

application INFO available

UC1872 UC2872 UC3872

Resonant Lamp Ballast Controller

FEATURES

- Controls Different Types of Lamps: Cold Cathode Fluorescent, Neon, and Gas Discharge
- Zero Voltage Switching (ZVS) of Push-Pull Drivers
- Accurate Control of Lamp Current
- Variable Lamp Intensity Control
- 1µA Disable Current
- 4.5V to 24V Operation
- Open Lamp Detection Circuitry

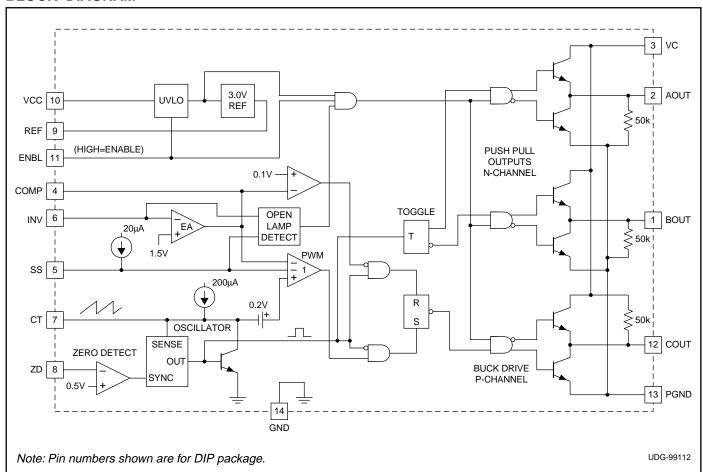
DESCRIPTION

The UC3872 is a resonant lamp ballast controller optimized for driving cold cathode fluorescent, neon, and other gas discharge lamps. The resonant power stage develops a sinusoidal lamp drive voltage, and minimizes switching loss and EMI generation. Lamp intensity adjustment is accomplished with a buck regulator, which is synchronized to the external power stage's resonant frequency. Suitable for automotive and battery powered applications, the UC3872 draws only 1µA when disabled.

Soft start and open lamp detect circuitry have been incorporated to minimize component stresses. Open lamp detection is enabled at the completion of a soft start cycle. The chip is optimized for smooth duty cycle control to 100%.

Other features include a precision 1.2% reference, undervoltage lockout, and accurate minimum and maximum frequency control.

BLOCK DIAGRAM

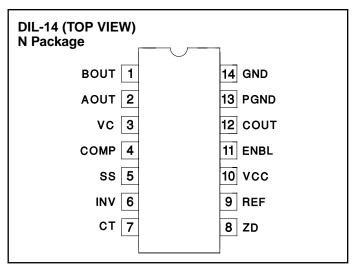


ABSOLUTE MAXIMUM RATINGS

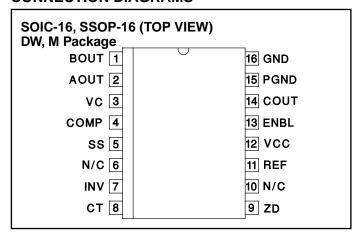
Analog Inputs
VCC, VC Voltage
ZD Input Current
High Impedance Source +10mA
ZD Input Voltage
Low Impedance Source +24V
Power Dissipation at TA = 25° C
Storage Temperature
Lead Temperature300° C

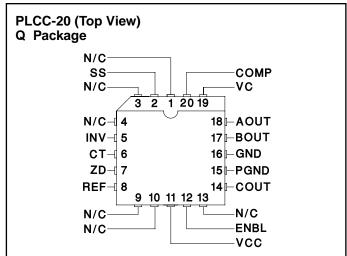
Note 1: Currents are positive into, negative out of the specified terminal.

Note 2: Consult Packaging Section of Databook for thermal limitations and considerations of package.



CONNECTION DIAGRAMS





ELECTRICAL CHARACTERISTICS: Unless otherwise stated, these parameters apply for T_J = -55°C to +125°C for the UC1872, -40°C to +85°C for the UC2872, -0°C to +70°C for the UC3872; VCC= 5V, VC = 15V, V_{ENBL} = 5V, CT = 1nF, ZD = 1V.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Reference Section					
Output Voltage	T _J = 25° C	2.963	3.000	3.037	V
	Over Temperature	2.940	3.000	3.060	V
Line Regulation	VCC = 4.75V to 18V			10	mV
Load Regulation	$I_O = 0$ to -5 mA			10	mV
Oscillator Section					
Free Running Frequency	T _J = 25° C	57	68	78	kHz
Maximum Synchronization Frequency	T _J = 25° C	160	200	240	kHz
Charge Current	V _{CT} = 1.5V	180	200	220	μΑ
Voltage Stability				2	%
Temperature Stability			4	8	%
Zero Detect Threshold		0.46	0.5	0.56	V
Error Amp Section					
Input Voltage	$V_0 = 2V$	1.445	1.475	1.505	V
Input Bias Current			-0.4	-2	μΑ
Open Loop Gain	$V_0 = 0.5 \text{ to } 3V$	65	90		dB
Output High	V _{INV} = 1.3V	3.1	3.5	3.9	V

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PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Error Amp Section (cont.)			1		
Output Low	V _{INV} = 1.7V		0.1	0.2	V
Output Source Current	$V_{INV} = 1.3V, V_{O} = 2V$	-350	-500		μА
Output Sink Current	$V_{INV} = 1.7V, V_{O} = 2V$	10	20		mA
Common Mode Range		0		V _{IN} -1V	V
Unity Gain Bandwidth	T _J = 25° C (Note 4)		1		MHz
Open Lamp Detect Section				1	
Soft Start Threshold	V _{INV} = 0V	2.9	3.4	3.8	V
Open Lamp Detect Threshold	V _{SS} = 4.2V	0.6	1.0	1.4	V
Soft Start Current	V _{SS} = 2V	10	20	40	μΑ
Output Section		•			
Output Low Level	I _{OUT} = 0, Outputs A and B		0.05	0.2	V
	I _{OUT} = 10mA		0.1	0.4	V
	I _{OUT} = 100mA		1.5	2.2	V
Output High Level	I _{OUT} = 0, Output C	13.9	14.9		V
	$I_{OUT} = -10 \text{mA}$	13.5	14.3		V
	$I_{OUT} = -100 \text{mA}$	12.5	13.5		V
Rise Time	T _J = 25° C, Cl = 1nF (Note 4)		30	80	ns
Fall Time	T _J = 25° C, CI = 1nF (Note 4)		30	80	ns
Output Dynamics					
Out A and B Duty Cycle		48	49.9	50	%
Out C Max Duty Cycle	$V_{INV} = 1V$	100			%
Out C Min Duty Cycle	$V_{INV} = 2V$			0	%
Under Voltage Lockout Section					
Startup Threshold Voltage		3.7	4.2	4.5	V
Hysteresis		120	200	280	mV
Enable Section					
Input High Threshold		2			V
Input Low Threshold				0.8	V
Input Current	V _{ENBL} = 5V		150	400	μΑ
Supply Current Section					
VCC Supply Current	VCC = 24V		6	14	mA
VC Supply Current	VC = 24V		5	12	mA
ICC Disabled	VCC = 24V, VENBL = 0V		1	10	μΑ

Note 3: Unless otherwise specified, all voltages are with respect to ground. Currents are positive into, and negative out of the specified terminal.

Note 4: Guaranteed by design. Not 100% tested in production.

PIN DESCRIPTIONS

AOUT, BOUT: These outputs provide complementary drive signals for the push-pull N-channel MOSFETs. Each one is high for 50% of the time, switching states each time a zero-detect is sensed.

COMP: COMP is the output terminal of the error amplifier. Compensation components are normally connected between COMP and INV. Connecting a capacitor from this pin to ground limits turn on current and blanks the open lamp detect signal allowing the lamp to start.

COUT: This output directly drives the bulk regulator P-channel MOSFET. COUT turn-on is synchronized to each zero-detect, and therefore switches at twice the frequency of AOUT and BOUT. The modulator controlling COUT is designed to provide smooth control up to 100% duty cycle.

CT: A capacitor connected between this pin and GND ground sets the synchronization frequency range. The capacitor is charged with approximately $200\mu A$, creating a linear ramp which is used by COUT's (buck regulator driver) PWM comparator.

ENBL: When ENBL is driven high the device is enabled. When ENBL is pulled low, the IC is shut down and typically draws 1μ A.

GND: This pin is the ground reference point for the internal reference and all thresholds.

INV: This pin is the inverting input to the error amplifier and the input for the open lamp detect circuitry. If the voltage at INV is below the 1V open lamp detect threshold, the outputs are disabled.

PGND: This pin is the high current ground connection for the three output drivers.

REF: This pin is connected to the 3V reference voltage which is used for the internal logic. Bypass REF to ground with a $0.01\mu F$ ceramic capacitor for proper operation.

VC: VC is the power supply voltage connection for the output drivers. Bypass it to ground with a $0.1\mu F$ ceramic capacitor for proper operation.

VCC: VCC is the positive supply voltage for the chip. Its operating range is from 4.2V to 24V. Bypass VCC to ground with a $0.1\mu F$ ceramic capacitor for proper operation.

ZD: The zero-detect input senses when the transformer's primary center tap voltage falls to zero to synchronize the sawtooth voltage waveform on CT. The threshold is approximately 0.5V, providing a small amount of offset such that with propagation delay, zero-volt switching occurs. A resistor (typically 10k) should be connected between ZD and the primary center tap to limit input current at turn off.

APPLICATION INFORMATION

Figure 1 shows a complete application circuit using the UC3872 Resonant Lamp Ballast Controller. The IC provides all drive, control and housekeeping functions. The buck output voltage (transformer center-tap) provides the zero crossing and synchronization signals.

The buck modulator drives a P-channel MOSFET directly, and operates over a 0-100% duty-cycle range. The modulation range includes 100%, allowing operation with minimal headroom.

The oscillator and synchronization circuitry are shown in Figure 2. The oscillator is designed to synchronize over a 3:1 frequency range. In an actual application however, the frequency range is only about 1.5:1. A zero detect comparator senses the primary center-tap voltage, generating a synchronization pulse when the resonant wave-

form falls to zero. The actual threshold is 0.5 volts, providing a small amount of anticipation to offset propagation delay.

The synchronization pulse width is the time required for the 4mA current sink to discharge the timing capacitor to 0.1 volts. This pulse width limits the minimum linear control range of the buck regulator. The 200µA current source charges the capacitor to a maximum of 3 volts. A comparator blanks the zero detect signal until the capacitor voltage exceeds 1 volt, preventing multiple synchronization pulse generation and setting the maximum frequency. If the capacitor voltage reaches 3 volts (a zero detection has not occurred) an internal clock pulse is generated to limit the minimum frequency.

APPLICATION INFORMATION (cont.)

A unique protection feature incorporated in the UC3872 is the Open Lamp Detect circuit. An open lamp interrupts the current feedback loop and causes very high secondary voltage. Operation in this mode will usually breakdown the transformer's insulation, causing permanent damage to the converter. The open lamp detect circuit, shown in Figure 3 senses the lamp current feedback signal at the error amplifier's input, and shuts down the outputs if insufficient signal is present. Soft start circuitry limits initial turn-on currents and blanks the open lamp detect signal.

Other features are included to minimize external circuitry requirements. A logic level enable pin shuts down the IC, allowing direct connection to a battery. During shutdown, the IC typically draws less than $1\mu A.$ The UC3872, operating from 4.5V to 24V, is compatible with almost all battery voltages used in portable computers and automotive applications. Undervoltage lockout circuitry disables operation until sufficient supply voltage is available, and a 1% voltage reference insures accurate operation.

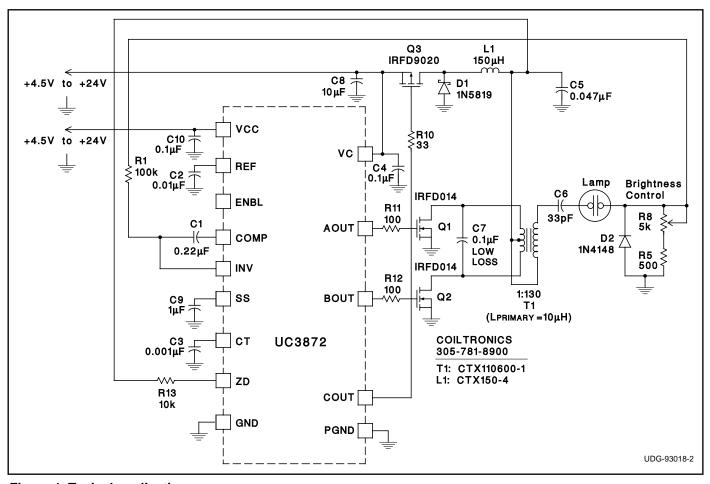


Figure 1. Typical application.

APPLICATIONS INFORMATION

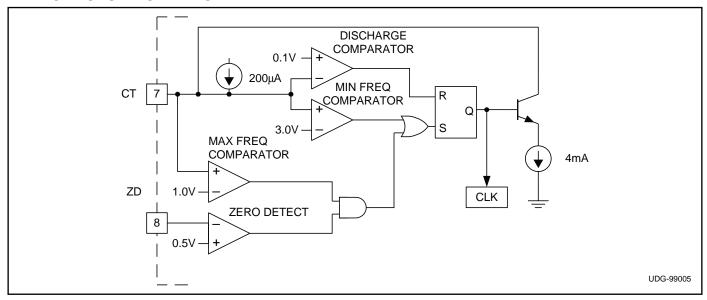


Figure 2. UC3872 oscillator section.

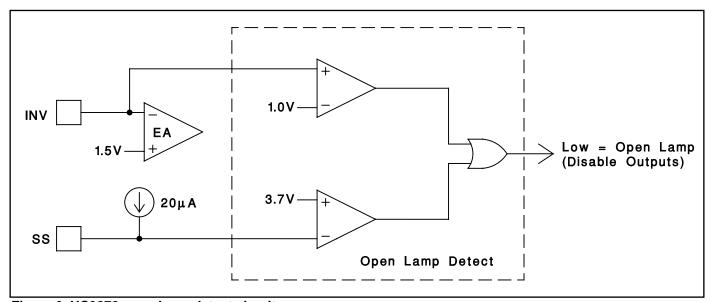


Figure 3. UC3872 open lamp detect circuitry.

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PRODUCT SUPPORT: APPLICATIONS

UC3872, Resonant Lamp Ballast Controller

DEVICE STATUS: ACTIVE

PARAMETER NAME	<u>UC2872</u>	UC3872
Vin (max) (V)	24	24
Vin (min) (V)	4.5	4.5
Vref tol (%)	1.2	1.2
Pin Count	14	14
Control Topology	Push-Pull	Push-Pull
Open Lamp Detect	Yes	Yes
External Sync	Yes	Yes
Frequency (max) (kHz)	200	200
Shutdown Current (mA)	1	1
Supply Current (mA)	11	11
Lamp Intensity Control	Yes	Yes
Soft Start	Yes	Yes

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TECHNICAL RESOURCES

Product Folder: UC3872, Resonant Lamp Ballast Controller

To download a document to your hard drive, right-click on the link and choose 'Save'.

DATASHEET Back to Top

Full datasheet in Acrobat PDF: uc3872.pdf (190 KB) (Updated: 09/05/1999)

APPLICATION NOTES

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- DN-75 Using the UC3871 and UC3872 Resonant Fluorescent Lamp Drivers in Floating (SLUA090 Updated: 09/05/1999)
- U-141 Resonant Fluorescent Lamp Converter Provides Efficient and Compact (SLUA056 Updated: 09/05/1999)
- U-148 Dimmable Cold-Cathode Fluorescent Lamp Ballast Design Using the UC3871 (SLUA088 Updated: 09/05/1999)

RELATED DOCUMENTS

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- Military Analog Selection Guide (SGLB002, 318 KB Updated: 11/09/2000)
- Military Semiconductors Selection Guide 2002 (Rev. B) (SGYC003B, 1648 KB Updated: 04/22/2002)

BLOCK DIAGRAMS

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Generic Web Access Device

SAMPLES	<u>▲Back to Top</u>					
ORDERABLE DEVICE	<u>PACKAGE</u>	<u>PINS</u>	TEMP (°C)	<u>STATUS</u>	<u>SAMPLES</u>	
UC3872DW	<u>DW</u>	16	0 TO 70	ACTIVE	Request Samples	
UC3872M	DBQ	16	0 TO 70	ACTIVE	Request Samples	
UC3872N	<u>N</u>	14	0 TO 70	ACTIVE	Request Samples	

PRICING/AVAILABIL	RICING/AVAILABILITY/PKG				<u> </u>		
ORDERABLE DEVICE	PACKAGE	<u>PINS</u>	TEMP (°C)	STATUS	BUDGETARY PRICE US\$/UNIT QTY=1000+	PACK QTY	PRICING/AVAILABILITY/PKG
UC3872DW	<u>DW</u>	16	0 TO 70	ACTIVE	2.43	40	Check stock or order
UC3872DWTR	<u>DW</u>	16	0 TO 70	ACTIVE	2.43	2000	Check stock or order
UC3872M	<u>DBQ</u>	16	0 TO 70	ACTIVE	2.43	90	Check stock or order
UC3872MTR	<u>DBQ</u>	16	0 TO 70	ACTIVE	2.43	2500	Check stock or order
UC3872N	<u>N</u>	14	0 TO 70	ACTIVE	2.43	25	Check stock or order
UC3872Q	<u>FN</u>	20	0 TO 70	ACTIVE	2.43	46	Check stock or order
UC3872QTR	<u>FN</u>	20	0 TO 70	ACTIVE	2.43	1000	Check stock or order

Table Data Updated on: 7/25/2002

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