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- Floating Bootstrap or Ground-Reference High-Side Driver
- Adaptive Dead-Time Control
- 50-ns Max Rise/Fall Times With 3.3-nF Load
- 2.4-A Typical Output Current
- 4.5-V to 15-V Supply Voltage Range
- TTL-Compatible Inputs
- Internal Schottky Bootstrap Diode
- Low Supply Current....3 mA Typical
- Ideal for High-Current Single or Multiphase Power Supplies
- 40°C to 125°C Operating Virtual Junction-Temperature Range

#### description

The TPS2836 and TPS2837 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not have MOSFET drivers. The drivers are designed to deliver minimum 2-A peak currents into large capacitive loads. The high-side driver can be configured as ground-reference or as floating-bootstrap. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions and provides high efficiency for the buck regulator.

The TPS2836 has a noninverting input, while the TPS2837 has an inverting input. These drivers, available in 8-terminal SOIC packages, operate over a junction temperature range of  $-40^{\circ}$ C to 125°C.

	PACKAGED DEVICES
Tj	SOIC (D)
– 40°C to 125°C	TPS2836D TPS2837D

#### **AVAILABLE OPTIONS**

The D package is available taped and reeled. Add R suffix to device type (e.g., TPS2836DR)

#### **Related Synchronous MOS FET Drivers**

DEVICE NAME	ADDITIONAL FEATURES	INPUTS			
TPS2830		01100	Noninverted		
TPS2831	ENABLE, SYNC and CROWBAR	CMOS	Inverted		
TPS2832			Noninverted		
TPS2833	W/O ENABLE, SYNC and CROWBAR	CMOS	Inverted		
TPS2834			Noninverted		
TPS2835	ENABLE, SYNC and CROWBAR	TTL	Inverted		

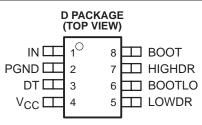


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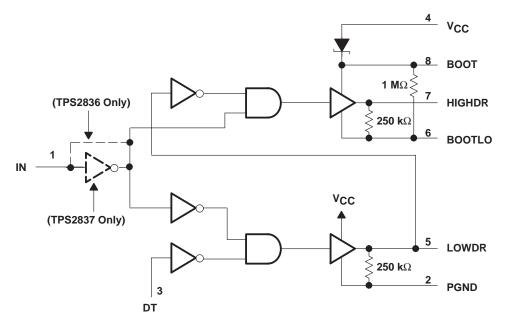


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# TPS2836, TPS2837 SYNCHRONOUS-BUCK MOSFET DRIVER WITH DEAD-TIME CONTROL SLVS224B – NOVEMBER 1999 – REVISED AUGUST 2002

## functional block diagram



### **Terminal Functions**

TERMIN	NAL		DECODIDITION
NAME	NO.	1/0	DESCRIPTION
BOOT	8	I	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 $\mu$ F and 1 $\mu$ F.
BOOTLO	6	0	This terminal connects to the junction of the high-side and low-side MOSFETs.
DT	3	I.	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs
HIGHDR	7	0	Output drive for the high-side power MOSFET
IN	1	Ι	Input signal to the MOSFET drivers (noninverting input for the TPS2836; inverting input for the TPS2837).
LOWDR	5	0	Output drive for the low-side power MOSFET
PGND	2		Power ground. Connect to the FET power ground.
VCC	4	Ι	Input supply. Recommended that a 1 $\mu\text{F}$ capacitor be connected from VCC to PGND.



#### detailed description

#### low-side driver

The low-side driver is designed to drive low r<sub>DS(on)</sub> N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

#### high-side driver

The high-side driver is designed to drive low  $r_{DS(on)}$  N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a ground-reference driver or a floating bootstrap driver. The internal bootstrap diode is a Schottky for improved drive efficiency. The maximum voltage that can be applied between the BOOT terminal and ground is 30 V.

#### dead-time (DT) control

Dead-time control prevents shoot-through current from flowing through the main power FETs during switching transitions by controlling the turnon times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the TTL-compatible DT terminal connects to the junction of the power FETs.

#### IN

The IN terminal is a TTL-compatible digital terminal that is the input control signal for the drivers. The TPS2836 has a noninverting input; the TPS2837 has an inverting input.

#### absolute maximum ratings over operating free-air temperature (unless otherwise noted)<sup>†</sup>

Supply voltage range, V <sub>CC</sub> (see Note 1)	) 16 V
Input voltage range: BOOT to PGND (high-side driver ON)	) 30 V
BOOTLO to PGND	) 16 V
BOOT to BOOTLO	) 16 V
IN	) 16 V
DT	) 30 V
Continuous total power dissipation See Dissipation Rating	Table
Operating virtual junction temperature range, T <sub>J</sub> –40°C to 1	125°C
Storage temperature range, T <sub>stg</sub>	150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. NOTE 1: Unless otherwise specified, all voltages are with respect to PGND.

## DISSIPATION RATING TABLE

PACKAGE	$T_A \le 25^{\circ}C$	DERATING FACTOR	T <sub>A</sub> = 70°C	T <sub>A</sub> = 85°C
	POWER RATING	ABOVE T <sub>A</sub> = 25°C	POWER RATING	POWER RATING
D	600 mW	6.0 mW/°C	330 mW	240 mW

### recommended operating conditions

		MIN	NOM MAX	UNIT
Supply voltage,	Vcc	4.5	15	V
Input voltage	BOOT to PGND	4.5	28	V



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#### electrical characteristics over recommended operating virtual junction temperature range, $V_{CC} = 6.5 \text{ V}, C_{L} = 3.3 \text{ nF}$ (unless otherwise noted)

#### supply current

	PARAMETER	TEST CC	MIN	TYP	MAX	UNIT	
	Supply voltage range			4.5		15	V
	Quiescent current	V <sub>CC</sub> =15 V,	V <sub>(ENABLE)</sub> = LOW			100	•
Vcc	Quiescent current	V <sub>CC</sub> =15 V,	V <sub>(ENABLE)</sub> = HIGH		300	400	μA
	Quiescent current	V <sub>CC</sub> =12 V, f <sub>SWX</sub> = 200 kHz, C <sub>HIGHDR</sub> = 50 pF,	BOOTLO grounded, C <sub>LOWDR</sub> = 50 pF, See Note 2		3		mA

NOTE 2: Ensured by design, not production tested.

#### output drivers

PARAMETER			TEST CONDIT	TIONS	MIN	TYP	MAX	UNIT			
		Duty cycle < 2%,	VBOOT - VBOOTLO = 4.5 V	, VHIGHDR = 4 V	0.7	1.1					
	High-side sink (see Note 4)	t <sub>pw</sub> < 100 μs	$V_{BOOT} - V_{BOOTLO} = 6.5 V_{POOTLO}$	, V <sub>HIGHDR</sub> = 5 V	1.1	1.5		А			
		(see Note 3)	VBOOT – VBOOTLO = 12 V,	V <sub>HIGHDR</sub> = 10.5 V	2	2.4		7			
	High-side	Duty cycle < 2%,	V <sub>BOOT</sub> - V <sub>BOOTLO</sub> = 4.5 V	, V <sub>HIGHDR</sub> = 0.5V	1.2	1.4					
	source	t <sub>pw</sub> < 100 μs	$V_{BOOT} - V_{BOOTLO} = 6.5 V_{e}$	, V <sub>HIGHDR</sub> = 1.5 V	1.3	1.6		А			
Peak output-	(see Note 4)	(see Note 3)	VBOOT - VBOOTLO = 12 V,	V <sub>HIGHDR</sub> = 1.5 V	2.3	2.7					
current		Duty cycle < 2%,	V <sub>CC</sub> = 4.5 V,	$V_{LOWDR} = 4 V$	1.3	1.8					
	Low-side sink (see Note 4)	t <sub>pw</sub> < 100 μs	V <sub>CC</sub> = 6.5 V,	$V_{LOWDR} = 5 V$	2	2.5		А			
		(see Note 3)	V <sub>CC</sub> = 12 V,	$V_{LOWDR} = 10.5 V$	3	3.5					
	Low-side source (see Note 4)	Duty cycle < 2%,	V <sub>CC</sub> = 4.5 V,	$V_{LOWDR} = 0.5V$	1.4	1.7					
		t <sub>pw</sub> < 100 μs	V <sub>CC</sub> = 6.5 V,	$V_{LOWDR} = 1.5 V$	2	2.4		А			
		(see Note 3)	V <sub>CC</sub> = 12 V,	$V_{LOWDR} = 1.5 V$	2.5	3					
			$V_{BOOT} - V_{BOOTLO} = 4.5 V_{BOOTLO}$	, V <sub>HIGHDR</sub> = 0.5 V			5	;			
	High-side sink (see Note 4)		$V_{BOOT} - V_{BOOTLO} = 6.5 V_{e}$	, V <sub>HIGHDR</sub> = 0.5 V			5	Ω			
			$V_{BOOT} - V_{BOOTLO} = 12 V,$	VHIGHDR = 0.5 V			5				
			VBOOT - VBOOTLO = 4.5 V	, V <sub>HIGHDR</sub> = 4 V			75				
	High-side source	e (see Note 4)	$V_{BOOT} - V_{BOOTLO} = 6.5 V_{e}$	, V <sub>HIGHDR</sub> = 6 V			75	Ω			
Output			$V_{BOOT} - V_{BOOTLO} = 12 V,$	VHIGHDR =11.5 V			75				
resistance			V <sub>DRV</sub> = 4.5 V,	$V_{LOWDR} = 0.5 V$			9				
	Low-side sink (se	ee Note 4)	V <sub>DRV</sub> = 6.5 V	$V_{LOWDR} = 0.5 V$			7.5	Ω			
			V <sub>DRV</sub> = 12 V,	$V_{LOWDR} = 0.5 V$			6				
			V <sub>DRV</sub> = 4.5 V,	$V_{LOWDR} = 4 V$			75				
	Low-side source	(see Note 4)	V <sub>DRV</sub> = 6.5 V,	$V_{LOWDR} = 6 V$			75	Ω			
			V <sub>DRV</sub> = 12 V,	V <sub>LOWDR</sub> = 11.5 V			75				

NOTES: 3. Ensured by design, not production tested.

4. The pullup/pulldown circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the rDS(on) of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.



# electrical characteristics over recommended operating virtual junction temperature range, $V_{CC} = 6.5 \text{ V}$ , $C_L = 3.3 \text{ nF}$ (unless otherwise noted) (continued)

#### dead-time

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input voltage			0.7V <sub>CC</sub>			V
$V_{\text{IL}}$	Low-level input voltage	LOWDR	Over the V <sub>CC</sub> range (see Note 3)			1	V
VIH	High-level input voltage	DT		2			V
$V_{IL}$	Low-level input voltage	וט	Over the V <sub>CC</sub> range			1	V

NOTE 3: Ensured by design, not production tested.

#### digital control terminals (IN)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{\text{IH}}$	High-level input voltage	Over the Vere range	2			V
VIL	Low-level input voltage	Over the V <sub>CC</sub> range			1	V

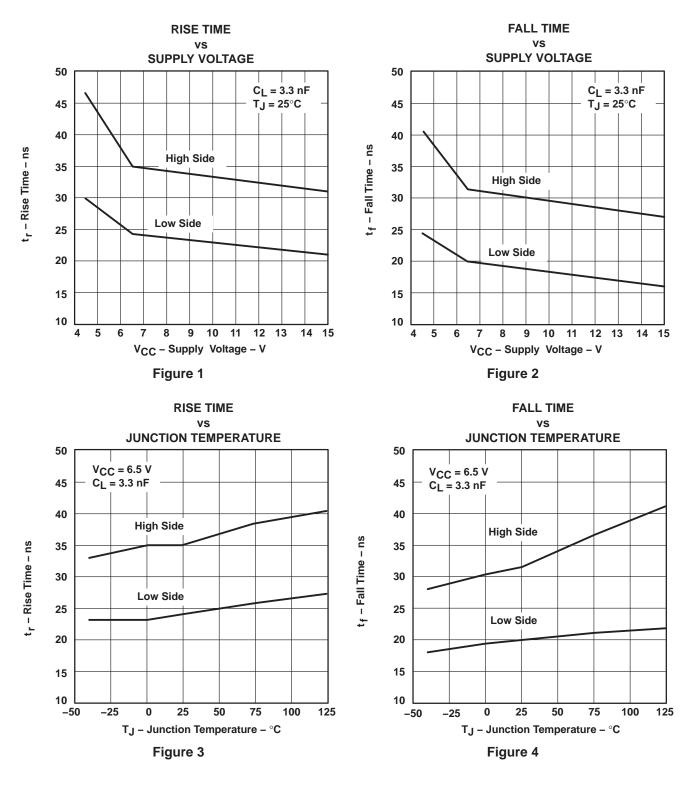
# switching characteristics over recommended operating virtual junction temperature range, $C_L$ = 3.3 nF (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT			
		V <sub>BOOT</sub> = 4.5 V,	V <sub>BOOTLO</sub> = 0 V			60				
Rise time	HIGHDR output (see Note 3)	V <sub>BOOT</sub> = 6.5 V,	V <sub>BOOTLO</sub> = 0 V			50	ns			
		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			50				
		$V_{CC} = 4.5 V$				40				
	LOWDR output (see Note 3)	$V_{CC} = 6.5 V$				30	ns			
		V <sub>CC</sub> = 12 V				30				
		V <sub>BOOT</sub> = 4.5 V,	V <sub>BOOTLO</sub> = 0 V			50				
	HIGHDR output (see Note 3)	V <sub>BOOT</sub> = 6.5 V,	V <sub>BOOTLO</sub> = 0 V			40	ns			
Fall time		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			40				
raii ume		$V_{CC} = 4.5 V$				40				
	LOWDR output (see Note 3)	V <sub>CC</sub> = 6.5 V				30	ns			
		V <sub>CC</sub> = 12 V				30				
		V <sub>BOOT</sub> = 4.5 V,	V <sub>BOOTLO</sub> = 0 V			95				
	HIGHDR going low (excluding dead- time) (see Note 3)	V <sub>BOOT</sub> = 6.5 V,	V <sub>BOOTLO</sub> = 0 V			80	ns			
Propagation delay time		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			65				
Propagation delay time	LOW/DD seine high (augludiae	V <sub>BOOT</sub> = 4.5 V,	V <sub>BOOTLO</sub> = 0 V			80				
	LOWDR going high (excluding dead-time) (see Note 3)	V <sub>BOOT</sub> = 6.5 V,	V <sub>BOOTLO</sub> = 0 V			70	ns			
		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			60				
		V <sub>CC</sub> = 4.5 V				80				
Propagation delay time	LOWDR going low (excluding dead- time) (see Note 3)	V <sub>CC</sub> = 6.5 V				70	ns			
		V <sub>CC</sub> = 12 V				60				
		V <sub>CC</sub> = 4.5 V		40		170				
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 3)	$V_{CC} = 6.5 V$		25		135	ns			
		V <sub>CC</sub> = 12 V		15		85				

NOTE 3: Ensured by design, not production tested.

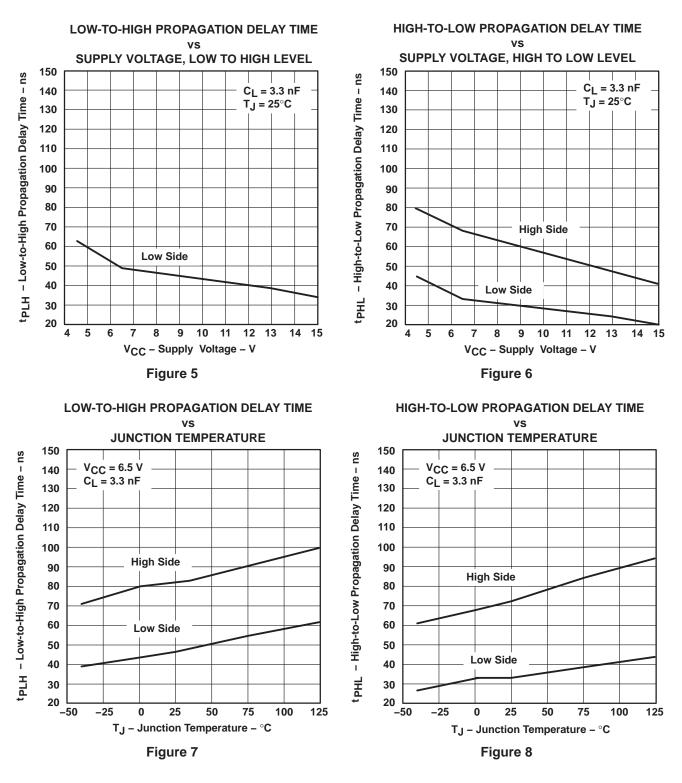


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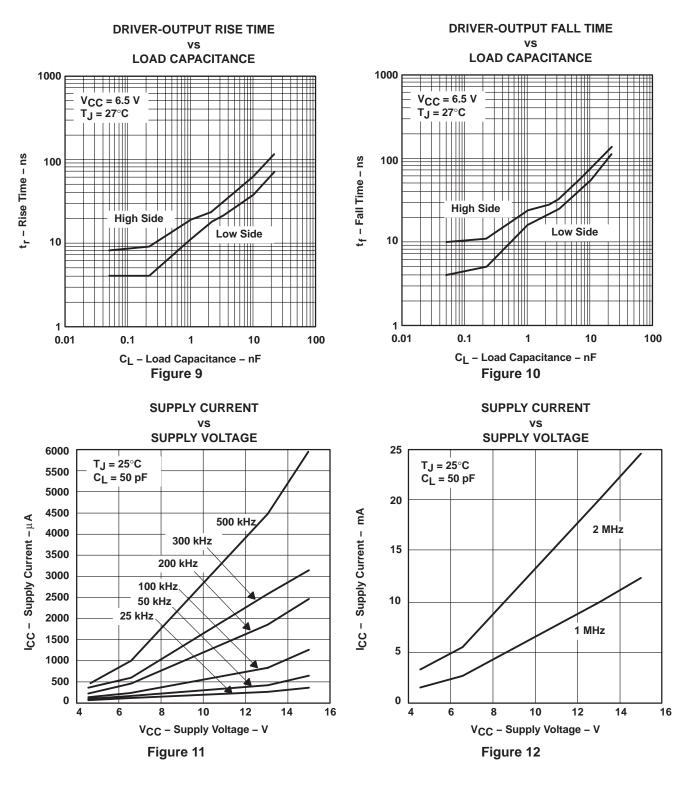


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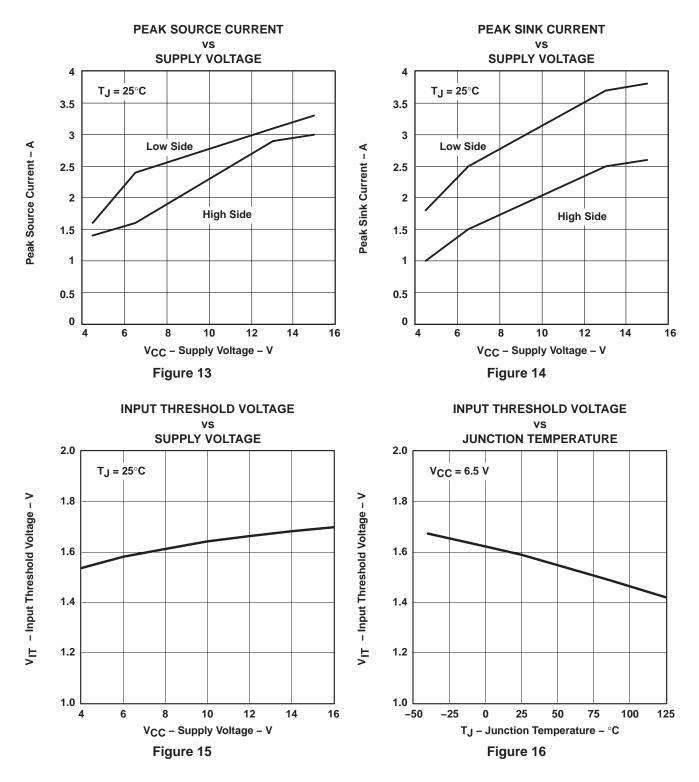




#### **TPS2836, TPS2837** SYNCHRONOUS-BUCK MOSFET DRIVER WITH DEAD-TIME CONTROL SLVS224B - NOVEMBER 1999 - REVISED AUGUST 2002









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## **APPLICATION INFORMATION**

Figure 17 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2837 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3-V output. The circuit can supply 3 A continuous load and the transient load is 5 A. The converter achieves an efficiency of 94% for  $V_{IN} = 5$  V,  $I_{load} = 1$  A, and 93% for  $V_{IN} = 5$  V,  $I_{load} = 3$  A.

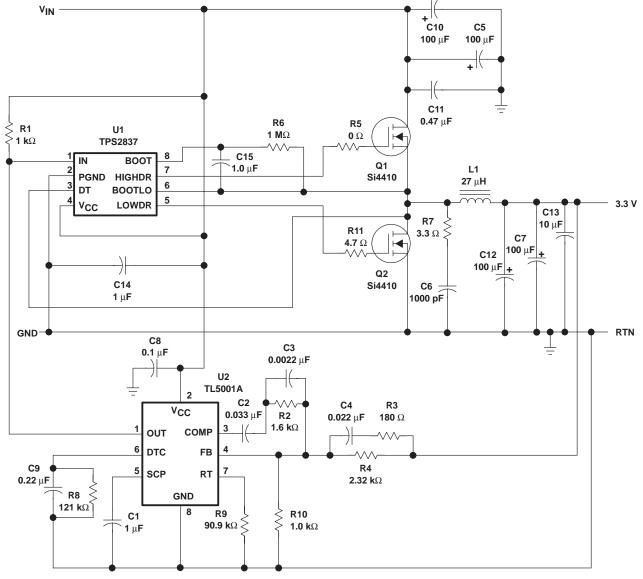


Figure 17. 3.3 V 3 A Synchronous-Buck Converter Circuit



## **APPLICATION INFORMATION**

Great care should be taken when laying out the PC board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across  $V_{CC}$  and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A). This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have other EMI problems and the power supply will be relatively free of noise.



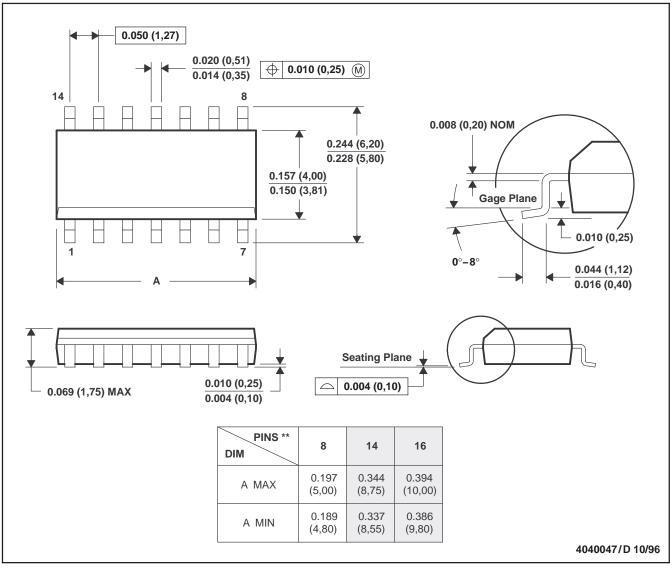
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**MECHANICAL DATA** 

#### D (R-PDSO-G\*\*)

#### PLASTIC SMALL-OUTLINE PACKAGE

#### **14 PIN SHOWN**



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-012





## PACKAGING INFORMATION

Orderable Device	Status	Package Type	•	Pins	Package	Eco Plan	Lead finish/	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	Ball material	(3)		(4/5)	
							(6)				
TPS2836D	LIFEBUY	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2836	
TPS2836DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2836	Samples
TPS2837D	LIFEBUY	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2837	
TPS2837DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2837	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW**: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION





### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



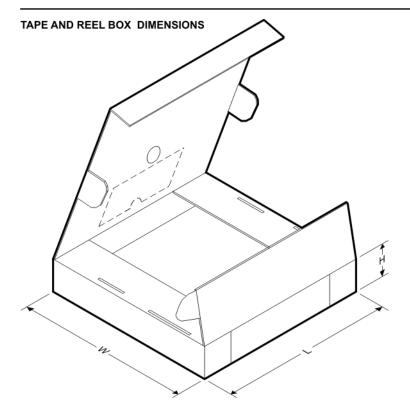
*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2836DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS2837DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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# PACKAGE MATERIALS INFORMATION

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2836DR	SOIC	D	8	2500	340.5	336.1	25.0
TPS2837DR	SOIC	D	8	2500	340.5	336.1	25.0



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## TUBE



#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
TPS2836D	D	SOIC	8	75	507	8	3940	4.32
TPS2837D	D	SOIC	8	75	507	8	3940	4.32

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