	STOP BITS	EVEN/ODD PARITY	PARITY ENABLE	LOOP BACK ENA	ADDR/IDLE MODE	SCI CHAR2	SCI CHAR1	SCI CHAR0	SCICCR
07051h	_	RX ERR INT ENA	SW RESET	_	TXWAKE	SLEEP	TXENA	RXENA	SCICTL1
07052h	BAUD15 (MSB)	BAUD14	BAUD13	BAUD12	BAUD11	BAUD10	BAUD9	BAUD8	SCIHBAUD
07053h	BAUD7	BAUD6	BAUD5	BAUD4	BAUD3	BAUD2	BAUD1	BAUD0 (LSB)	SCILBAUD
07054h	TXRDY	TX EMPTY	_	_			RX/BK INT ENA	TX INT ENA	SCICTL2
07055h	RX ERROR	RXRDY	BRKDT	FE	OE	PE	RXWAKE		SCIRXST
07056h	ERXDT7	ERXDT6	ERXDT5	ERXDT4	ERXDT3	ERXDT2	ERXDT1	ERXDT0	SCIRXEMU
07057h	RXDT7	RXDT6	RXDT5	RXDT4	RXDT3	RXDT2	RXDT1	RXDT0	SCIRXBUF
07058h					legal				
07059h	TXDT7	TXDT6	TXDT5	TXDT4	TXDT3	TXDT2	TXDT1	TXDT0	SCITXBUF
0705Ah to 0705Eh				II	legal				
0705Fh	_	SCITX PRIORITY	SCIRX PRIORITY	SCI SOFT	SCI FREE		—		SCIPRI
07060h to 0706Fh				II	legal				
			EXTER	NAL INTERRUP	T CONTROL RE	GISTERS			
07070h	XINT1 FLAG	_	_	-	_	_	—	_	XINT1CR
0/0/01	_	_	_	_	_	XINT1 POLARITY	XINT1 PRIORITY	XINT1 ENA	XINTION
07071h	XINT2 FLAG	_		_					XINT2CR
	_	_	_	_	_	XINT2 POLARITY	XINT2 PRIORITY	XINT2 ENA	_
07072h to 0708Fh				II	legal				
				DIGITAL I/O CON	NTROL REGISTI	ERS			
07000	MCRA.15	MCRA.14	MCRA.13	MCRA.12	MCRA.11	MCRA.10	MCRA.9	MCRA.8	
07090h	MCRA.7	MCRA.6	MCRA.5	MCRA.4	MCRA.3	MCRA.2	MCRA.1	MCRA.0	MCRA
07091h				II	legal				
07000	MCRB.15	MCRB.14	MCRB.13	MCRB.12	MCRB.11	MCRB.10	MCRB.9	MCRB.8	MODD
07092h	MCRB.7	MCRB.6	MCRB.5	MCRB.4	MCRB.3	MCRB.2	MCRB.1	MCRB.0	MCRB
07093h				II	legal				
	MCRC.15	MCRC.14	MCRC.13	MCRC.12	MCRC.11	MCRC.10	MCRC.9	MCRC.8	
	MCRC.7	MCRC.6	MCRC.5	MCRC.4	MCRC.3	MCRC.2	MCRC.1	MCRC.0	MCRC
07094h									
07094h 07095h	E7DIR	E6DIR	E5DIR	E4DIR	E3DIR	E2DIR	E1DIR	E0DIR	PEDATDIR

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)



### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

ADDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	REG
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
			DIGITAL	I/O CONTROL R	EGISTERS (CO	NTINUED)			
7096h	—	F6DIR	F5DIR	F4DIR	F3DIR	F2DIR	F1DIR	F0DIR	PFDATDIR
109011		IOPF6	IOPF5	IOPF4	IOPF3	IOPF2	IOPF1	IOPF0	FIDAIDIR
7098h	A7DIR	A6DIR	A5DIR	A4DIR	A3DIR	A2DIR	A1DIR	A0DIR	PADATDIR
03011	IOPA7	IOPA6	IOPA5	IOPA4	IOPA3	IOPA2	IOPA1	IOPA0	TADATOIR
7099h				Ille	egal				
709Ah	B7DIR	B6DIR	B5DIR	B4DIR	B3DIR	B2DIR	B1DIR	B0DIR	PBDATDIR
00/ 11	IOPB7	IOPB6	IOPB5	IOPB4	IOPB3	IOPB2	IOPB1	IOPB0	
09Bh			1	Ille	egal				4
09Ch	C7DIR	C6DIR	C5DIR	C4DIR	C3DIR	C2DIR	C1DIR	C0DIR	PCDATDIR
	IOPC7	IOPC6	IOPC5	IOPC4	IOPC3	IOPC2	IOPC1	IOPC0	
709Dh			1	1	egal				4
709Eh	D7DIR	D6DIR	D5DIR	D4DIR	D3DIR	D2DIR	D1DIR	D0DIR	PDDATDIR
002	IOPD7	IOPD6	IOPD5	IOPD4	IOPD3	IOPD2	IOPD1	IOPD0	
709Fh				Ille	egal				
			ANALOG-TO	D-DIGITAL CON	VERTER (ADC)	REGISTERS			
	—	ADC S/W RESET	SOFT	FREE	ACQ PRESCALE3	ACQ PRESCALE2	ACQ PRESCALE1	ACQ PRESCALE0	
70A0h	CONV PRE- SCALE (CPS)	CONTIN- UOUS RUN	INT PRIORITY	SEQ1/2 CASCADE	CALIB EN	BRIDGE EN	HI / LO	FSTEST EN	ADCCTRL1
	EVB SOC EN SEQ1	Reset SEQ1 Start CALIB	SOC SEQ1	SEQ1 BUSY	INT ENA SEQ1 Mode1	INT ENA SEQ1 Mode0	INT FLAG SEQ1	EVA SOC EN SEQ1	
70A1h	EXT SOC EN SEQ1	Reset SEQ2	SOC SEQ2	SEQ2 BUSY	INT ENA SEQ2 Mode1	INT ENA SEQ2 Mode0	INT FLAG SEQ2	EVB SOC EN SEQ2	ADCCTRL2
	_	_	_	_	_		_	_	1
70A2h	_	MAXCONV2 2	MAXCONV2 1	MAXCONV2 0	MAXCONV1 3	MAXCONV1 2	MAXCONV1 1	MAXCONV1 0	MAXCONV
	CONV 3	CONV 3	CONV 3	CONV 3	CONV 2	CONV 2	CONV 2	CONV 2	
70A3h	CONV 1	CONV 1	CONV 1	CONV 1	CONV 0	CONV 0	CONV 0	CONV 0	CHSELSEQ1
70 4 41-	CONV 7	CONV 7	CONV 7	CONV 7	CONV 6	CONV 6	CONV 6	CONV 6	
70A4h	CONV 5	CONV 5	CONV 5	CONV 5	CONV 4	CONV 4	CONV 4	CONV 4	CHSELSEQ2
70A5h	CONV 11	CONV 11	CONV 11	CONV 11	CONV 10	CONV 10	CONV 10	CONV 10	CHSELSEQ3
UASII	CONV 9	CONV 9	CONV 9	CONV 9	CONV 8	CONV 8	CONV 8	CONV 8	CHSELSEQS
70A6h	CONV 15	CONV 15	CONV 15	CONV 15	CONV 14	CONV 14	CONV 14	CONV 14	CHSELSEQ4
TUAON	CONV 13	CONV 13	CONV 13	CONV 13	CONV 12	CONV 12	CONV 12	CONV 12	CH3EL3EQ4
		_	—	—	SEQ CNTR3	SEQ CNTR2	SEQ CNTR1	SEQ CNTR0	
70A7h	SEQ2 STATE 3	SEQ2 STATE 2	SEQ2 STATE 1	SEQ2 STATE 0	SEQ1 STATE 3	SEQ1 STATE 2	SEQ1 STATE 1	SEQ1 STATE 0	AUTO_SEQ_SR
	D9	D8	D7	D6	D5	D4	D3	D2	1
70A8h	D1	D0	0	0	0	0	0	0	RESULT0
	D9	D8	D7	D6	D5	D4	D3	D2	1
70A9h	D1	D0	0	0	0	0	0	0	RESULT1
	D9	D8	D7	D6	D5	D4	D3	D2	1
'0AAh	D1	D0	0	0	0	0	0	0	RESULT2



# peripheral register description (continued)

ADDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	250
ADDR	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
		AN	ALOG-TO-DIGI	TAL CONVERT	ER (ADC) REGIS	STERS (CONTIN	IUED)		
70ABh	D9	D8	D7	D6	D5	D4	D3	D2	RESULT3
TUABIT	D1	D0	0	0	0	0	0	0	RESULIS
70ACh	D9	D8	D7	D6	D5	D4	D3	D2	RESULT4
//UACII	D1	D0	0	0	0	0	0	0	RESULI4
70ADh	D9	D8	D7	D6	D5	D4	D3	D2	RESULT5
, OADII	D1	D0	0	0	0	0	0	0	RESOLIS
070AEh	D9	D8	D7	D6	D5	D4	D3	D2	RESULT6
	D1	D0	0	0	0	0	00	0	RESOLIO
070AFh	D9	D8	D7	D6	D5	D4	D3	D2	RESULT7
	D1	D0	0	0	0	0	0	0	RESOLIT
070B0h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT8
	D1	D0	0	0	0	0	0	0	RECOLIO
070B1h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT9
	D1	D0	0	0	0	0	0	0	
070B2h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT10
0700211	D1	D0	0	0	0	0	0	0	
070B3h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT11
	D1	D0	0	0	0	0	0	0	
070B4h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT12
0/00411	D1	D0	0	0	0	0	0	0	
070B5h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT13
	D1	D0	0	0	0	0	0	0	RECOLING
070B6h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT14
or oboin	D1	D0	0	0	0	0	0	0	
070B7h	D9	D8	D7	D6	D5	D4	D3	D2	RESULT15
	D1	D0	0	0	0	0	0	0	
070B8h									CALIBRATIC
-									
070B9h					llegal				
to 070FFh				I	liegai				
F		CONTRO	LLER AREA NI	ETWORK (CAN	) CONFIGURATI	ON CONTROL F	REGISTERS		-
[	_	_	_	_	—	—	_	_	┫
07100h	MD3	MD2	ME5	ME4	ME3	ME2	ME1	ME0	MDER
[	TA5	TA4	TA3	TA2	AA5	AA4	AA3	AA2	<b>-</b>
07101h	TRS5	TRS4	TRS3	TRS2	TRR5	TRR4	TRR3	TRR2	TCR
F	RFP3	RFP2	RFP1	RFP0	RML3	RML2	RML1	RML0	-
07102h	RMP3	RMP2	RMP1	RMP0	OPC3	OPC2	OPC1	OPC0	RCR
	_	—	SUSP	CCR	PDR	DBO	WUBA	CDR	1
07103h	ABO	STM	i –	i –	i –	—	MBNR1	MBNR0	MCR
	_	_	_	_	_	_	_	<u> </u>	−
07104h	BRP7	BRP6	BRP5	BRP4	BRP3	BRP2	BRP1	BRP0	BCR2

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)



### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

DDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	250
ואטכ	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
Ī	С	ONTROLLER A	REA NETWORI	(CAN) CONFI	GURATION CON	TROL REGIST	RS (CONTINUE	ED)	1
_	_	_	—	—	—	SBG	SJW1	SJW0	
)5h	SAM	TSEG1–3	TSEG1–2	TSEG1-1	TSEG1-0	TSEG2–2	TSEG2–1	TSEG2-0	BCR1
	_	—	_	_	_	_	_	FER	
)6h	BEF	SA1	CRCE	SER	ACKE	BO	EP	EW	ESR
, [	—	—	—	—	—	—	—	—	COD
)7h	—	—	SMA	CCE	PDA	_	RM	TM	GSR
	TEC7	TEC6	TEC5	TEC4	TEC3	TEC2	TEC1	TEC0	050
08h	REC7	REC6	REC5	REC4	REC3	REC2	REC1	REC0	CEC
	_	—	MIF5	MIF4	MIF3	MIF2	MIF1	MIF0	
09h	_	RMLIF	AAIF	WDIF	WUIF	BOIF	EPIF	WLIF	CAN_IFR
	MIL	—	MIM5	MIM4	MIM3	MIM2	MIM1	MIMO	
)Ah	EIL	RMLIM	AAIM	WDIM	WUIM	BOIM	EPIM	WLIM	CAN_IMR
	LAMI	_	—	LAM0-28	LAM0-27	LAM0-26	LAM0-25	LAM0-24	
0Bh	LAM0-23	LAM0-22	LAM0-21	LAM0-20	LAM0-19	LAM0-18	LAM0-17	LAM0-16	LAM0_H
	LAM0-15	LAM0-14	LAM0-13	LAM0-12	LAM0-11	LAM0-10	LAM0–9	LAM0–8	
0Ch	LAM0-7	LAM0-6	LAM0–5	LAM0-4	LAM0-3	LAM0-2	LAM0-1	LAM0-0	LAM0_L
	LAMI	—	_	LAM1-28	LAM1-27	LAM1-26	LAM1-25	LAM1-24	<b>1</b>
0Dh	LAM1-23	LAM1-22	LAM1-21	LAM1-20	LAM1-19	LAM1-18	LAM1-17	LAM1-16	LAM1_H
[	LAM1-15	LAM1-14	LAM1-13	LAM1-12	LAM1-11	LAM1-10	LAM1–9	LAM1–8	7
0Eh	LAM1-7	LAM1–6	LAM1–5	LAM1-4	LAM1–3	LAM1–2	LAM1-1	LAM1-0	LAM1_L
Fh			-	•	•		•	-	7
)				III	egal				
FFh									4
		1	1		e Object #0	1		1	4
200h	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	MSGIDOL
	IDL-7	IDL–6	IDL–5	IDL-4	IDL–3	IDL–2	IDL-1	IDL-0	
01h	IDE	AME	AAM	IDH–28	IDH–27	IDH–26	IDH–25	IDH–24	MSGID0H
	IDH–23	IDH-22	IDH–21	IDH–20	IDH–19	IDH–18	IDH-17	IDH–16	
202h	_	<u> </u>							MSGCTRLO
		—	_	RTR	DLC3	DLC2	DLC1	DLC0	_
203h				Res	erved				
204h	D15	D14	D13	D12	D11	D10	D9	D8	MBX0A
	D7	D6	D5	D4	D3	D2	D1	D0	
05h	D15	D14	D13	D12	D11	D10	D9	D8	MBX0B
	D7	D6	D5	D4	D3	D2	D1	D0	
206h	D15	D14	D13	D12	D11	D10	D9	D8	MBX0C
	D7	D6	D5	D4	D3	D2	D1	D0	
207h	D15	D14	D13	D12	D11	D10	D9	D8	MBX0D
	D7	D6	D5	D4	D3	D2	D1	D0	



ADDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	REG
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	
	c	ONTROLLER A	REA NETWOR	K (CAN) CONFI	GURATION CON	ITROL REGIST	ERS (CONTINU	ED)	
				Message	e Object #1				
070001	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	MOOIDAI
07208h	IDL-7	IDL-6	IDL-5	IDL-4	IDL–3	IDL-2	IDL-1	IDL-0	MSGID1L
070001	IDE	AME	AAM	IDH–28	IDH–27	IDH–26	IDH-25	IDH–24	MOOIDALL
07209h	IDH–23	IDH-22	IDH-21	IDH–20	IDH-19	IDH–18	IDH-17	IDH-16	MSGID1H
0700.41	—	—	—	—	—	—	—	—	MOOOTDLA
0720Ah	_	—	—	RTR	DLC3	DLC2	DLC1	DLC0	MSGCTRL
0720Bh				Res	erved				
070001	D15	D14	D13	D12	D11	D10	D9	D8	
0720Ch	D7	D6	D5	D4	D3	D2	D1	D0	MBX1A
070004	D15	D14	D13	D12	D11	D10	D9	D8	
0720Dh	D7	D6	D5	D4	D3	D2	D1	D0	MBX1B
	D15	D14	D13	D12	D11	D10	D9	D8	
0720Eh	D7	D6	D5	D4	D3	D2	D1	D0	MBX1C
	D15	D14	D13	D12	D11	D10	D9	D8	
0720Fh	D7	D6	D5	D4	D3	D2	D1	D0	MBX1D
ĺ			•	Message	Object #2		•	•	
	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	-
07210h	IDL-7	IDL-6	IDL–5	IDL-4	IDL–3	IDL-2	IDL-1	IDL-0	MSGID2L
	IDE	AME	AAM	IDH–28	IDH–27	IDH–26	IDH-25	IDH–24	-
07211h	IDH–23	IDH-22	IDH-21	IDH–20	IDH–19	IDH–18	IDH-17	IDH–16	MSGID2H
		_					_		-
07212h		_		RTR	DLC3	DLC2	DLC1	DLC0	MSGCTRL2
07213h			1	Res	erved				-
	D15	D14	D13	D12	D11	D10	D9	D8	-
07214h	D7	D6	D5	D4	D3	D2	D1	D0	MBX2A
	D15	D14	D13	D12	D11	D10	D9	D8	
07215h	D7	D6	D5	D4	D3	D2	D1	D0	MBX2B
	D15	D14	D13	D12	D11	D10	D9	D8	-
07216h	D7	D6	D5	D4	D3	D2	D1	D0	MBX2C
	D15	D14	D13	D12	D11	D10	D9	D8	-
07217h	D7	D6	D5	D4	D3	D2	D1	D0	MBX2D
		1		Message	Object #3		1	1	-
	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	_
07218h	IDL-7	IDL-6	IDL-5	IDL-4	IDL-3	IDL-2	IDL-1	IDL-0	MSGID3L
	IDE	AME	AAM	IDH-28	IDH-27	IDH-26	IDH-25	IDH-24	_
07219h	IDH–23	IDH-22	IDH–21	IDH-20	IDH-19	IDH–18	IDH-17	IDH-16	MSGID3H
		_	_			_	_		-
0721Ah		_	l _	RTR	DLC3	DLC2	DLC1	DLC0	MSGCTRL
0721Bh		<u>I</u>	1		erved	5102			-
, <u>,</u> , , <u>,</u> , , <u>, , , , </u>	D15	D14	D13	D12	D11	D10	D9	D8	-
0721Ch	D13	D14 D6	D13	D12	D3	D10	D3	D0	MBX3A
l	וט	00	00	U4	60	02		50	

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

21Dh	BIT 7	BIT 6	BIT 5	BIT 4					REG
⊢				BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	
⊢	D15	CONTROLLER	AREA NETWOR	K (CAN) CONFI	GURATION CON	NTROL REGISTE	ERS (CONTINUE	D)	
⊢	015	D14	D13	D12	D11	D10	D9	D8	МВХЗВ
	D7	D6	D5	D4	D3	D2	D1	D0	IVIDA3D
1 E h	D15	D14	D13	D12	D11	D10	D9	D8	МВХЗС
1Eh	D7	D6	D5	D4	D3	D2	D1	D0	WIDAGC
1Fh	D15	D14	D13	D12	D11	D10	D9	D8	MBX3D
۳Ľ	D7	D6	D5	D4	D3	D2	D1	D0	WIDX3D
L				Messag	e Object #4				
20h	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	MSGID4L
201	IDL-7	IDL–6	IDL-5	IDL-4	IDL-3	IDL-2	IDL-1	IDL-0	WISGID4L
	IDE	AME	AAM	IDH–28	IDH–27	IDH–26	IDH–25	IDH–24	MECIDALI
21h	IDH-23	IDH-22	IDH-21	IDH–20	IDH–19	IDH-18	IDH-17	IDH-16	MSGID4H
	_	—	—	—	—	—	—	-	MSGCTRL
222h	—	—	—	RTR	DLC3	DLC2	DLC1	DLC0	MISGUTRL
23h				Re	served				
	D15	D14	D13	D12	D11	D10	D9	D8	
24h	D7	D6	D5	D4	D3	D2	D1	D0	MBX4A
225h	D15	D14	D13	D12	D11	D10	D9	D8	MBX4B
22311	D7	D6	D5	D4	D3	D2	D1	D0	IVIDA4D
	D15	D14	D13	D12	D11	D10	D9	D8	MBX4C
26h	D7	D6	D5	D4	D3	D2	D1	D0	
27h	D15	D14	D13	D12	D11	D10	D9	D8	MBX4D
2/11	D7	D6	D5	D4	D3	D2	D1	D0	IVIDA4D
				Messag	e Object #5				
	IDL-15	IDL-14	IDL-13	IDL-12	IDL-11	IDL-10	IDL-9	IDL-8	
228h	IDL-7	IDL-6	IDL–5	IDL-4	IDL-3	IDL-2	IDL-1	IDL-0	MSGID5L
	IDE	AME	AAM	IDH–28	IDH–27	IDH–26	IDH–25	IDH–24	
229h	IDH-23	IDH-22	IDH–21	IDH-20	IDH–19	IDH–18	IDH-17	IDH-16	MSGID5H
	_	—	—	—	—	—	—	—	MOOOTDI
2Ah	_	-	—	RTR	DLC3	DLC2	DLC1	DLC0	MSGCTRL
2Bh				Re	served				
	D15	D14	D13	D12	D11	D10	D9	D8	
2Ch	D7	D6	D5	D4	D3	D2	D1	D0	MBX5A
	D15	D14	D13	D12	D11	D10	D9	D8	
2Dh	D7	D6	D5	D4	D3	D2	D1	D0	MBX5B
	D15	D14	D13	D12	D11	D10	D9	D8	
2Eh	D7	D6	D5	D4	D3	D2	D1	D0	MBX5C
	D15	D14	D13	D12	D11	D10	D9	D8	MDYED
2Fh	D7	D6	D5	D4	D3	D2	D1	D0	MBX5D
:30h		•		•			•		1
o IFFh				11	legal				



# peripheral register description (continued)

ADDR	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	
ADDR	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
		GENERA	L-PURPOSE (G	P) TIMER CONF	IGURATION CO	ONTROL REGIS	TERS – EVA	•	
	_	T2STAT	T1STAT	_	_	T2TC	ADC	T1TOADC(1)	1
07400h	T1TOADC(0)	TCOMPOE	- 1	_	T2	PIN	T.	1PIN	GPTCONA
	D15	D14	D13	D12	D11	D10	D9	D8	
07401h	D7	D6	D5	D4	D3	D2	D1	D0	T1CNT
074005	D15	D14	D13	D12	D11	D10	D9	D8	TIONED
07402h	D7	D6	D5	D4	D3	D2	D1	D0	T1CMPR
074005	D15	D14	D13	D12	D11	D10	D9	D8	TIDD
07403h	D7	D6	D5	D4	D3	D2	D1	D0	T1PR
074045	FREE	SOFT	-	TMODE1	TMODE0	TPS2	TPS1	TPS0	TIOON
07404h	—	TENABLE	TCLKS1	TCLKS0	TCLD1	TCLD0	TECMPR	—	T1CON
07405	D15	D14	D13	D12	D11	D10	D9	D8	TOOLIT
07405h	D7	D6	D5	D4	D3	D2	D1	D0	T2CNT
07400	D15	D14	D13	D12	D11	D10	D9	D8	TOOLIDD
07406h	D7	D6	D5	D4	D3	D2	D1	D0	T2CMPR
074075	D15	D14	D13	D12	D11	D10	D9	D8	TODD
07407h	D7	D6	D5	D4	D3	D2	D1	D0	T2PR
074005	FREE	SOFT	-	TMODE1	TMODE0	TPS2	TPS1	TPS0	TROOM
07408h	T2SWT1	TENABLE	TCLKS1	TCLKS0	TCLD1	TCLD0	TECMPR	SELT1PR	T2CON
07409h to 07410h			FULL AND		legal	STERS - EVA			-
								PDPINTA	-
07411h	CENABLE	CLD1	CLD0	SVENABLE	ACTRLD1	ACTRLD0	FCOMPOE	STATUS	COMCONA
	—	_	l –	—	_	—	_	—	
07412h				11	legal				
074405	SVRDIR	D2	D1	D0	CMP6ACT1	CMP6ACT0	CMP5ACT1	CMP5ACT0	
07413h	CMP4ACT1	CMP4ACT0	CMP3ACT1	CMP3ACT0	CMP2ACT1	CMP2ACT0	CMP1ACT1	CMP1ACT0	ACTRA
07414h				11	legal				
	—	_	_	_	DBT3	DBT2	DBT1	DBT0	
07415h	EDBT3	EDBT2	EDBT1	DBTPS2	DBTPS1	DBTPS0	i –	i —	DBTCONA
07416h			•		legal	•	•	•	
	D15	D14	D13	D12	D11	D10	D9	D8	1
	D7	D6	D5	D4	D3	D2	D1	D0	CMPR1
07417h		D14	D13	D12	D11	D10	D9	D8	0.4555
	D15	-	D5	D4	D3	D2	D1	D0	CMPR2
07417h 07418h	D15 D7	D6	05			<b>D</b> 10	D9	DO	1
07418h		D6 D14	D13	D12	D11	D10	Da	D8	
	D7			D12 D4	D11 D3	D10 D2	D3 D1	D8 D0	CMPR3
07418h	D7 D15	D14	D13	1				1	CMPR3
07418h 07419h	D7 D15	D14	D13	D4				1	CMPR3

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)



# Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

R	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	REG
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
				CAPTURE UNIT	REGISTERS -	EVA			
	CAPRES	CAP	QEPN	CAP3EN		CAP3TSEL	CAP12TSEL	CAP3TOADC	CAPCONA
	CAP1	EDGE	CAP2	2EDGE	CAPS	BEDGE		_	
			1	I	llegal				
	-	-	CAP	3FIFO	CAP	2FIFO	CAF	P1FIFO	CAPFIFOA
•	_	—		—		-		—	
	D15	D14	D13	D12	D11	D10	D9	D8	CAP1FIFO
•	D7	D6	D5	D4	D3	D2	D1	D0	4
	D15	D14	D13	D12	D11	D10	D9	D8	CAP2FIFO
•	D7	D6	D5	D4	D3	D2	D1	D0	-
	D15	D14	D13	D12	D11	D10	D9	D8	CAP3FIFO
•	D7	D6	D5	D4	D3	D2	D1	D0	-
-		544	D (a	T	llegal	5.40			
	D15	D14	D13	D12	D11	D10	D9	D8	CAP1FBOT
ŀ	D7	D6	D5	D4	D3	D2	D1	D0	-
	D15	D14	D13	D12	D11	D10	D9	D8	CAP2FBOT
	D7	D6	D5	D4	D3	D2	D1	D0	-
	D15 D7	D14 D6	D13 D5	D12 D4	D11 D3	D10 D2	D9 D1	D8 D0	CAP3FBOT
	DI	Do	D5	D4	03	DZ	DI	DU	-
				I	llegal				
									4
L		1	EVENT MAN	AGER (EVA) INT	ERRUPT CONT		S		_
	_				-	T1OFINT ENA	T1UFINT ENA	T1CINT ENA	EVAIMRA
	T1PINT ENA	—	_	_	CMP3INT ENA	CMP2INT ENA	CMP1INT ENA	PDPINTA ENA	
Ļ	_	—	-	-	-		-		
	—	_	_	_	T2OFINT ENA	T2UFINT ENA	T2CINT ENA	T2PINT ENA	EVAIMRB
Ļ	_		-		—	—	—	—	
L	—	_			_	CAP3INT ENA	CAP2INT ENA	CAP1INT ENA	EVAIMRC
	—	—	-	-	-	T1OFINT FLAG	T1UFINT FLAG	T1CINT FLAG	EVAIFRA
Ì	T1PINT FLAG	—	_	_	CMP3INT FLAG	CMP2INT FLAG	CMP1INT FLAG	PDPINTA FLAG	EVAIERA
	—	—	-	—	—	—	—	—	
	_	—	-	_	T2OFINT FLAG	T2UFINT FLAG	T2CINT FLAG	T2PINT FLAG	EVAIFRB
ſ	_	—	—	—	—	-	—	—	
	_	_	-	-	-	CAP3INT FLAG	CAP2INT FLAG	CAP1INT FLAG	EVAIFRC
Г			-	•	•	•	-	-	1
١				1	llegal				



# peripheral register description (continued)

	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	
ADDR	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
		GENERA	L-PURPOSE (G	P) TIMER CONF	IGURATION CO	NTROL REGIST	ERS – EVB		
	_	T4STAT	T3STAT	_	-	T4TO	ADC	T3TOADC(1)	
07500h	T3TOADC(0)	TCOMPOEB	-	-	T4	PIN	т	3PIN	GPTCONB
075041	D15	D14	D13	D12	D11	D10	D9	D8	TOOLIT
07501h	D7	D6	D5	D4	D3	D2	D1	D0	T3CNT
075006	D15	D14	D13	D12	D11	D10	D9	D8	T3CMPR
07502h	D7	D6	D5	D4	D3	D2	D1	D0	T3CMPR
075006	D15	D14	D13	D12	D11	D10	D9	D8	TODD
07503h	D7	D6	D5	D4	D3	D2	D1	D0	T3PR
075046	FREE	SOFT	_	TMODE1	TMODE0	TPS2	TPS1	TPS0	TROOM
07504h	—	TENABLE	TCLKS3	TCLKS0	TCLD1	TCLD0	TECMPR	—	T3CON
07505h	D15	D14	D13	D12	D11	D10	D9	D8	T4CNT
07505h	D7	D6	D5	D4	D3	D2	D1	D0	14CN1
OZEOCh	D15	D14	D13	D12	D11	D10	D9	D8	TACMOD
07506h	D7	D6	D5	D4	D3	D2	D1	D0	T4CMPR
07507h	D15	D14	D13	D12	D11	D10	D9	D8	T4PR
0750711	D7	D6	D5	D4	D3	D2	D1	D0	14FK
07509h	FREE	SOFT	_	TMODE1	TMODE0	TPS2	TPS1	TPS0	T4CON
07508h	T4SWT3	TENABLE	TCLKS1	TCLKS0	TCLD1	TCLD0	TECMPR	SELT3PR	14CON
07509h to				Po	served				
07510h				ite.	Serveu				
			FULL AND	SIMPLE COMP	ARE UNIT REG	STERS- EVB			
07511h	CENABLE	CLD1	CLD0	SVENABLE	ACTRLD1	ACTRLD0	FCOMPOEB	PDPINTB STATUS	COMCONB
	-	—	_	—	_	—	—	—	
07512h				Re	served				
075406	SVRDIR	D2	D1	D0	CMP12ACT1	CMP12ACT0	CMP11ACT1	CMP11ACT0	
07513h	CMP10ACT1	CMP10ACT0	CMP9ACT1	CMP9ACT0	CMP8ACT1	CMP8ACT0	CMP7ACT1	CMP7ACT0	ACTRB
07514h				Re	served				
075455	—	—	_	—	DBT3	DBT2	DBT1	DBT0	DETOONE
07515h	EDBT3	EDBT2	EDBT1	DBTPS2	DBTPS1	DBTPS0	—	—	DBTCONB
07516h		-	-	Re	served	-	-	-	
07547	D15	D14	D13	D12	D11	D10	D9	D8	
07517h	D7	D6	D5	D4	D3	D2	D1	D0	CMPR4
075405	D15	D14	D13	D12	D11	D10	D9	D8	OMPDE
07518h	D7	D6	D5	D4	D3	D2	D1	D0	CMPR5
075404	D15	D14	D13	D12	D11	D10	D9	D8	CMDDC
07519h	D7	D6	D5	D4	D3	D2	D1	D0	CMPR6
0751Ah				_					]
to 0751Fh				Re	served				
0101111									

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)



### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

, L	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	REG
	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
				CAPTURE UNIT	REGISTERS-E	EVB			
	CAPRES	CAP	QEPN	CAP6EN	—	CAP6TSEL	CAP45SEL	CAP6TOADC	
	CAP4	EDGE	CAPS	EDGE	CAP6	EDGE		_	CAFCON
				Re	served				
	-	_	CAP	6FIFO	CAP	5FIFO	CAF	24FIFO	CAPFIFO
	_	—			—	—	—	—	
	D15	D14	D13	D12	D11	D10	D9	D8	CAP4FIFC
L	D7	D6	D5	D4	D3	D2	D1	D0	
	D15	D14	D13	D12	D11	D10	D9	D8	CAP5FIF
	D7	D6	D5	D4	D3	D2	D1	D0	
	D15	D14	D13	D12	D11	D10	D9	D8	CAP6FIF
	D7	D6	D5	D4	D3	D2	D1	D0	
				Re	served				
	D15	D14	D13	D12	D11	D10	D9	D8	CAP4FBC
L	D7	D6	D5	D4	D3	D2	D1	D0	_
	D15	D14	D13	D12	D11	D10	D9	D8	CAP5FBC
L	D7	D6	D5	D4	D3	D2	D1	D0	_
	D15	D14	D13	D12	D11	D10	D9	D8	CAP6FBC
	D7	D6	D5	D4	D3	D2	D1	D0	
				Re	served				
					001100				
			EVENT MAN	AGER (EVB) INT	ERRUPT CONT	ROL REGISTER	S		
	—	—	—	—	_	T3OFINT ENA	T3UFINT ENA	T3CINT ENA	EVBIMRA
	T3PINT ENA	_	_	-	CMP6INT ENA	CMP5INT ENA	CMP4INT ENA	PDPINTB ENA	
	_	-	-		—	—	—		
	_	—			T4OFINT ENA	T4UFINT ENA	T4CINT ENA	T4PINT ENA	EVBIMRB
	_	-		—	—	—	—	—	
	_	_			—	CAP6INT ENA	CAP5INT ENA	CAP4INT ENA	EVBIMRC
	_	_	_	_	_	T3OFINT FLAG	T3UFINT FLAG	T3CINT FLAG	EVBIFRA
	T3PINT FLAG	-	-	-	CMP6INT FLAG	CMP5INT FLAG	CMP4INT FLAG	PDPINTB FLAG	EVDIFKA
	_	-	-	—	—	—	—	—	
	_	_	_	-	T4OFINT FLAG	T4UFINT FLAG	T4CINT FLAG	T4PINT FLAG	EVBIFRB
	_	—	_	—	—	—	—	—	
	_	—	-	-	-	CAP6INT FLAG	CAP5INT FLAG	CAP4INT FLAG	EVBIFRC
							-		1
				Pa	served				



## peripheral register description (continued)

	BIT 15	BIT 14	BIT 13	BIT 12	BIT 11	BIT 10	BIT 9	BIT 8	7
ADDR	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	REG
			PROGRA	M MEMORY SP	PACE – FLASH F	REGISTERS			1
0001	—	—	—	_	—	—	—	—	PMPC
0xx00h	—	_	—	_	PWR DWN	KEY1	KEY0	EXEC	PMPC
0xx01h	—	—	—	—	—	—	WSVER EN	PRECND Mode1	CTRL <sup>†</sup>
0000111	PRECND Mode0	ENG/R Mode2	ENG/R Mode1	ENG/R Mode0	FCM3	FCM2	FCM1	FCM0	CIRE
0xx02h									WADDR
0xx03h									WDATA
		_	_	_	_	_	_		1
0xx04h	_	— —	_	_	_	_	—	_	TCR
0xx05h	—	—	—	—	—	—	—	—	ENAB
0770311	_	_	_	_	_	_	_		LINAD
	_	—	—	_	—		—	_	
0xx06h	—	_	—	—	SECT 4 ENABLE	SECT 3 ENABLE	SECT 2 ENABLE	SECT 1 ENABLE	SECT
				I/O MEM	ORY SPACE	-			7
0FF0Fh	—	—	—	—	—	—	—	—	FCMR
UFFUFII		_	—	_	—	—	—		FCIVIK
			WAIT-S	TATE GENERAT	OR CONTROL	REGISTER			
0FFFFh	—		_	_	-	BVIS.1	BVIS.0	ISWS.2	WSGR
VI I I II	ISWS.1	ISWS.0	DSWS.2	DSWS.1	DSWS.0	PSWS.2	PSWS.1	PSWS.0	

### Table 18. LF240xA/LC240xA DSP Peripheral Register Description (Continued)

Indicates change with respect to the F243/F241, C242 device register maps.

<sup>†</sup>Register shown with bits set in **register mode**.



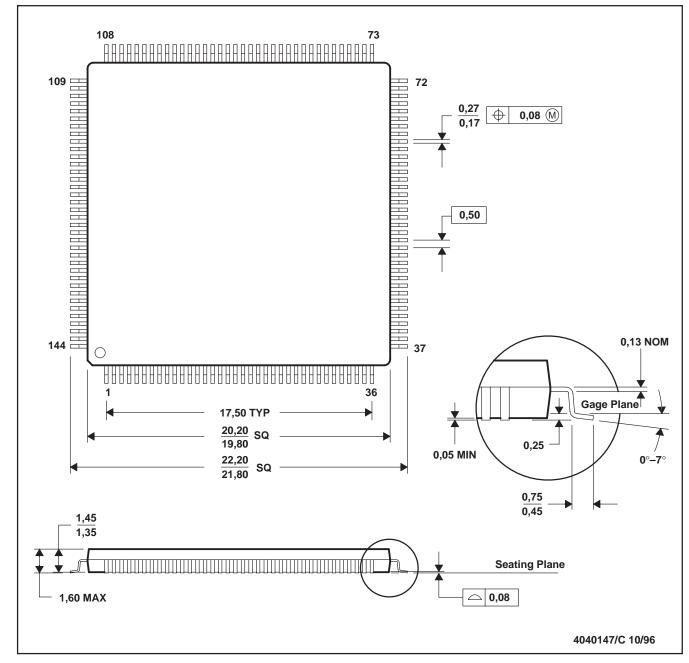
# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS

SPRS145B - JULY 2000 - REVISED DECEMBER 2000

**MECHANICAL DATA** 

### PGE (S-PQFP-G144)

#### PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-026

**Typical Thermal Resistance Characteristics** 

PARAMETER	DESCRIPTION	°C/W
ΘJA	Junction-to-ambient	32
ΘJC	Junction-to-case	8



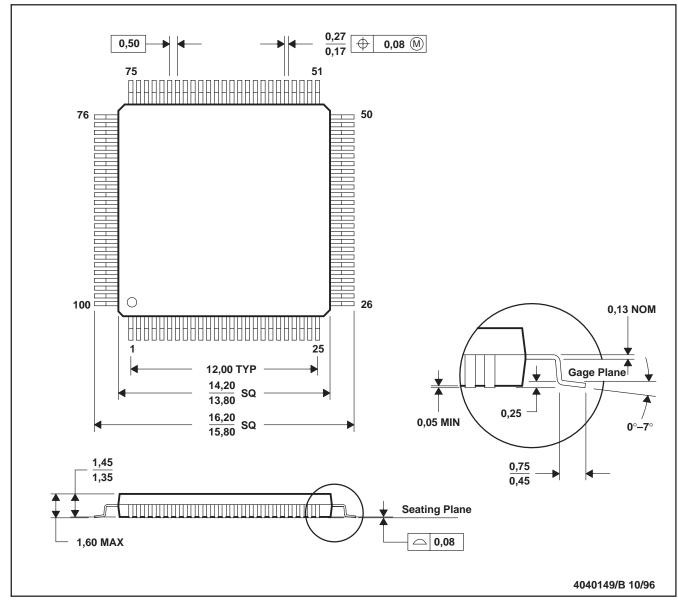
# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS

SPRS145B - JULY 2000 - REVISED DECEMBER 2000

MECHANICAL DATA

### PZ (S-PQFP-G100)

### PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026

Typical Thermal Resistance Characteristics

PARAMETER	DESCRIPTION	°C/W
$\Theta_{JA}$	Junction-to-ambient	42
ΘJC	Junction-to-case	8



**ADVANCE INFORMATION** 

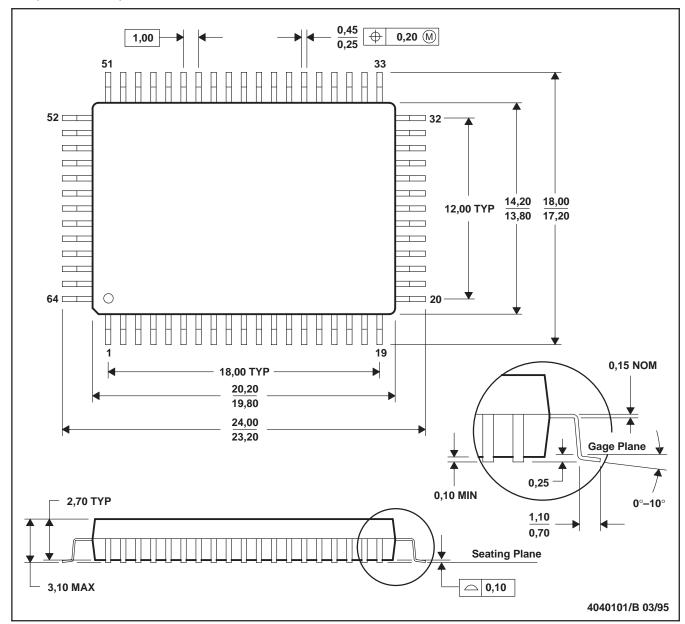
# TMS320LF2407A, TMS320LF2406A, TMS320LF2402A TMS320LC2406A, TMS320LC2404A, TMS320LC2402A DSP CONTROLLERS

SPRS145B - JULY 2000 - REVISED DECEMBER 2000

**MECHANICAL DATA** 

### PG (R-PQFP-G64)

PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Contact field sales office to determine if a tighter coplanarity requirement is available for this package.

**Typical Thermal Resistance Characteristics** 

PARAMETER	DESCRIPTION	°C/W
$\Theta_{JA}$	Junction-to-ambient	35
Θ <sup>JC</sup>	Junction-to-case	11



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